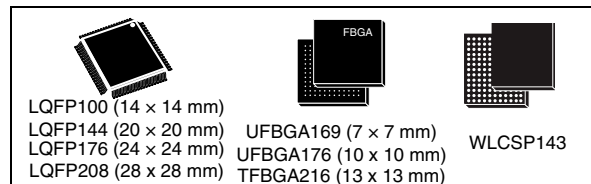


ARM Cortex-M4 32b MCU+FPU, 225DMIPS, up to 2MB Flash/256+4KB RAM, crypto, USB OTG HS/FS, Ethernet, 17 TIMs, 3 ADCs, 20 comm. interfaces, camera&LCD-TFT

Datasheet - production data

## Features

- Core: ARM 32-bit Cortex™-M4 CPU with FPU, Adaptive real-time accelerator (ART Accelerator™) allowing 0-wait state execution from Flash memory, frequency up to 180 MHz, MPU, 225 DMIPS/1.25 DMIPS/MHz (Dhrystone 2.1), and DSP instructions
- Memories
  - Up to 2 MB of Flash memory organized into two banks allowing read-while-write
  - Up to 256+4 KB of SRAM including 64-KB of CCM (core coupled memory) data RAM
  - Flexible external memory controller with up to 32-bit data bus: SRAM,PSRAM,SDRAM/LPSDR SDRAM , Compact Flash/NOR/NAND memories
- LCD parallel interface, 8080/6800 modes
- LCD-TFT controller up to SVGA resolution with dedicated Chrom-ART Accelerator™ for enhanced graphic content creation (DMA2D)
- Clock, reset and supply management
  - 1.7 V (PDR OFF) or 1.8 V to 3.6 V application supply and I/Os
  - POR, PDR, PVD and BOR
  - 4-to-26 MHz crystal oscillator
  - Internal 16 MHz factory-trimmed RC (1% accuracy)
  - 32 kHz oscillator for RTC with calibration
  - Internal 32 kHz RC with calibration
- Low power
  - Sleep, Stop and Standby modes
  - V<sub>BAT</sub> supply for RTC, 20×32 bit backup registers + optional 4 KB backup SRAM
- 3×12-bit, 2.4 MSPS ADC: up to 24 channels and 7.2 MSPS in triple interleaved mode
- 2×12-bit D/A converters
- General-purpose DMA: 16-stream DMA controller with FIFOs and burst support
- Up to 17 timers: up to twelve 16-bit and two 32-bit timers up to 180 MHz, each with up to 4 IC/OC/PWM or pulse counter and quadrature (incremental) encoder input
- Debug mode
  - SWD & JTAG interfaces



- Cortex-M4 Trace Macrocell™
- Up to 168 I/O ports with interrupt capability
  - Up to 164 fast I/Os up to 90 MHz
  - Up to 166 5 V-tolerant I/Os
- Up to 21 communication interfaces
  - Up to 3 × I<sup>2</sup>C interfaces (SMBus/PMBus)
  - Up to 4 USARTs/4 UARTs (11.25 Mbit/s, ISO7816 interface, LIN, IrDA, modem control)
  - Up to 6 SPIs (45 Mbits/s), 2 with muxed full-duplex I<sup>2</sup>S for audio class accuracy via internal audio PLL or external clock
  - 1 × SAI (serial audio interface)
  - 2 × CAN (2.0B Active) and SDIO interface
- Advanced connectivity
  - USB 2.0 full-speed device/host/OTG controller with on-chip PHY
  - USB 2.0 high-speed/full-speed device/host/OTG controller with dedicated DMA, on-chip full-speed PHY and ULPI
  - 10/100 Ethernet MAC with dedicated DMA: supports IEEE 1588v2 hardware, MII/RMII
- 8- to 14-bit parallel camera interface up to 54 Mbytes/s
- Cryptographic acceleration: hardware acceleration for AES 128, 192, 256, Triple DES, HASH (MD5, SHA-1, SHA-2), and HMAC
- True random number generator
- CRC calculation unit
- RTC: subsecond accuracy, hardware calendar
- 96-bit unique ID

Table 1. Device summary

Reference	Part number
STM32F437xx	STM32F437VG, STM32F437ZG, STM32F437IG, STM32F437VI, STM32F437ZI, STM32F437II, STM32F437AI
STM32F439xx	STM32F439VI, STM32F439VG, STM32F439ZG, STM32F439ZI, STM32F439IG, STM32F439II, STM32F439BG, STM32F439BI, STM32F439NI, STM32F439AI, STM32F439NG

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# 1 Introduction

This datasheet provides the description of the STM32F437xx and STM32F439xx line of microcontrollers. For more details on the whole STMicroelectronics STM32™ family, please refer to [Section 2.1: Full compatibility throughout the family](#).

The STM32F437xx and STM32F439xx datasheet should be read in conjunction with the STM32F4xx reference manual.

For information on the Cortex™-M4 core, please refer to the Cortex™-M4 programming manual (PM0214), available from the [www.st.com](http://www.st.com).

## 2 Description

The STM32F437xx and STM32F439xx devices are based on the high-performance ARM® Cortex™-M4 32-bit RISC core operating at a frequency of up to 180 MHz. The Cortex-M4 core features a Floating point unit (FPU) single precision which supports all ARM single-precision data-processing instructions and data types. It also implements a full set of DSP instructions and a memory protection unit (MPU) which enhances application security.

The STM32F437xx and STM32F439xx devices incorporate high-speed embedded memories (Flash memory up to 2 Mbyte, up to 256 Kbytes of SRAM), up to 4 Kbytes of backup SRAM, and an extensive range of enhanced I/Os and peripherals connected to two APB buses, two AHB buses and a 32-bit multi-AHB bus matrix.

All devices offer three 12-bit ADCs, two DACs, a low-power RTC, twelve general-purpose 16-bit timers including two PWM timers for motor control, two general-purpose 32-bit timers, a true random number generator (RNG), and a cryptographic acceleration cell. They also feature standard and advanced communication interfaces.

- Up to three I<sup>2</sup>Cs
- Six SPIs, two I<sup>2</sup>Ss full duplex. To achieve audio class accuracy, the I<sup>2</sup>S peripherals can be clocked via a dedicated internal audio PLL or via an external clock to allow synchronization.
- Four USARTs plus four UARTs
- An USB OTG full-speed and a USB OTG high-speed with full-speed capability (with the ULPI),
- Two CANs
- One SAI serial audio interface
- An SDIO/MMC interface
- Ethernet and the camera interface
- LCD-TFT display controller
- Chrom-ART Accelerator™.

Advanced peripherals include an SDIO, a flexible memory control (FMC) interface, a camera interface for CMOS sensors and a cryptographic acceleration cell. Refer to [Table 2: STM32F437xx and STM32F439xx features and peripheral counts](#) for the list of peripherals available on each part number.

The STM32F437xx and STM32F439xx devices operates in the –40 to +105 °C temperature range from a 1.8 to 3.6 V power supply.

The supply voltage can drop to 1.7 V with the use of an external power supply supervisor (refer to [Section 3.17.2: Internal reset OFF](#)). A comprehensive set of power-saving mode allows the design of low-power applications.

The STM32F437xx and STM32F439xx devices offers devices in 8 packages ranging from 100 pins to 216 pins. The set of included peripherals changes with the device chosen.

These features make the STM32F437xx and STM32F439xx microcontrollers suitable for a wide range of applications:

- Motor drive and application control
- Medical equipment
- Industrial applications: PLC, inverters, circuit breakers
- Printers, and scanners
- Alarm systems, video intercom, and HVAC
- Home audio appliances

Figure 4 show the general block diagram of the device family.

**Table 2. STM32F437xx and STM32F439xx features and peripheral counts**

Peripherals		STM32F437 Vx		STM32F439 Vx		STM32F437Zx		STM32F437AI	STM32F439AI	STM32F439Zx		STM32F437Ix		STM32F439Ix		STM32F439Bx		STM32F439Nx	
		1024	2048	1024	2048	1024	2048	2048	2048	1024	2048	1024	2048	1024	2048	1024	2048	1024	2048
Flash memory in Kbytes		1024	2048	1024	2048	1024	2048	2048	2048	1024	2048	1024	2048	1024	2048	1024	2048	1024	2048
SRAM in Kbytes	System	256(112+16+64+64)																	
	Backup	4																	
FMC memory controller		Yes <sup>(1)</sup>																	
Ethernet		Yes																	
Timers	General-purpose	10																	
	Advanced-control	2																	
	Basic	2																	
Random number generator		Yes																	

**Table 2. STM32F437xx and STM32F439xx features and peripheral counts (continued)**

Peripherals	STM32F437 Vx	STM32F439 Vx	STM32F437Zx	STM32F437AI	STM32F439AI	STM32F439Zx	STM32F437Ix	STM32F439Ix	STM32F439Bx	STM32F439Nx
Communication interfaces	SPI / I <sup>2</sup> S	6/2 (full duplex) <sup>(2)</sup>								
	I <sup>2</sup> C	3								
	USART/ UART	4/4								
	USB OTG FS	Yes								
	USB OTG HS	Yes								
	CAN	2								
	SAI	1								
	SDIO	Yes								
Camera interface	Yes									
LCD-TFT	No	Yes	No	Yes	Yes	No	Yes	Yes	Yes	Yes
Chrom-ART Accelerator™ (DMA2D)	Yes									
Cryptography	Yes									
GPIOs	82		114			140		168		168
12-bit ADC Number of channels	3									
	16		24							
12-bit DAC Number of channels	Yes 2									
Maximum CPU frequency	180 MHz									
Operating voltage	1.7/ 1.8 to 3.6 V <sup>(3)</sup>									
Operating temperatures	Ambient temperatures: -40 to +85 °C / -40 to +105 °C									
	Junction temperature: -40 to + 125 °C									
Package	LQFP100		WLCSP143 LQFP144	UFBGA169		WLCSP143 LQFP144	UFBGA176 LQFP176		LQFP208	TFBGA216

1. For the LQFP100 package, only FMC Bank1 or Bank2 are available. Bank1 can only support a multiplexed NOR/PSRAM memory using the NE1 Chip Select. Bank2 can only support a 16- or 8-bit NAND Flash memory using the NCE2 Chip Select. The interrupt line cannot be used since Port G is not available in this package.
2. The SPI2 and SPI3 interfaces give the flexibility to work in an exclusive way in either the SPI mode or the I2S audio mode.
3. V<sub>DD</sub>/V<sub>DDA</sub> minimum value of 1.7 V is obtained with the use of an external power supply supervisor (refer to [Section 3.17.2: Internal reset OFF](#)).

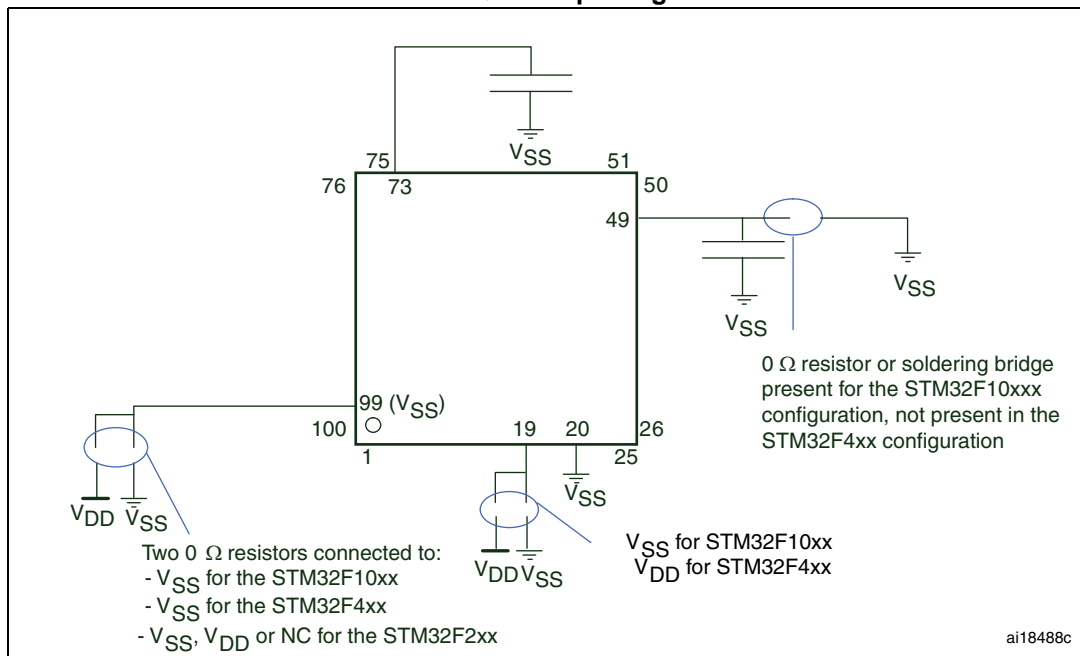
## 2.1 Full compatibility throughout the family

The STM32F437xx and STM32F439xx devices are part of the STM32F4 family. They are fully pin-to-pin, software and feature compatible with the STM32F2xx devices, allowing the user to try different memory densities, peripherals, and performances (FPU, higher frequency) for a greater degree of freedom during the development cycle.

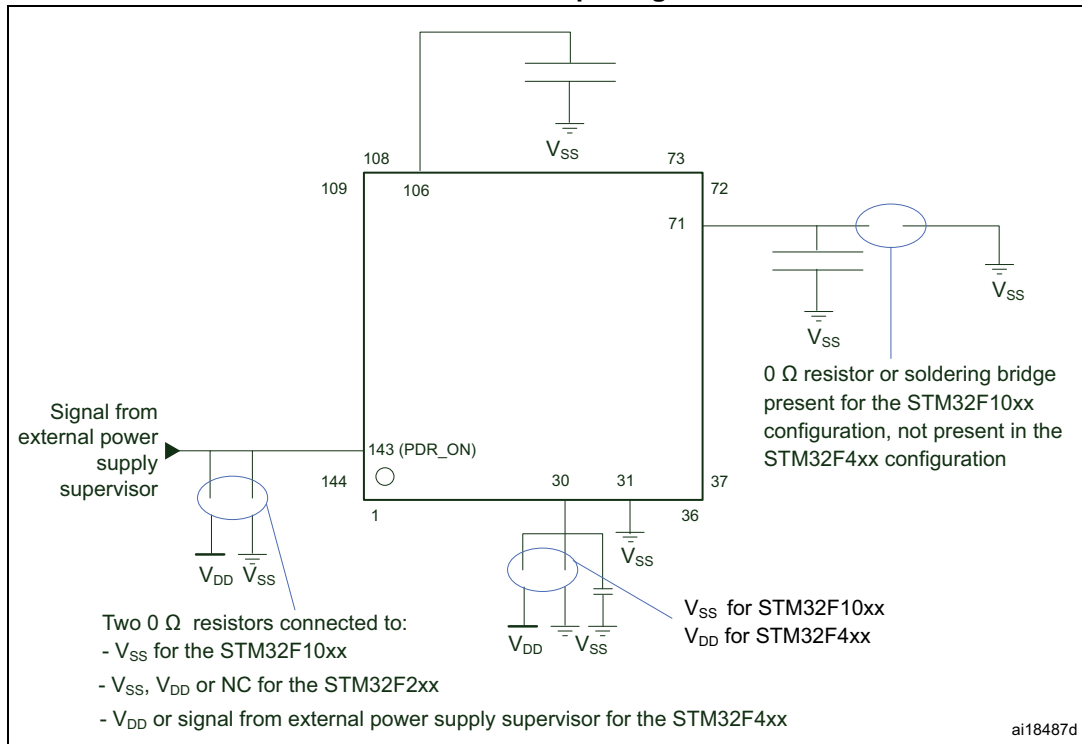
The STM32F437xx and STM32F439xx devices maintain a close compatibility with the whole STM32F10xx family. All functional pins are pin-to-pin compatible. The STM32F437xx and STM32F439xx, however, are not drop-in replacements for the STM32F10xx devices: the two families do not have the same power scheme, and so their power pins are different. Nonetheless, transition from the STM32F10xx to the STM32F43x family remains simple as only a few pins are impacted.

[Figure 1](#), [Figure 2](#), and [Figure 3](#), give compatible board designs between the STM32F4xx, STM32F2xx, and STM32F10xx families.

**Figure 1. Compatible board design STM32F10xx/STM32F2xx/STM32F4xx for LQFP100 package**



**Figure 2. Compatible board design between STM32F10xx/STM32F2xx/STM32F4xx for LQFP144 package**



**Figure 3. Compatible board design between STM32F2xx and STM32F4xx for LQFP176 and UFBGA176 packages**

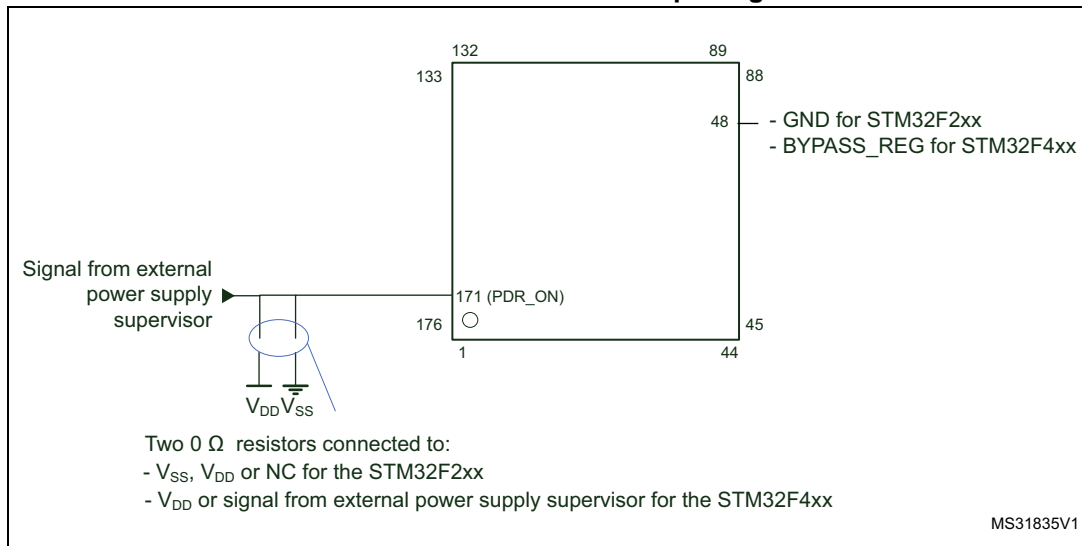
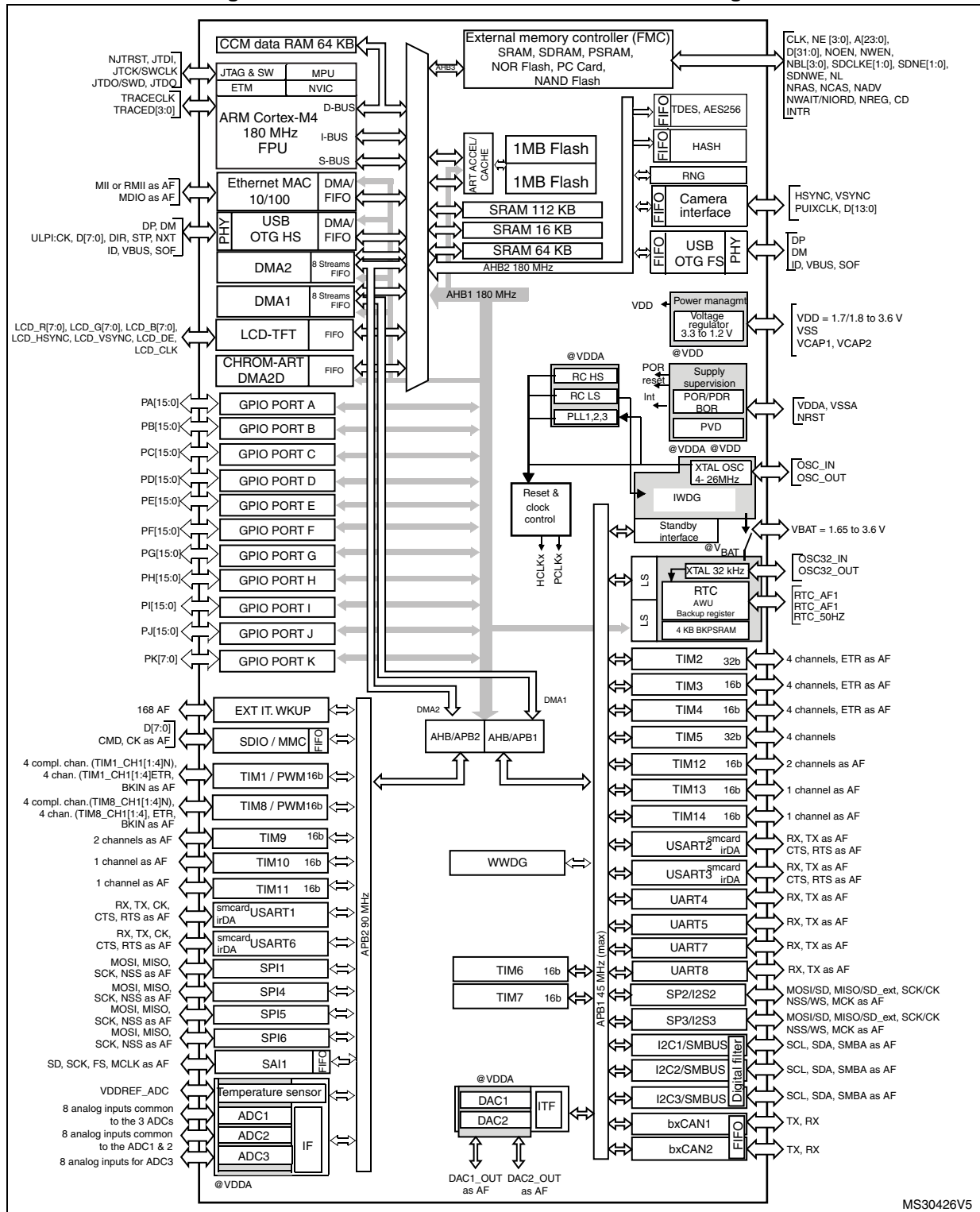


Figure 4. STM32F437xx and STM32F439xx block diagram



1. The timers connected to APB2 are clocked from TIMxCLK up to 180 MHz, while the timers connected to APB1 are clocked from TIMxCLK either up to 90 MHz or 180 MHz depending on TIMPRE bit configuration in the RCC\_DCKCFGR register.
2. The LCD-TFT is available only on STM32F439xx devices.

## 3 Functional overview

### 3.1 ARM<sup>®</sup> Cortex<sup>™</sup>-M4 with FPU and embedded Flash and SRAM

The ARM Cortex-M4 with FPU processor is the latest generation of ARM processors for embedded systems. It was developed to provide a low-cost platform that meets the needs of MCU implementation, with a reduced pin count and low-power consumption, while delivering outstanding computational performance and an advanced response to interrupts.

The ARM Cortex-M4 with FPU core is a 32-bit RISC processor that features exceptional code-efficiency, delivering the high-performance expected from an ARM core in the memory size usually associated with 8- and 16-bit devices.

The processor supports a set of DSP instructions which allow efficient signal processing and complex algorithm execution.

Its single precision FPU (floating point unit) speeds up software development by using metalanguage development tools, while avoiding saturation.

The STM32F43x family is compatible with all ARM tools and software.

*Figure 4: STM32F437xx and STM32F439xx block diagram* shows the general block diagram of the STM32F43x family.

*Note:* Cortex-M4 with FPU core is binary compatible with the Cortex-M3 core.

### 3.2 Adaptive real-time memory accelerator (ART Accelerator<sup>™</sup>)

The ART Accelerator<sup>™</sup> is a memory accelerator which is optimized for STM32 industry-standard ARM<sup>®</sup> Cortex<sup>™</sup>-M4 with FPU processors. It balances the inherent performance advantage of the ARM Cortex-M4 with FPU over Flash memory technologies, which normally requires the processor to wait for the Flash memory at higher frequencies.

To release the processor full 225 DMIPS performance at this frequency, the accelerator implements an instruction prefetch queue and branch cache, which increases program execution speed from the 128-bit Flash memory. Based on CoreMark benchmark, the performance achieved thanks to the ART Accelerator is equivalent to 0 wait state program execution from Flash memory at a CPU frequency up to 180 MHz.

### 3.3 Memory protection unit

The memory protection unit (MPU) is used to manage the CPU accesses to memory to prevent one task to accidentally corrupt the memory or resources used by any other active task. This memory area is organized into up to 8 protected areas that can in turn be divided up into 8 subareas. The protection area sizes are between 32 bytes and the whole 4 gigabytes of addressable memory.

The MPU is especially helpful for applications where some critical or certified code has to be protected against the misbehavior of other tasks. It is usually managed by an RTOS (real-time operating system). If a program accesses a memory location that is prohibited by the MPU, the RTOS can detect it and take action. In an RTOS environment, the kernel can dynamically update the MPU area setting, based on the process to be executed.

The MPU is optional and can be bypassed for applications that do not need it.

### 3.4 Embedded Flash memory

The devices embed a Flash memory of up to 2 Mbytes available for storing programs and data.

### 3.5 CRC (cyclic redundancy check) calculation unit

The CRC (cyclic redundancy check) calculation unit is used to get a CRC code from a 32-bit data word and a fixed generator polynomial.

Among other applications, CRC-based techniques are used to verify data transmission or storage integrity. In the scope of the EN/IEC 60335-1 standard, they offer a means of verifying the Flash memory integrity. The CRC calculation unit helps compute a software signature during runtime, to be compared with a reference signature generated at link-time and stored at a given memory location.

### 3.6 Embedded SRAM

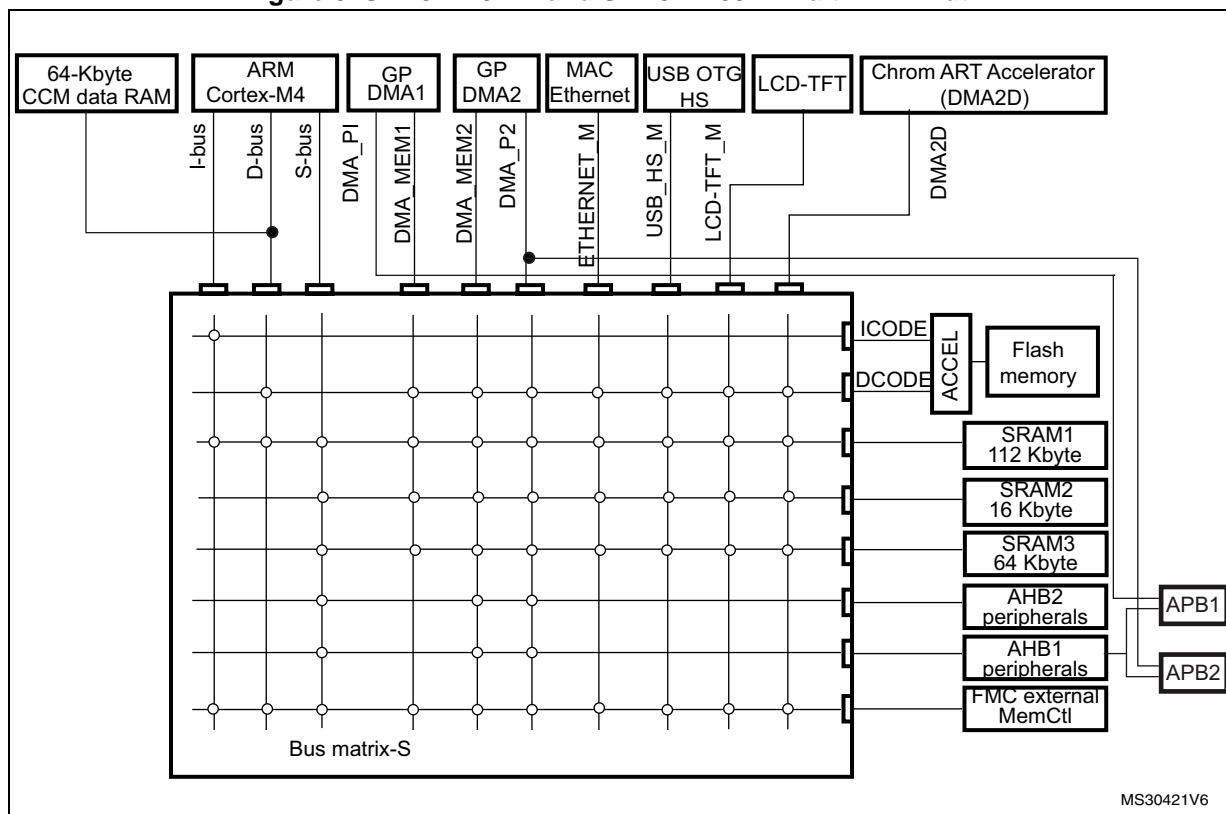
All devices embed:

- Up to 256 Kbytes of system SRAM including 64 Kbytes of CCM (core coupled memory) data RAM  
RAM memory is accessed (read/write) at CPU clock speed with 0 wait states.
- 4 Kbytes of backup SRAM  
This area is accessible only from the CPU. Its content is protected against possible unwanted write accesses, and is retained in Standby or VBAT mode.

### 3.7 Multi-AHB bus matrix

The 32-bit multi-AHB bus matrix interconnects all the masters (CPU, DMAs, Ethernet, USB HS, the LCD-TFT, and the DMA2D) and the slaves (Flash memory, RAM, FMC, AHB and APB peripherals) and ensures a seamless and efficient operation even when several high-speed peripherals work simultaneously.

Figure 5. STM32F437xx and STM32F439xx Multi-AHB matrix



### 3.8 DMA controller (DMA)

The devices feature two general-purpose dual-port DMAs (DMA1 and DMA2) with 8 streams each. They are able to manage memory-to-memory, peripheral-to-memory and memory-to-peripheral transfers. They feature dedicated FIFOs for APB/AHB peripherals, support burst transfer and are designed to provide the maximum peripheral bandwidth (AHB/APB).

The two DMA controllers support circular buffer management, so that no specific code is needed when the controller reaches the end of the buffer. The two DMA controllers also have a double buffering feature, which automates the use and switching of two memory buffers without requiring any special code.

Each stream is connected to dedicated hardware DMA requests, with support for software trigger on each stream. Configuration is made by software and transfer sizes between source and destination are independent.

The DMA can be used with the main peripherals:

- SPI and I<sup>2</sup>S
- I<sup>2</sup>C
- USART
- General-purpose, basic and advanced-control timers TIMx
- DAC
- SDIO
- Cryptographic acceleration
- Camera interface (DCMI)
- ADC
- SAI1.

### 3.9 Flexible memory controller (FMC)

All devices embed an FMC. It has four Chip Select outputs supporting the following modes: PCCard/Compact Flash, SDRAM/LPSDR SDRAM, SRAM, PSRAM, NOR Flash and NAND Flash.

Functionality overview:

- 8-, 16-, 32-bit data bus width
- Read FIFO for SDRAM controller
- Write FIFO
- Maximum FMC\_CLK/FMC\_SDCLK frequency for synchronous accesses is 90 MHz.

#### LCD parallel interface

The FMC can be configured to interface seamlessly with most graphic LCD controllers. It supports the Intel 8080 and Motorola 6800 modes, and is flexible enough to adapt to specific LCD interfaces. This LCD parallel interface capability makes it easy to build cost-effective graphic applications using LCD modules with embedded controllers or high performance solutions using external controllers with dedicated acceleration.

### 3.10 LCD-TFT controller (available only on STM32F439xx)

The LCD-TFT display controller provides a 24-bit parallel digital RGB (Red, Green, Blue) and delivers all signals to interface directly to a broad range of LCD and TFT panels up to SVGA (800x600) resolution with the following features:

- 2 displays layers with dedicated FIFO (64x32-bit)
- Color Look-Up table (CLUT) up to 256 colors (256x24-bit) per layer
- Up to 8 Input color formats selectable per layer
- Flexible blending between two layers using alpha value (per pixel or constant)
- Flexible programmable parameters for each layer
- Color keying (transparency color)
- Up to 4 programmable interrupt events.

### 3.11 Chrom-ART Accelerator™ (DMA2D)

The Chrom-Art Accelerator™ (DMA2D) is a graphic accelerator which offers advanced bit blitting, row data copy and pixel format conversion. It supports the following functions:

- Rectangle filling with a fixed color
- Rectangle copy
- Rectangle copy with pixel format conversion
- Rectangle composition with blending and pixel format conversion.

Various image format coding are supported, from indirect 4bpp color mode up to 32bpp direct color. It embeds dedicated memory to store color lookup tables.

An interrupt can be generated when an operation is complete or at a programmed watermark.

All the operations are fully automatized and are running independently from the CPU or the DMAs.

### 3.12 Nested vectored interrupt controller (NVIC)

The devices embed a nested vectored interrupt controller able to manage 16 priority levels, and handle up to 91 maskable interrupt channels plus the 16 interrupt lines of the Cortex™-M4 with FPU core.

- Closely coupled NVIC gives low-latency interrupt processing
- Interrupt entry vector table address passed directly to the core
- Allows early processing of interrupts
- Processing of late arriving, higher-priority interrupts
- Support tail chaining
- Processor state automatically saved
- Interrupt entry restored on interrupt exit with no instruction overhead

This hardware block provides flexible interrupt management features with minimum interrupt latency.

### 3.13 External interrupt/event controller (EXTI)

The external interrupt/event controller consists of 23 edge-detector lines used to generate interrupt/event requests. Each line can be independently configured to select the trigger event (rising edge, falling edge, both) and can be masked independently. A pending register maintains the status of the interrupt requests. The EXTI can detect an external line with a pulse width shorter than the Internal APB2 clock period. Up to 168 GPIOs can be connected to the 16 external interrupt lines.

### 3.14 Clocks and startup

On reset the 16 MHz internal RC oscillator is selected as the default CPU clock. The 16 MHz internal RC oscillator is factory-trimmed to offer 1% accuracy over the full temperature range. The application can then select as system clock either the RC oscillator or an external 4-26 MHz clock source. This clock can be monitored for failure. If a failure is

detected, the system automatically switches back to the internal RC oscillator and a software interrupt is generated (if enabled). This clock source is input to a PLL thus allowing to increase the frequency up to 180 MHz. Similarly, full interrupt management of the PLL clock entry is available when necessary (for example if an indirectly used external oscillator fails).

Several prescalers allow the configuration of the two AHB buses, the high-speed APB (APB2) and the low-speed APB (APB1) domains. The maximum frequency of the two AHB buses is 180 MHz while the maximum frequency of the high-speed APB domains is 90 MHz. The maximum allowed frequency of the low-speed APB domain is 45 MHz.

The devices embed a dedicated PLL (PLL12S) and PLLSAI which allows to achieve audio class performance. In this case, the I<sup>2</sup>S master clock can generate all standard sampling frequencies from 8 kHz to 192 kHz.

### 3.15 Boot modes

At startup, boot pins are used to select one out of three boot options:

- Boot from user Flash
- Boot from system memory
- Boot from embedded SRAM

The boot loader is located in system memory. It is used to reprogram the Flash memory through a serial interface. Refer to application note AN2606 for details.

### 3.16 Power supply schemes

- $V_{DD} = 1.7/1.8$  to 3.6 V: external power supply for I/Os and the internal regulator (when enabled), provided externally through  $V_{DD}$  pins.
- $V_{SSA}, V_{DDA} = 1.7/1.8$  to 3.6 V: external analog power supplies for ADC, DAC, Reset blocks, RCs and PLL.  $V_{DDA}$  and  $V_{SSA}$  must be connected to  $V_{DD}$  and  $V_{SS}$ , respectively.
- $V_{BAT} = 1.65$  to 3.6 V: power supply for RTC, external clock 32 kHz oscillator and backup registers (through power switch) when  $V_{DD}$  is not present.

*Note:*  $V_{DD}/V_{DDA}$  minimum value of 1.7 V is obtained with the use of an external power supply supervisor (refer to [Section 3.17.2: Internal reset OFF](#)). Refer to [Table 3: Voltage regulator configuration mode versus device operating mode](#) to identify the packages supporting this option.

### 3.17 Power supply supervisor

#### 3.17.1 Internal reset ON

On packages embedding the PDR\_ON pin, the power supply supervisor is enabled by holding PDR\_ON high. On the other package, the power supply supervisor is always enabled.

The device has an integrated power-on reset (POR)/ power-down reset (PDR) circuitry coupled with a Brownout reset (BOR) circuitry. At power-on, POR/PDR is always active and ensures proper operation starting from 1.8 V. After the 1.8 V POR threshold level is

reached, the option byte loading process starts, either to confirm or modify default BOR thresholds, or to disable BOR permanently. Three BOR thresholds are available through option bytes. The device remains in reset mode when  $V_{DD}$  is below a specified threshold,  $V_{POR/PDR}$  or  $V_{BOR}$ , without the need for an external reset circuit.

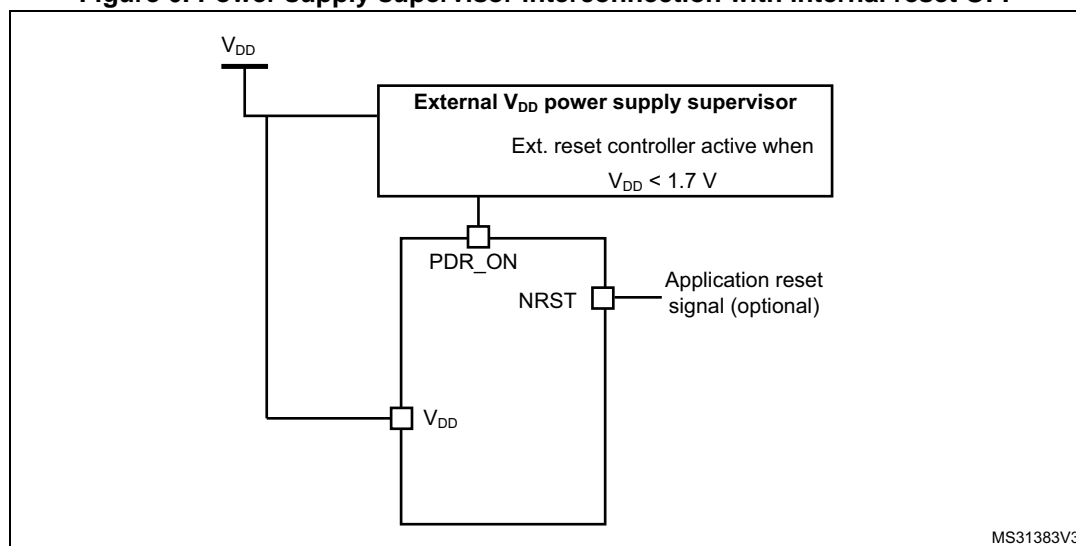
The device also features an embedded programmable voltage detector (PVD) that monitors the  $V_{DD}/V_{DDA}$  power supply and compares it to the  $V_{PVD}$  threshold. An interrupt can be generated when  $V_{DD}/V_{DDA}$  drops below the  $V_{PVD}$  threshold and/or when  $V_{DD}/V_{DDA}$  is higher than the  $V_{PVD}$  threshold. The interrupt service routine can then generate a warning message and/or put the MCU into a safe state. The PVD is enabled by software.

### 3.17.2 Internal reset OFF

This feature is available only on packages featuring the PDR\_ON pin. The internal power-on reset (POR) / power-down reset (PDR) circuitry is disabled through the PDR\_ON pin.

An external power supply supervisor should monitor  $V_{DD}$  and should maintain the device in reset mode as long as  $V_{DD}$  is below a specified threshold. PDR\_ON should be connected to this external power supply supervisor. Refer to [Figure 6: Power supply supervisor interconnection with internal reset OFF](#).

**Figure 6. Power supply supervisor interconnection with internal reset OFF**



The  $V_{DD}$  specified threshold, below which the device must be maintained under reset, is 1.7 V (see [Figure 7](#)).

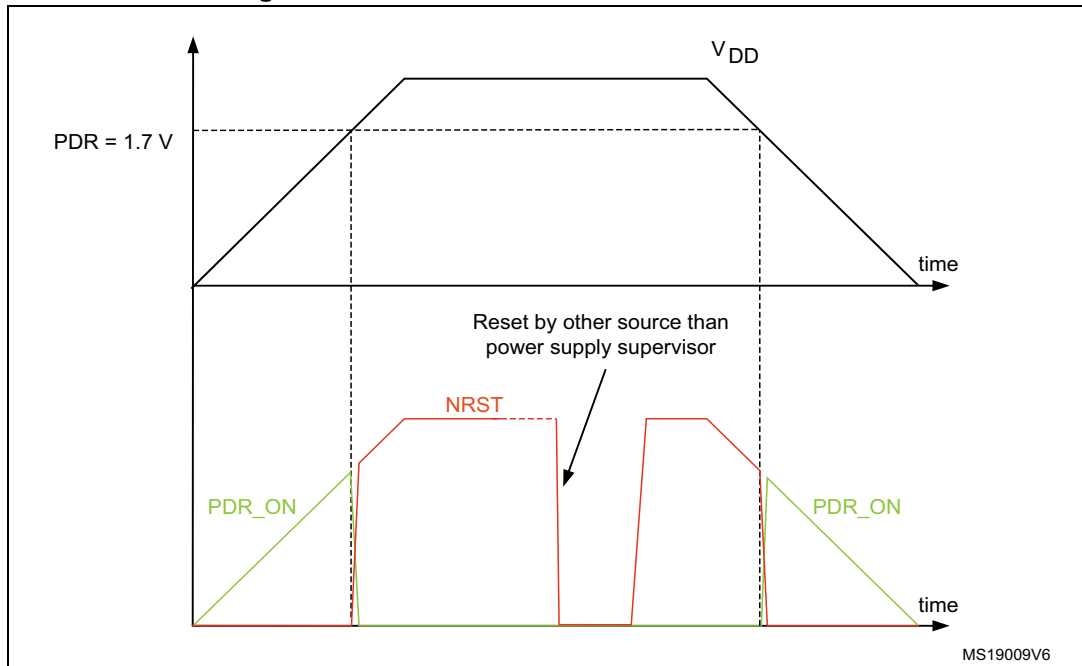
A comprehensive set of power-saving mode allows to design low-power applications.

When the internal reset is OFF, the following integrated features are no more supported:

- The integrated power-on reset (POR) / power-down reset (PDR) circuitry is disabled
- The brownout reset (BOR) circuitry must be disabled
- The embedded programmable voltage detector (PVD) is disabled
- $V_{BAT}$  functionality is no more available and  $V_{BAT}$  pin should be connected to  $V_{DD}$ .

All packages, except for the LQFP100, allow to disable the internal reset through the PDR\_ON signal.

Figure 7. PDR\_ON control with internal reset OFF



### 3.18 Voltage regulator

The regulator has four operating modes:

- Regulator ON
  - Main regulator mode (MR)
  - Low power regulator (LPR)
  - Power-down
- Regulator OFF

#### 3.18.1 Regulator ON

On packages embedding the BYPASS\_REG pin, the regulator is enabled by holding BYPASS\_REG low. On all other packages, the regulator is always enabled.

There are three power modes configured by software when the regulator is ON:

- MR mode used in Run/sleep modes or in Stop modes
  - In Run/Sleep mode
    - The MR mode is used either in the normal mode (default mode) or the over-drive mode (enabled by software). Different voltages scaling are provided to reach the best compromise between maximum frequency and dynamic power consumption.

The over-drive mode allows operating at a higher frequency than the normal mode for a given voltage scaling.

- In Stop modes

The MR can be configured in two ways during stop mode:

MR operates in normal mode (default mode of MR in stop mode)

MR operates in under-drive mode (reduced leakage mode).

- LPR is used in the Stop modes:

The LP regulator mode is configured by software when entering Stop mode.

Like the MR mode, the LPR can be configured in two ways during stop mode:

- LPR operates in normal mode (default mode when LPR is ON)
- LPR operates in under-drive mode (reduced leakage mode).

- Power-down is used in Standby mode.

The Power-down mode is activated only when entering in Standby mode. The regulator output is in high impedance and the kernel circuitry is powered down, inducing zero consumption. The contents of the registers and SRAM are lost.

Refer to [Table 3](#) for a summary of voltage regulator modes versus device operating modes.

Two external ceramic capacitors should be connected on V<sub>CAP\_1</sub> and V<sub>CAP\_2</sub> pin. Refer to [Figure 22: Power supply scheme](#) and [Table 19: VCAP1/VCAP2 operating conditions](#).

All packages have the regulator ON feature.

**Table 3. Voltage regulator configuration mode versus device operating mode<sup>(1)</sup>**

Voltage regulator configuration	Run mode	Sleep mode	Stop mode	Standby mode
Normal mode	MR	MR	MR or LPR	-
Over-drive mode <sup>(2)</sup>	MR	MR	-	-
Under-drive mode	-	-	MR or LPR	-
Power-down mode	-	-	-	Yes

1. '-' means that the corresponding configuration is not available.

2. The over-drive mode is not available when V<sub>DD</sub> = 1.7/1.8 to 2.1 V.

### 3.18.2 Regulator OFF

This feature is available only on packages featuring the BYPASS\_REG pin. The regulator is disabled by holding BYPASS\_REG high. The regulator OFF mode allows to supply externally a V<sub>12</sub> voltage source through V<sub>CAP\_1</sub> and V<sub>CAP\_2</sub> pins.

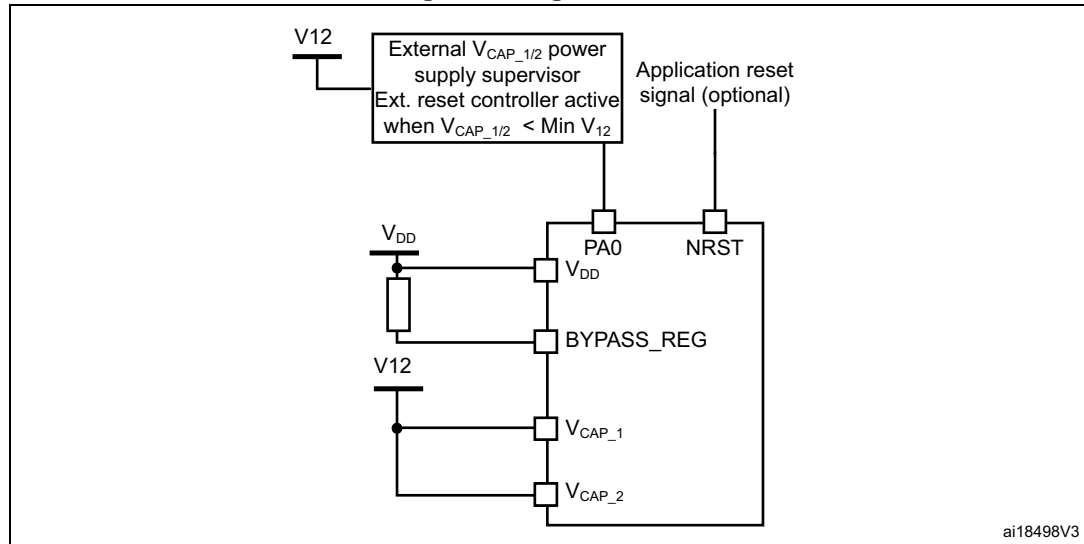
Since the internal voltage scaling is not managed internally, the external voltage value must be aligned with the targeted maximum frequency. Refer to [Table 17: General operating conditions](#). The two 2.2 µF ceramic capacitors should be replaced by two 100 nF decoupling capacitors. Refer to [Figure 22: Power supply scheme](#).

When the regulator is OFF, there is no more internal monitoring on V<sub>12</sub>. An external power supply supervisor should be used to monitor the V<sub>12</sub> of the logic power domain. PA0 pin should be used for this purpose, and act as power-on reset on V<sub>12</sub> power domain.

In regulator OFF mode, the following features are no more supported:

- PA0 cannot be used as a GPIO pin since it allows to reset a part of the  $V_{12}$  logic power domain which is not reset by the NRST pin.
- As long as PA0 is kept low, the debug mode cannot be used under power-on reset. As a consequence, PA0 and NRST pins must be managed separately if the debug connection under reset or pre-reset is required.
- The over-drive and under-drive modes are not available.

Figure 8. Regulator OFF

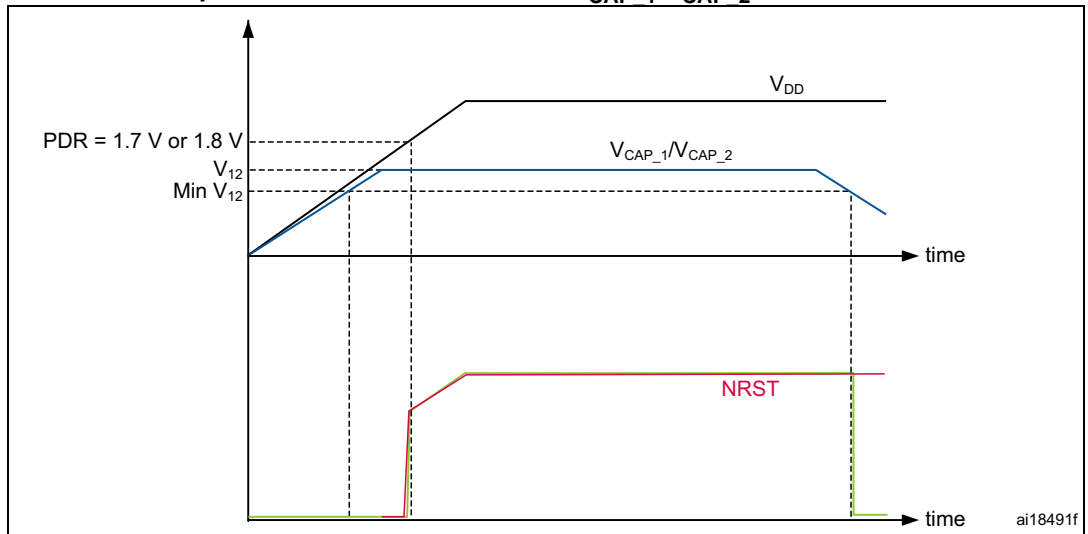


The following conditions must be respected:

- $V_{DD}$  should always be higher than  $V_{CAP\_1}$  and  $V_{CAP\_2}$  to avoid current injection between power domains.
- If the time for  $V_{CAP\_1}$  and  $V_{CAP\_2}$  to reach  $V_{12}$  minimum value is faster than the time for  $V_{DD}$  to reach 1.8 V, then PA0 should be kept low to cover both conditions: until  $V_{CAP\_1}$  and  $V_{CAP\_2}$  reach  $V_{12}$  minimum value and until  $V_{DD}$  reaches 1.8 V (see [Figure 9](#)).
- Otherwise, if the time for  $V_{CAP\_1}$  and  $V_{CAP\_2}$  to reach  $V_{12}$  minimum value is slower than the time for  $V_{DD}$  to reach 1.8 V, then PA0 could be asserted low externally (see [Figure 10](#)).
- If  $V_{CAP\_1}$  and  $V_{CAP\_2}$  go below  $V_{12}$  minimum value and  $V_{DD}$  is higher than 1.8 V, then a reset must be asserted on PA0 pin.

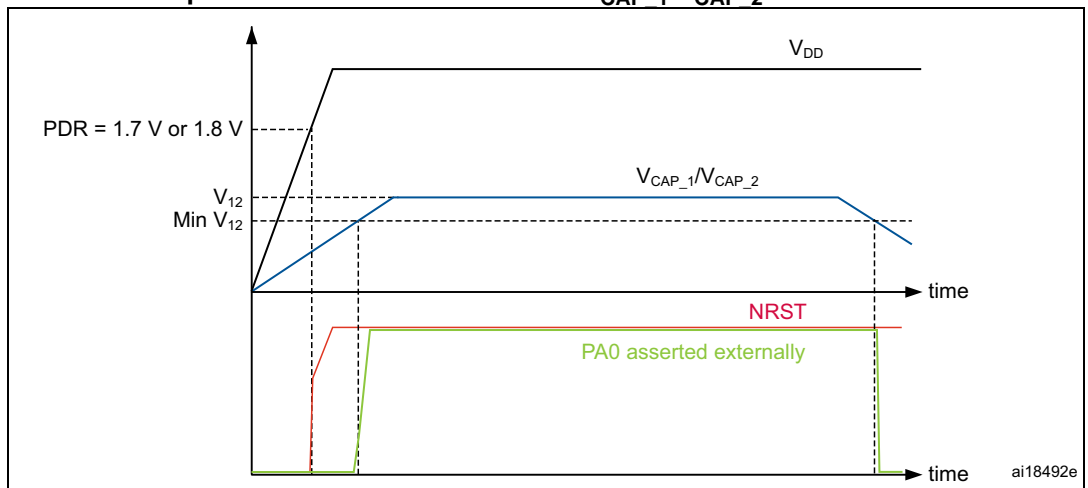
Note: *The minimum value of  $V_{12}$  depends on the maximum frequency targeted in the application (see [Table 17: General operating conditions](#)).*

**Figure 9. Startup in regulator OFF: slow  $V_{DD}$  slope - power-down reset risen after  $V_{CAP\_1}/V_{CAP\_2}$  stabilization**



1. This figure is valid whatever the internal reset mode (ON or OFF).

**Figure 10. Startup in regulator OFF mode: fast  $V_{DD}$  slope - power-down reset risen before  $V_{CAP\_1}/V_{CAP\_2}$  stabilization**



1. This figure is valid whatever the internal reset mode (ON or OFF).

### 3.18.3 Regulator ON/OFF and internal reset ON/OFF availability

Table 4. Regulator ON/OFF and internal reset ON/OFF availability

Package	Regulator ON	Regulator OFF	Internal reset ON	Internal reset OFF
LQFP100	Yes	No	Yes	No
LQFP144			Yes PDR_ON set to $V_{DD}$	Yes PDR_ON connected to an external power supply supervisor
WLCSP143, LQFP176, UFBGA169, UFBGA176, LQFP208, TFBGA216	Yes BYPASS_REG set to $V_{SS}$	Yes BYPASS_REG set to $V_{DD}$		

### 3.19 Real-time clock (RTC), backup SRAM and backup registers

The backup domain includes:

- The real-time clock (RTC)
- 4 Kbytes of backup SRAM
- 20 backup registers

The real-time clock (RTC) is an independent BCD timer/counter. Dedicated registers contain the second, minute, hour (in 12/24 hour), week day, date, month, year, in BCD (binary-coded decimal) format. Correction for 28, 29 (leap year), 30, and 31 day of the month are performed automatically. The RTC provides a programmable alarm and programmable periodic interrupts with wakeup from Stop and Standby modes. The sub-seconds value is also available in binary format.

It is clocked by a 32.768 kHz external crystal, resonator or oscillator, the internal low-power RC oscillator or the high-speed external clock divided by 128. The internal low-speed RC has a typical frequency of 32 kHz. The RTC can be calibrated using an external 512 Hz output to compensate for any natural quartz deviation.

Two alarm registers are used to generate an alarm at a specific time and calendar fields can be independently masked for alarm comparison. To generate a periodic interrupt, a 16-bit programmable binary auto-reload downcounter with programmable resolution is available and allows automatic wakeup and periodic alarms from every 120  $\mu$ s to every 36 hours.

A 20-bit prescaler is used for the time base clock. It is by default configured to generate a time base of 1 second from a clock at 32.768 kHz.

The 4-Kbyte backup SRAM is an EEPROM-like memory area. It can be used to store data which need to be retained in VBAT and standby mode. This memory area is disabled by default to minimize power consumption (see [Section 3.20: Low-power modes](#)). It can be enabled by software.

The backup registers are 32-bit registers used to store 80 bytes of user application data when  $V_{DD}$  power is not present. Backup registers are not reset by a system, a power reset, or when the device wakes up from the Standby mode (see [Section 3.20: Low-power modes](#)).

Additional 32-bit registers contain the programmable alarm subseconds, seconds, minutes, hours, day, and date.

Like backup SRAM, the RTC and backup registers are supplied through a switch that is powered either from the  $V_{DD}$  supply when present or from the  $V_{BAT}$  pin.

## 3.20 Low-power modes

The devices support three low-power modes to achieve the best compromise between low power consumption, short startup time and available wakeup sources:

- **Sleep mode**

In Sleep mode, only the CPU is stopped. All peripherals continue to operate and can wake up the CPU when an interrupt/event occurs.

- **Stop mode**

The Stop mode achieves the lowest power consumption while retaining the contents of SRAM and registers. All clocks in the 1.2 V domain are stopped, the PLL, the HSI RC and the HSE crystal oscillators are disabled.

The voltage regulator can be put either in main regulator mode (MR) or in low-power mode (LPR). Both modes can be configured as follows (see [Table 5: Voltage regulator modes in stop mode](#)):

- Normal mode (default mode when MR or LPR is enabled)
- Under-drive mode.

The device can be woken up from the Stop mode by any of the EXTI line (the EXTI line source can be one of the 16 external lines, the PVD output, the RTC alarm / wakeup / tamper / time stamp events, the USB OTG FS/HS wakeup or the Ethernet wakeup).

**Table 5. Voltage regulator modes in stop mode**

Voltage regulator configuration	Main regulator (MR)	Low-power regulator (LPR)
Normal mode	MR ON	LPR ON
Under-drive mode	MR in under-drive mode	LPR in under-drive mode

- **Standby mode**

The Standby mode is used to achieve the lowest power consumption. The internal voltage regulator is switched off so that the entire 1.2 V domain is powered off. The PLL, the HSI RC and the HSE crystal oscillators are also switched off. After entering Standby mode, the SRAM and register contents are lost except for registers in the backup domain and the backup SRAM when selected.

The device exits the Standby mode when an external reset (NRST pin), an IWDG reset, a rising edge on the WKUP pin, or an RTC alarm / wakeup / tamper /time stamp event occurs.

The standby mode is not supported when the embedded voltage regulator is bypassed and the 1.2 V domain is controlled by an external power.

### 3.21 $V_{BAT}$ operation

The  $V_{BAT}$  pin allows to power the device  $V_{BAT}$  domain from an external battery, an external supercapacitor, or from  $V_{DD}$  when no external battery and an external supercapacitor are present.

$V_{BAT}$  operation is activated when  $V_{DD}$  is not present.

The  $V_{BAT}$  pin supplies the RTC, the backup registers and the backup SRAM.

*Note:* When the microcontroller is supplied from  $V_{BAT}$ , external interrupts and RTC alarm/events do not exit it from  $V_{BAT}$  operation.

*When PDR\_ON pin is not connected to  $V_{DD}$  (Internal Reset OFF), the  $V_{BAT}$  functionality is no more available and  $V_{BAT}$  pin should be connected to  $V_{DD}$ .*

### 3.22 Timers and watchdogs

The devices include two advanced-control timers, eight general-purpose timers, two basic timers and two watchdog timers.

All timer counters can be frozen in debug mode.

[Table 6](#) compares the features of the advanced-control, general-purpose and basic timers.

Table 6. Timer feature comparison

Timer type	Timer	Counter resolution	Counter type	Prescaler factor	DMA request generation	Capture/compare channels	Complementary output	Max interface clock (MHz)	Max timer clock (MHz) <sup>(1)</sup>
Advanced -control	TIM1, TIM8	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	Yes	90	180
General purpose	TIM2, TIM5	32-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	45	90/180
	TIM3, TIM4	16-bit	Up, Down, Up/down	Any integer between 1 and 65536	Yes	4	No	45	90/180
	TIM9	16-bit	Up	Any integer between 1 and 65536	No	2	No	90	180
	TIM10, TIM11	16-bit	Up	Any integer between 1 and 65536	No	1	No	90	180
	TIM12	16-bit	Up	Any integer between 1 and 65536	No	2	No	45	90/180
	TIM13, TIM14	16-bit	Up	Any integer between 1 and 65536	No	1	No	45	90/180
	Basic	TIM6, TIM7	16-bit	Up	Any integer between 1 and 65536	Yes	0	No	45

1. The maximum timer clock is either 90 or 180 MHz depending on TIMPRE bit configuration in the RCC\_DCKCFGR register.

### 3.22.1 Advanced-control timers (TIM1, TIM8)

The advanced-control timers (TIM1, TIM8) can be seen as three-phase PWM generators multiplexed on 6 channels. They have complementary PWM outputs with programmable inserted dead times. They can also be considered as complete general-purpose timers. Their 4 independent channels can be used for:

- Input capture
- Output compare
- PWM generation (edge- or center-aligned modes)
- One-pulse mode output

If configured as standard 16-bit timers, they have the same features as the general-purpose TIMx timers. If configured as 16-bit PWM generators, they have full modulation capability (0-100%).

The advanced-control timer can work together with the TIMx timers via the Timer Link feature for synchronization or event chaining.

TIM1 and TIM8 support independent DMA request generation.

### 3.22.2 General-purpose timers (TIMx)

There are ten synchronizable general-purpose timers embedded in the STM32F43x devices (see [Table 6](#) for differences).

- **TIM2, TIM3, TIM4, TIM5**

The STM32F43x include 4 full-featured general-purpose timers: TIM2, TIM5, TIM3, and TIM4. The TIM2 and TIM5 timers are based on a 32-bit auto-reload up/downcounter and a 16-bit prescaler. The TIM3 and TIM4 timers are based on a 16-bit auto-reload up/downcounter and a 16-bit prescaler. They all feature 4 independent channels for input capture/output compare, PWM or one-pulse mode output. This gives up to 16 input capture/output compare/PWMs on the largest packages.

The TIM2, TIM3, TIM4, TIM5 general-purpose timers can work together, or with the other general-purpose timers and the advanced-control timers TIM1 and TIM8 via the Timer Link feature for synchronization or event chaining.

Any of these general-purpose timers can be used to generate PWM outputs.

TIM2, TIM3, TIM4, TIM5 all have independent DMA request generation. They are capable of handling quadrature (incremental) encoder signals and the digital outputs from 1 to 4 hall-effect sensors.

- **TIM9, TIM10, TIM11, TIM12, TIM13, and TIM14**

These timers are based on a 16-bit auto-reload upcounter and a 16-bit prescaler. TIM10, TIM11, TIM13, and TIM14 feature one independent channel, whereas TIM9 and TIM12 have two independent channels for input capture/output compare, PWM or one-pulse mode output. They can be synchronized with the TIM2, TIM3, TIM4, TIM5 full-featured general-purpose timers. They can also be used as simple time bases.

### 3.22.3 Basic timers TIM6 and TIM7

These timers are mainly used for DAC trigger and waveform generation. They can also be used as a generic 16-bit time base.

TIM6 and TIM7 support independent DMA request generation.

### 3.22.4 Independent watchdog

The independent watchdog is based on a 12-bit downcounter and 8-bit prescaler. It is clocked from an independent 32 kHz internal RC and as it operates independently from the main clock, it can operate in Stop and Standby modes. It can be used either as a watchdog to reset the device when a problem occurs, or as a free-running timer for application timeout management. It is hardware- or software-configurable through the option bytes.

### 3.22.5 Window watchdog

The window watchdog is based on a 7-bit downcounter that can be set as free-running. It can be used as a watchdog to reset the device when a problem occurs. It is clocked from the main clock. It has an early warning interrupt capability and the counter can be frozen in debug mode.

### 3.22.6 SysTick timer

This timer is dedicated to real-time operating systems, but could also be used as a standard downcounter. It features:

- A 24-bit downcounter
- Autoreload capability
- Maskable system interrupt generation when the counter reaches 0
- Programmable clock source.

## 3.23 Inter-integrated circuit interface (I<sup>2</sup>C)

Up to three I<sup>2</sup>C bus interfaces can operate in multimaster and slave modes. They can support the standard (up to 100 KHz) and fast (up to 400 KHz) modes. They support the 7/10-bit addressing mode and the 7-bit dual addressing mode (as slave). A hardware CRC generation/verification is embedded.

They can be served by DMA and they support SMBus 2.0/PMBus.

The devices also include programmable analog and digital noise filters (see [Table 7](#)).

**Table 7. Comparison of I2C analog and digital filters**

	Analog filter	Digital filter
Pulse width of suppressed spikes	≥ 50 ns	Programmable length from 1 to 15 I2C peripheral clocks

## 3.24 Universal synchronous/asynchronous receiver transmitters (USART)

The devices embed four universal synchronous/asynchronous receiver transmitters (USART1, USART2, USART3 and USART6) and four universal asynchronous receiver transmitters (UART4, UART5, UART7, and UART8).

These six interfaces provide asynchronous communication, IrDA SIR ENDEC support, multiprocessor communication mode, single-wire half-duplex communication mode and have LIN Master/Slave capability. The USART1 and USART6 interfaces are able to

communicate at speeds of up to 11.25 Mbit/s. The other available interfaces communicate at up to 5.62 bit/s.

USART1, USART2, USART3 and USART6 also provide hardware management of the CTS and RTS signals, Smart Card mode (ISO 7816 compliant) and SPI-like communication capability. All interfaces can be served by the DMA controller.

**Table 8. USART feature comparison<sup>(1)</sup>**

USART name	Standard features	Modem (RTS/CTS)	LIN	SPI master	irDA	Smartcard (ISO 7816)	Max. baud rate in Mbit/s (oversampling by 16)	Max. baud rate in Mbit/s (oversampling by 8)	APB mapping
USART1	X	X	X	X	X	X	5.62	11.25	APB2 (max. 90 MHz)
USART2	X	X	X	X	X	X	2.81	5.62	APB1 (max. 45 MHz)
USART3	X	X	X	X	X	X	2.81	5.62	APB1 (max. 45 MHz)
UART4	X	-	X	-	X	-	2.81	5.62	APB1 (max. 45 MHz)
UART5	X	-	X	-	X	-	2.81	5.62	APB1 (max. 45 MHz)
USART6	X	X	X	X	X	X	5.62	11.25	APB2 (max. 90 MHz)
UART7	X	-	X	-	X	-	2.81	5.62	APB1 (max. 45 MHz)
UART8	X	-	X	-	X	-	2.81	5.62	APB1 (max. 45 MHz)

1. X = feature supported.

### 3.25 Serial peripheral interface (SPI)

The devices feature up to six SPIs in slave and master modes in full-duplex and simplex communication modes. SPI1, SPI4, SPI5, and SPI6 can communicate at up to 45 Mbits/s, SPI2 and SPI3 can communicate at up to 22.5 Mbit/s. The 3-bit prescaler gives 8 master mode frequencies and the frame is configurable to 8 bits or 16 bits. The hardware CRC generation/verification supports basic SD Card/MMC modes. All SPIs can be served by the DMA controller.

The SPI interface can be configured to operate in TI mode for communications in master mode and slave mode.

### 3.26 Inter-integrated sound (I<sup>2</sup>S)

Two standard I<sup>2</sup>S interfaces (multiplexed with SPI2 and SPI3) are available. They can be operated in master or slave mode, in full duplex and simplex communication modes, and can be configured to operate with a 16-/32-bit resolution as an input or output channel. Audio sampling frequencies from 8 kHz up to 192 kHz are supported. When either or both of the I<sup>2</sup>S interfaces is/are configured in master mode, the master clock can be output to the external DAC/CODEC at 256 times the sampling frequency.

All I2Sx can be served by the DMA controller.

*Note:* For I2S2 full-duplex mode, I2S2\_CK and I2S2\_WS signals can be used only on GPIO Port B and GPIO Port D.

### 3.27 Serial Audio interface (SAI1)

The serial audio interface (SAI1) is based on two independent audio subblocks which can operate as transmitter or receiver with their FIFO. Many audio protocols are supported by each block: I2S standards, LSB or MSB-justified, PCM/DSP, TDM, AC'97 and SPDIF output, supporting audio sampling frequencies from 8 kHz up to 192 kHz. Both subblocks can be configured in master or in slave mode.

In master mode, the master clock can be output to the external DAC/CODEC at 256 times of the sampling frequency.

The two subblocks can be configured in synchronous mode when full-duplex mode is required.

SAI1 can be served by the DMA controller.

### 3.28 Audio PLL (PLLI2S)

The devices feature an additional dedicated PLL for audio I<sup>2</sup>S and SAI applications. It allows to achieve error-free I<sup>2</sup>S sampling clock accuracy without compromising on the CPU performance, while using USB peripherals.

The PLLI2S configuration can be modified to manage an I<sup>2</sup>S/SAI sample rate change without disabling the main PLL (PLL) used for CPU, USB and Ethernet interfaces.

The audio PLL can be programmed with very low error to obtain sampling rates ranging from 8 KHz to 192 KHz.

In addition to the audio PLL, a master clock input pin can be used to synchronize the I<sup>2</sup>S/SAI flow with an external PLL (or Codec output).

### 3.29 Audio and LCD PLL(PLLSAI)

An additional PLL dedicated to audio and LCD-TFT is used for SAI1 peripheral in case the PLLI2S is programmed to achieve another audio sampling frequency (49.152 MHz or 11.2896 MHz) and the audio application requires both sampling frequencies simultaneously.

The PLLSAI is also used to generate the LCD-TFT clock.

### 3.30 Secure digital input/output interface (SDIO)

An SD/SDIO/MMC host interface is available, that supports MultiMediaCard System Specification Version 4.2 in three different databus modes: 1-bit (default), 4-bit and 8-bit.

The interface allows data transfer at up to 48 MHz, and is compliant with the SD Memory Card Specification Version 2.0.

The SDIO Card Specification Version 2.0 is also supported with two different databus modes: 1-bit (default) and 4-bit.

The current version supports only one SD/SDIO/MMC4.2 card at any one time and a stack of MMC4.1 or previous.

In addition to SD/SDIO/MMC, this interface is fully compliant with the CE-ATA digital protocol Rev1.1.

### 3.31 Ethernet MAC interface with dedicated DMA and IEEE 1588 support

The devices provide an IEEE-802.3-2002-compliant media access controller (MAC) for ethernet LAN communications through an industry-standard medium-independent interface (MII) or a reduced medium-independent interface (RMII). The microcontroller requires an external physical interface device (PHY) to connect to the physical LAN bus (twisted-pair, fiber, etc.). The PHY is connected to the device MII port using 17 signals for MII or 9 signals for RMII, and can be clocked using the 25 MHz (MII) from the microcontroller.

The devices include the following features:

- Supports 10 and 100 Mbit/s rates
- Dedicated DMA controller allowing high-speed transfers between the dedicated SRAM and the descriptors (see the STM32F4xx reference manual for details)
- Tagged MAC frame support (VLAN support)
- Half-duplex (CSMA/CD) and full-duplex operation
- MAC control sublayer (control frames) support
- 32-bit CRC generation and removal
- Several address filtering modes for physical and multicast address (multicast and group addresses)
- 32-bit status code for each transmitted or received frame
- Internal FIFOs to buffer transmit and receive frames. The transmit FIFO and the receive FIFO are both 2 Kbytes.
- Supports hardware PTP (precision time protocol) in accordance with IEEE 1588 2008 (PTP V2) with the time stamp comparator connected to the TIM2 input
- Triggers interrupt when system time becomes greater than target time

### 3.32 Controller area network (bxCAN)

The two CANs are compliant with the 2.0A and B (active) specifications with a bitrate up to 1 Mbit/s. They can receive and transmit standard frames with 11-bit identifiers as well as extended frames with 29-bit identifiers. Each CAN has three transmit mailboxes, two receive

FIFOS with 3 stages and 28 shared scalable filter banks (all of them can be used even if one CAN is used). 256 bytes of SRAM are allocated for each CAN.

### 3.33 Universal serial bus on-the-go full-speed (OTG\_FS)

The devices embed an USB OTG full-speed device/host/OTG peripheral with integrated transceivers. The USB OTG FS peripheral is compliant with the USB 2.0 specification and with the OTG 1.0 specification. It has software-configurable endpoint setting and supports suspend/resume. The USB OTG full-speed controller requires a dedicated 48 MHz clock that is generated by a PLL connected to the HSE oscillator. The major features are:

- Combined Rx and Tx FIFO size of 320 × 35 bits with dynamic FIFO sizing
- Supports the session request protocol (SRP) and host negotiation protocol (HNP)
- 4 bidirectional endpoints
- 8 host channels with periodic OUT support
- HNP/SNP/IP inside (no need for any external resistor)
- For OTG/Host modes, a power switch is needed in case bus-powered devices are connected

### 3.34 Universal serial bus on-the-go high-speed (OTG\_HS)

The devices embed a USB OTG high-speed (up to 480 Mb/s) device/host/OTG peripheral. The USB OTG HS supports both full-speed and high-speed operations. It integrates the transceivers for full-speed operation (12 MB/s) and features a UTMI low-pin interface (ULPI) for high-speed operation (480 MB/s). When using the USB OTG HS in HS mode, an external PHY device connected to the ULPI is required.

The USB OTG HS peripheral is compliant with the USB 2.0 specification and with the OTG 1.0 specification. It has software-configurable endpoint setting and supports suspend/resume. The USB OTG full-speed controller requires a dedicated 48 MHz clock that is generated by a PLL connected to the HSE oscillator.

The major features are:

- Combined Rx and Tx FIFO size of 1 Kbit × 35 with dynamic FIFO sizing
- Supports the session request protocol (SRP) and host negotiation protocol (HNP)
- 6 bidirectional endpoints
- 12 host channels with periodic OUT support
- Internal FS OTG PHY support
- External HS or HS OTG operation supporting ULPI in SDR mode. The OTG PHY is connected to the microcontroller ULPI port through 12 signals. It can be clocked using the 60 MHz output.
- Internal USB DMA
- HNP/SNP/IP inside (no need for any external resistor)
- for OTG/Host modes, a power switch is needed in case bus-powered devices are connected

### 3.35 Digital camera interface (DCMI)

The devices embed a camera interface that can connect with camera modules and CMOS sensors through an 8-bit to 14-bit parallel interface, to receive video data. The camera interface can sustain a data transfer rate up to 54 Mbyte/s at 54 MHz. It features:

- Programmable polarity for the input pixel clock and synchronization signals
- Parallel data communication can be 8-, 10-, 12- or 14-bit
- Supports 8-bit progressive video monochrome or raw bayer format, YCbCr 4:2:2 progressive video, RGB 565 progressive video or compressed data (like JPEG)
- Supports continuous mode or snapshot (a single frame) mode
- Capability to automatically crop the image

### 3.36 Cryptographic acceleration

The devices embed a cryptographic accelerator. This cryptographic accelerator provides a set of hardware acceleration for the advanced cryptographic algorithms usually needed to provide confidentiality, authentication, data integrity and non repudiation when exchanging messages with a peer.

- These algorithms consists of:
  - Encryption/Decryption
    - DES/TDES (data encryption standard/triple data encryption standard): ECB (electronic codebook) and CBC (cipher block chaining) chaining algorithms, 64-, 128- or 192-bit key
    - AES (advanced encryption standard): ECB, CBC, GCM, CCM, and CTR (counter mode) chaining algorithms, 128, 192 or 256-bit key
  - Universal hash
    - SHA-1 and SHA-2 (secure hash algorithms)
    - MD5
    - HMAC

The cryptographic accelerator supports DMA request generation.

### 3.37 Random number generator (RNG)

All devices embed an RNG that delivers 32-bit random numbers generated by an integrated analog circuit.

### 3.38 General-purpose input/outputs (GPIOs)

Each of the GPIO pins can be configured by software as output (push-pull or open-drain, with or without pull-up or pull-down), as input (floating, with or without pull-up or pull-down) or as peripheral alternate function. Most of the GPIO pins are shared with digital or analog alternate functions. All GPIOs are high-current-capable and have speed selection to better manage internal noise, power consumption and electromagnetic emission.

The I/O configuration can be locked if needed by following a specific sequence in order to avoid spurious writing to the I/Os registers.

Fast I/O handling allowing maximum I/O toggling up to 90 MHz.

### 3.39 Analog-to-digital converters (ADCs)

Three 12-bit analog-to-digital converters are embedded and each ADC shares up to 16 external channels, performing conversions in the single-shot or scan mode. In scan mode, automatic conversion is performed on a selected group of analog inputs.

Additional logic functions embedded in the ADC interface allow:

- Simultaneous sample and hold
- Interleaved sample and hold

The ADC can be served by the DMA controller. An analog watchdog feature allows very precise monitoring of the converted voltage of one, some or all selected channels. An interrupt is generated when the converted voltage is outside the programmed thresholds.

To synchronize A/D conversion and timers, the ADCs could be triggered by any of TIM1, TIM2, TIM3, TIM4, TIM5, or TIM8 timer.

### 3.40 Temperature sensor

The temperature sensor has to generate a voltage that varies linearly with temperature. The conversion range is between 1.8 V and 3.6 V. The temperature sensor is internally connected to the same input channel as  $V_{BAT}$ , ADC1\_IN18, which is used to convert the sensor output voltage into a digital value. When the temperature sensor and  $V_{BAT}$  conversion are enabled at the same time, only  $V_{BAT}$  conversion is performed.

As the offset of the temperature sensor varies from chip to chip due to process variation, the internal temperature sensor is mainly suitable for applications that detect temperature changes instead of absolute temperatures. If an accurate temperature reading is needed, then an external temperature sensor part should be used.

### 3.41 Digital-to-analog converter (DAC)

The two 12-bit buffered DAC channels can be used to convert two digital signals into two analog voltage signal outputs.

This dual digital Interface supports the following features:

- two DAC converters: one for each output channel
- 8-bit or 10-bit monotonic output
- left or right data alignment in 12-bit mode
- synchronized update capability
- noise-wave generation
- triangular-wave generation
- dual DAC channel independent or simultaneous conversions
- DMA capability for each channel
- external triggers for conversion
- input voltage reference  $V_{REF+}$

Eight DAC trigger inputs are used in the device. The DAC channels are triggered through the timer update outputs that are also connected to different DMA streams.

### 3.42 Serial wire JTAG debug port (SWJ-DP)

The ARM SWJ-DP interface is embedded, and is a combined JTAG and serial wire debug port that enables either a serial wire debug or a JTAG probe to be connected to the target.

Debug is performed using 2 pins only instead of 5 required by the JTAG (JTAG pins could be re-use as GPIO with alternate function): the JTAG TMS and TCK pins are shared with SWDIO and SWCLK, respectively, and a specific sequence on the TMS pin is used to switch between JTAG-DP and SW-DP.

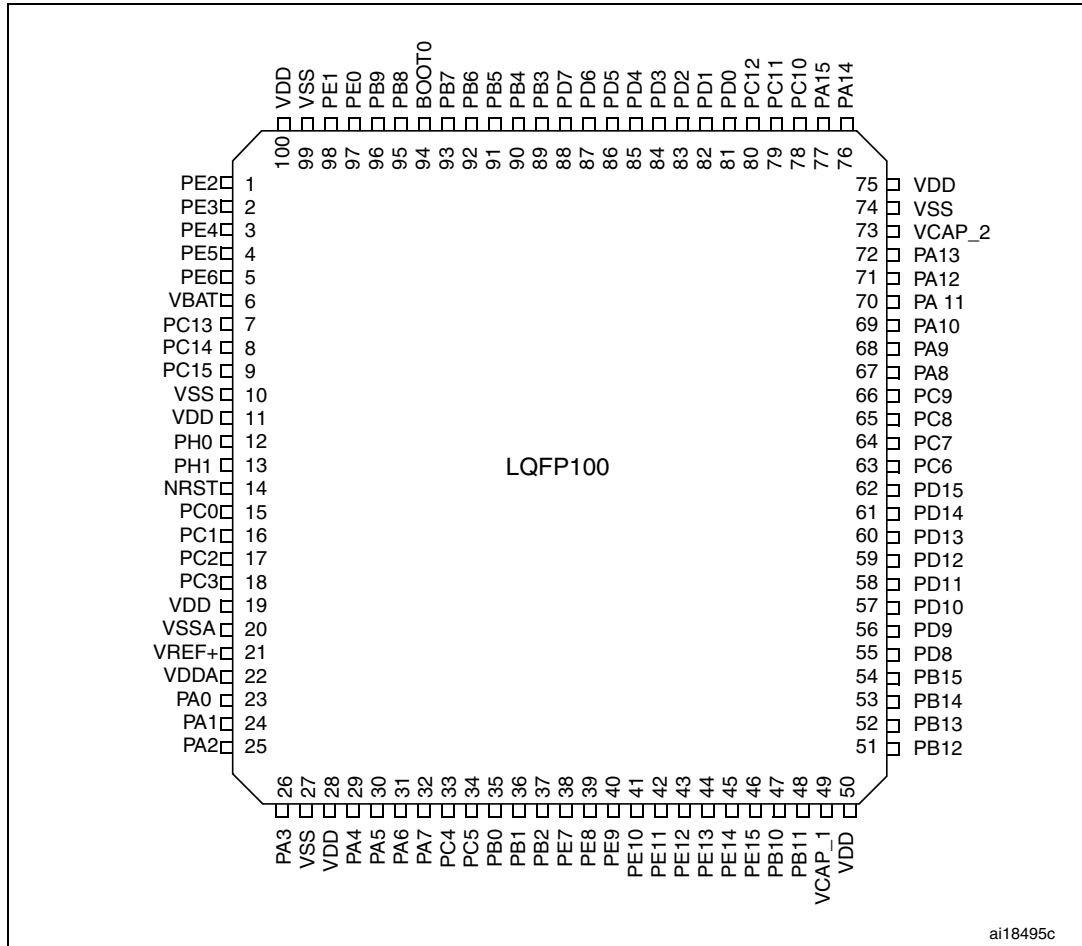
### 3.43 Embedded Trace Macrocell™

The ARM Embedded Trace Macrocell provides a greater visibility of the instruction and data flow inside the CPU core by streaming compressed data at a very high rate from the STM32F43x through a small number of ETM pins to an external hardware trace port analyzer (TPA) device. The TPA is connected to a host computer using USB, Ethernet, or any other high-speed channel. Real-time instruction and data flow activity can be recorded and then formatted for display on the host computer that runs the debugger software. TPA hardware is commercially available from common development tool vendors.

The Embedded Trace Macrocell operates with third party debugger software tools.

# 4 Pinouts and pin description

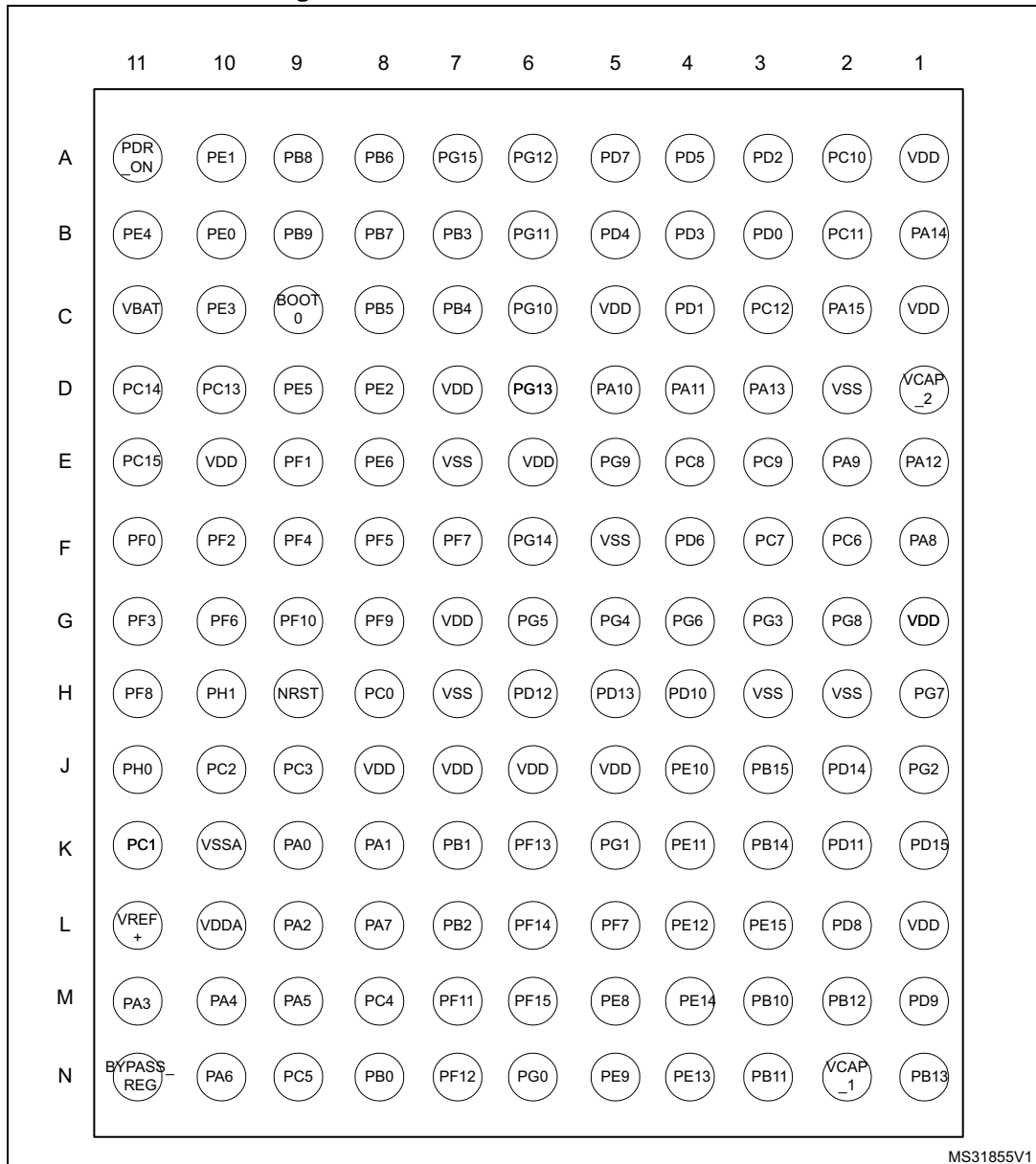
Figure 11. STM32F43x LQFP100 pinout



ai18495c

1. The above figure shows the package top view.

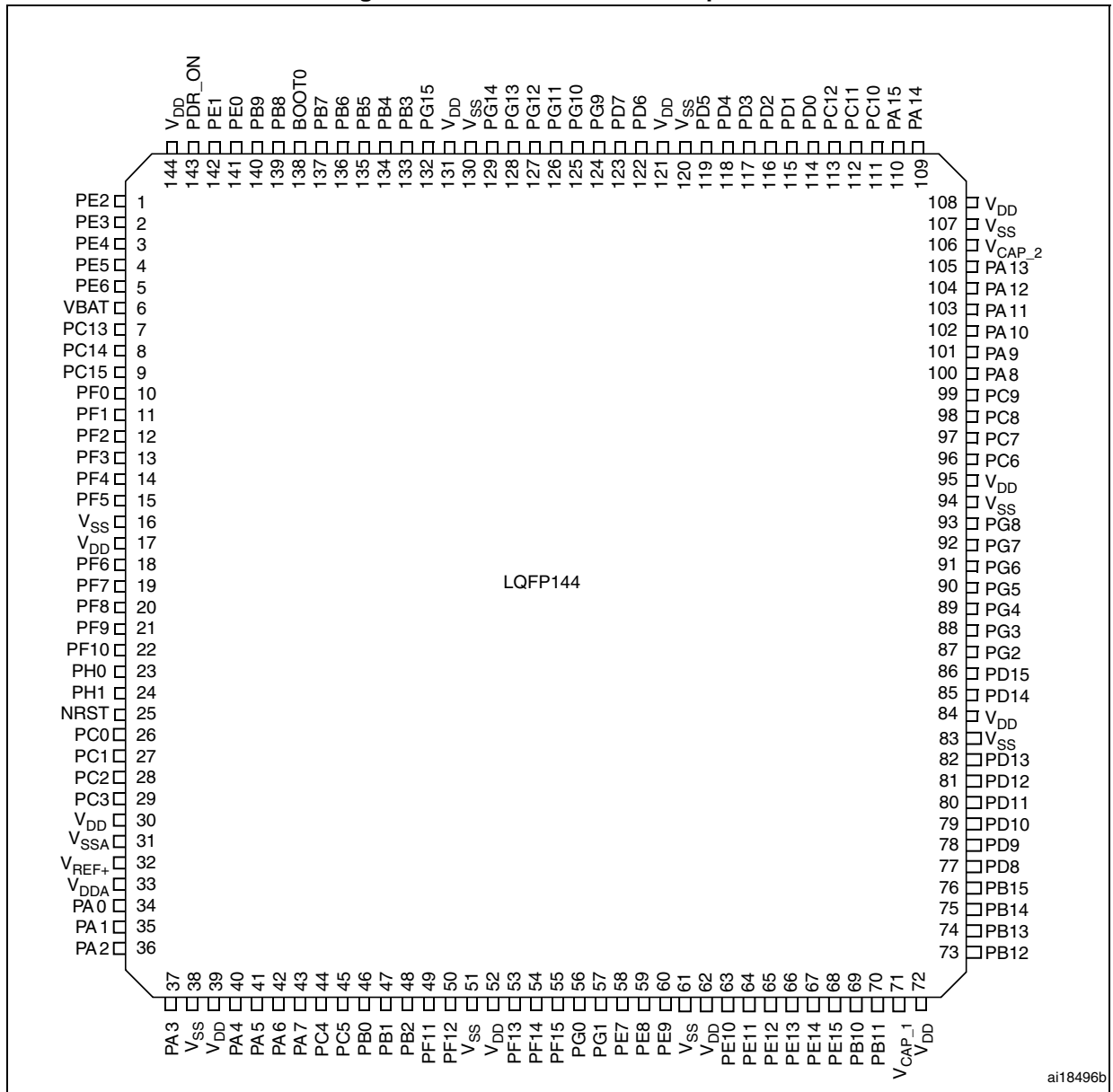
Figure 12. STM32F43x WLCSP143 ballout



MS31855V1

1. The above figure shows the package bump view.

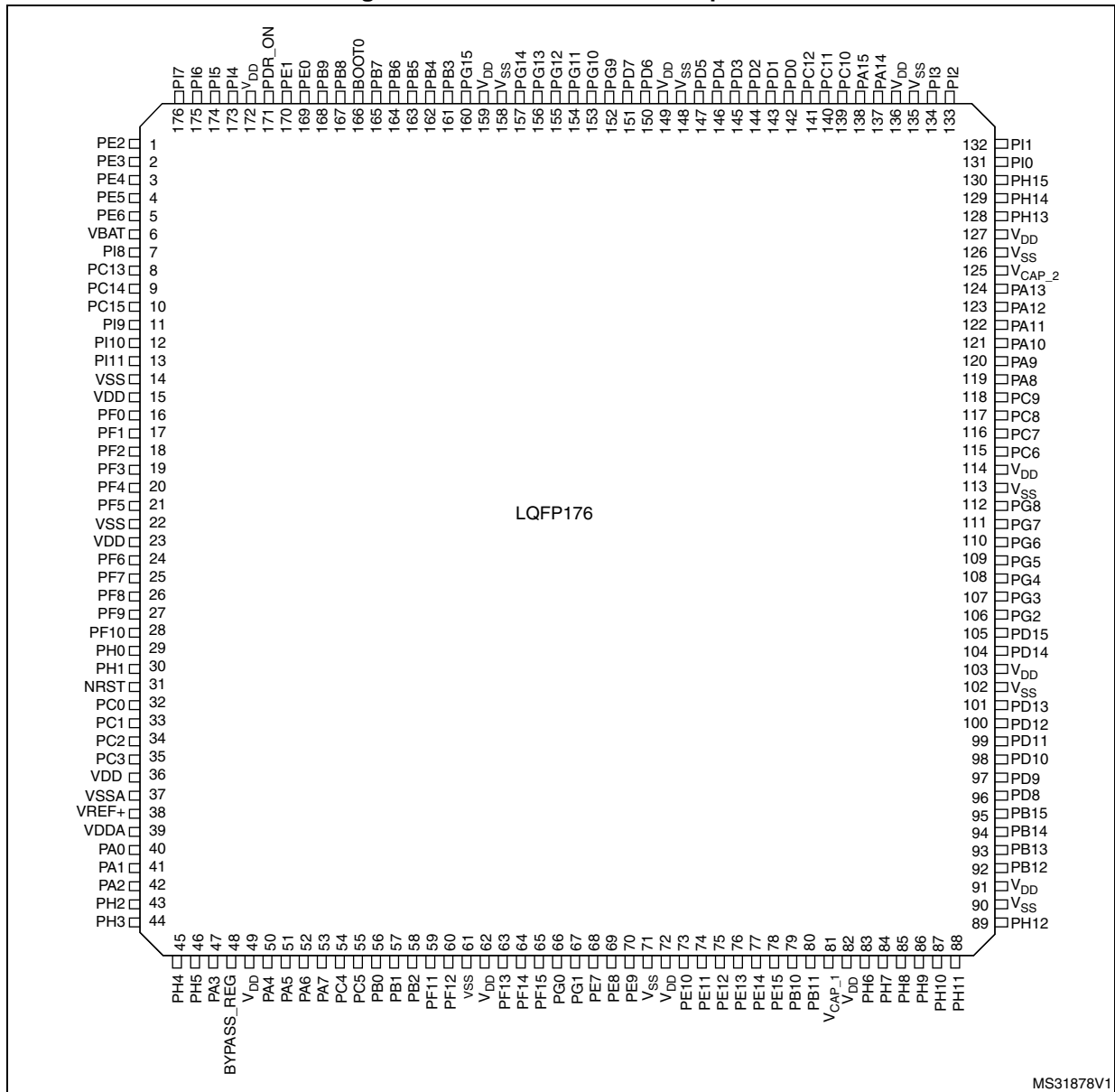
Figure 13. STM32F43x LQFP144 pinout



ai18496b

1. The above figure shows the package top view.

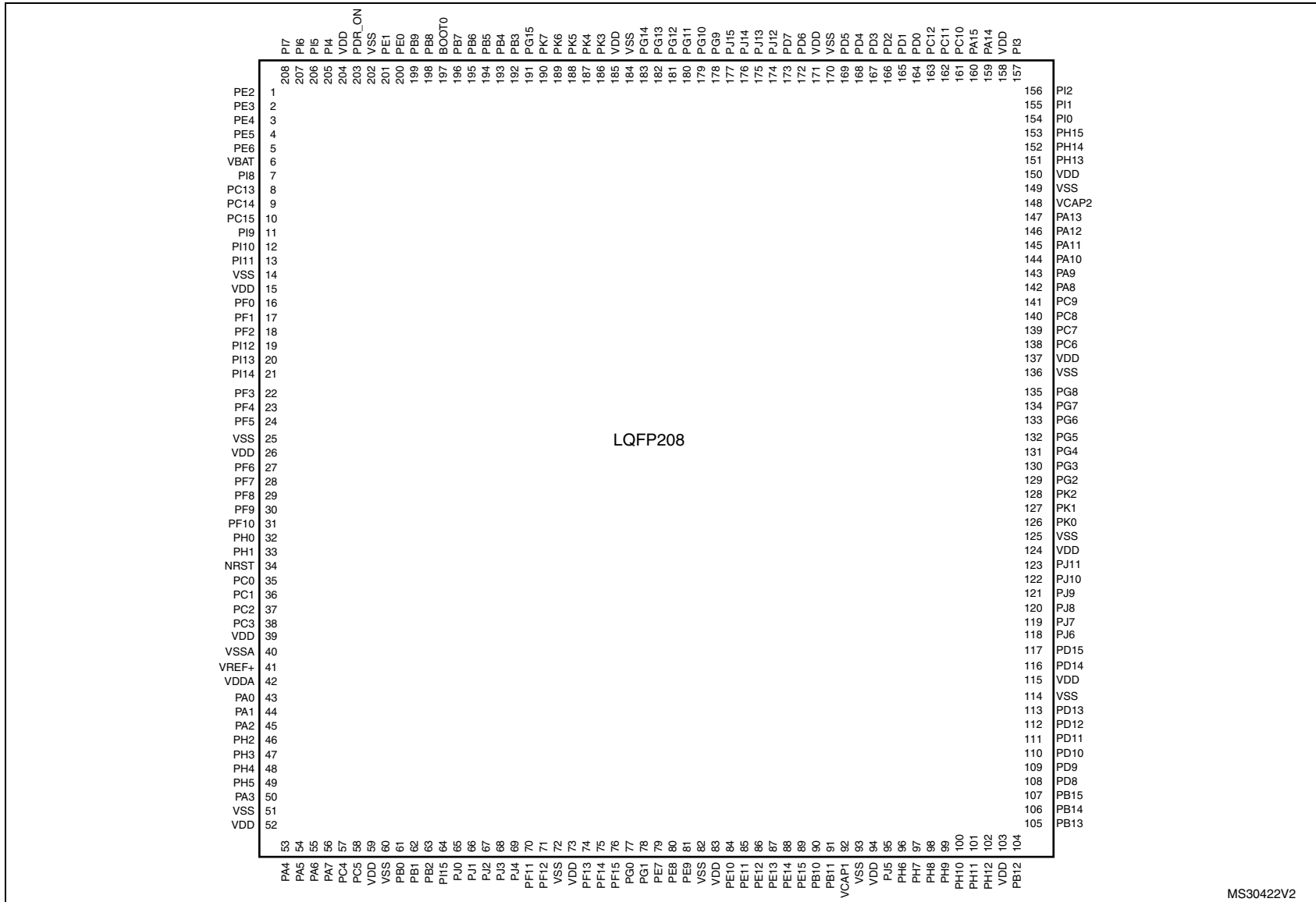
Figure 14. STM32F43x LQFP176 pinout



MS31878V1

1. The above figure shows the package top view.

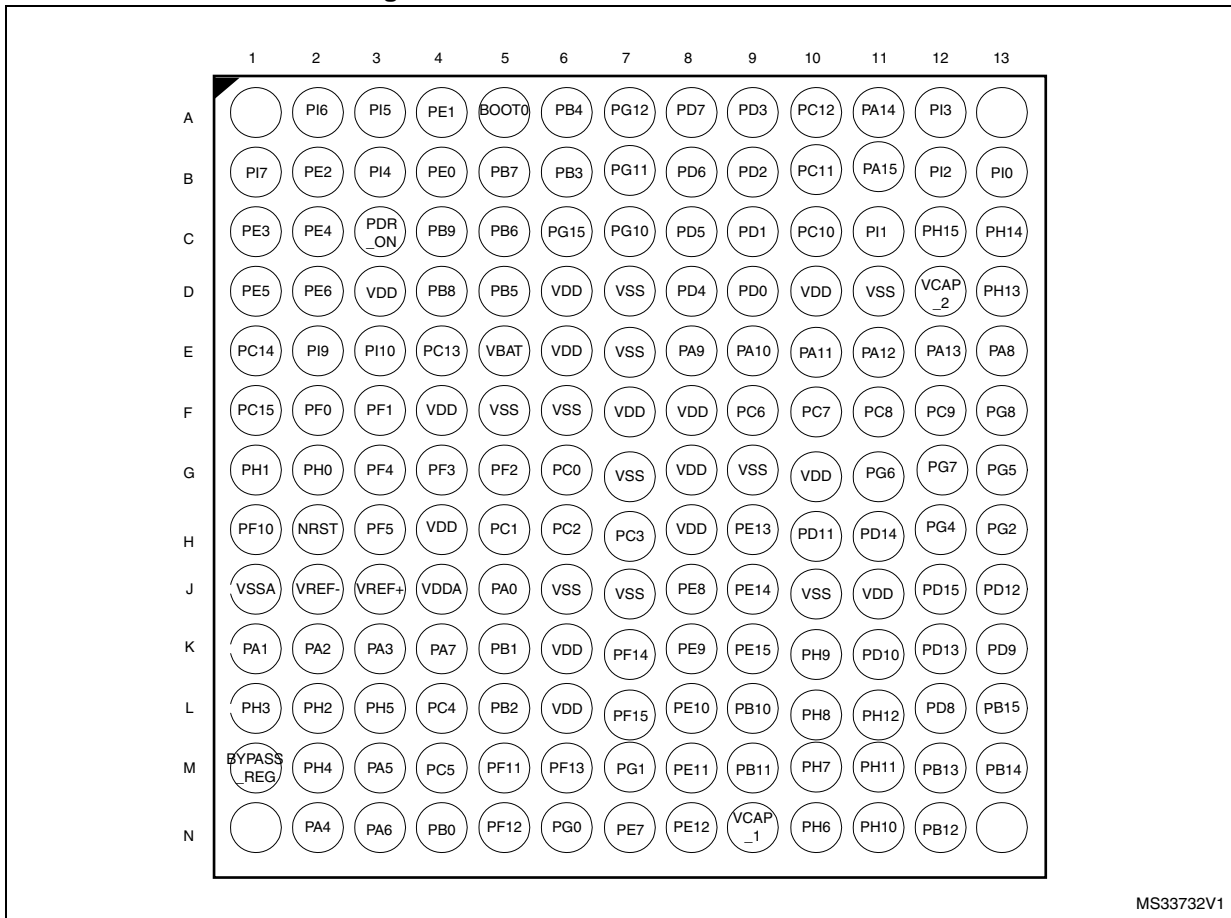
Figure 15. STM32F43x LQFP208 pinout



MS30422V2

1. The above figure shows the package top view.

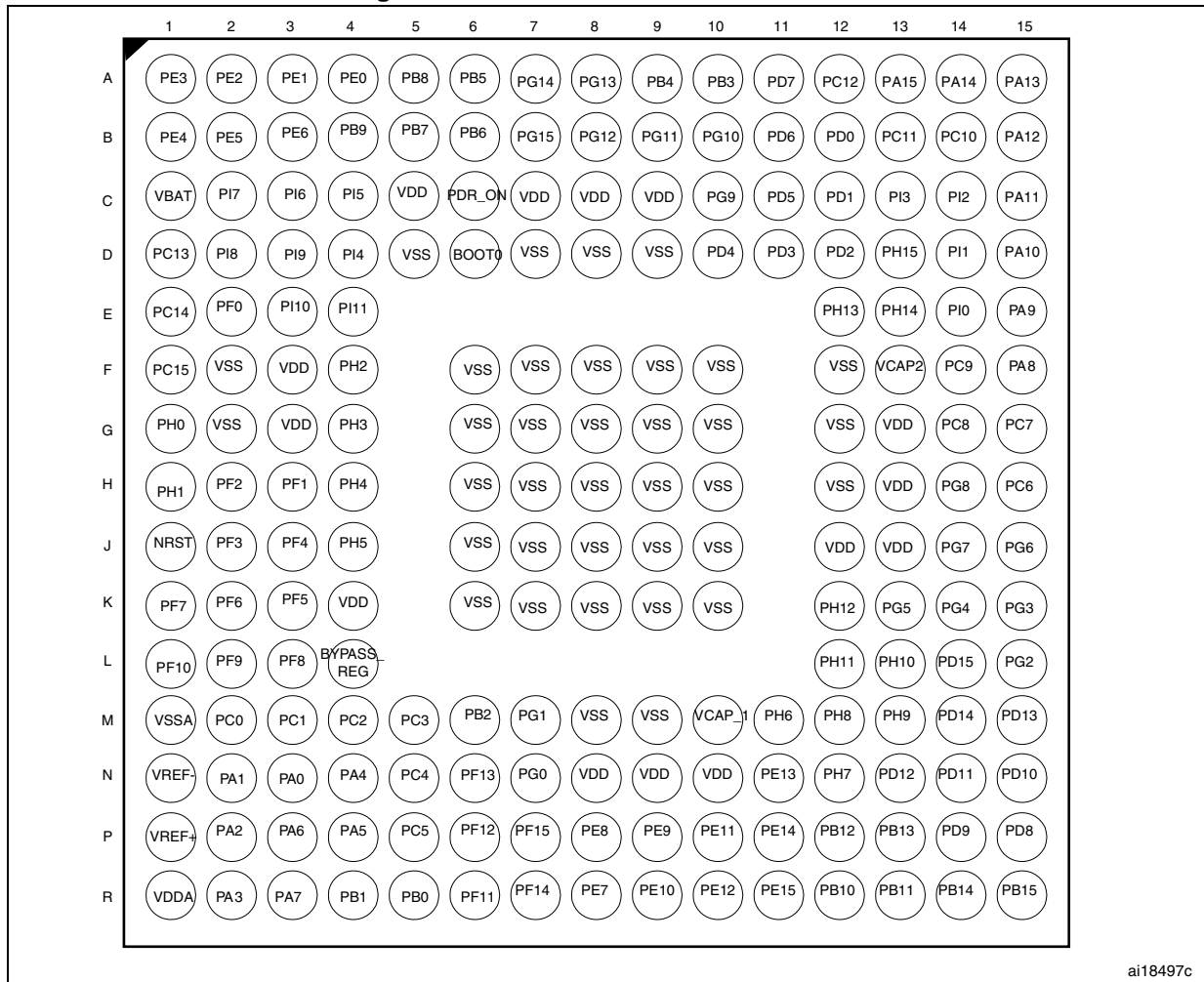
Figure 16. STM32F43x UFBGA169 ballout



MS33732V1

1. The above figure shows the package top view.

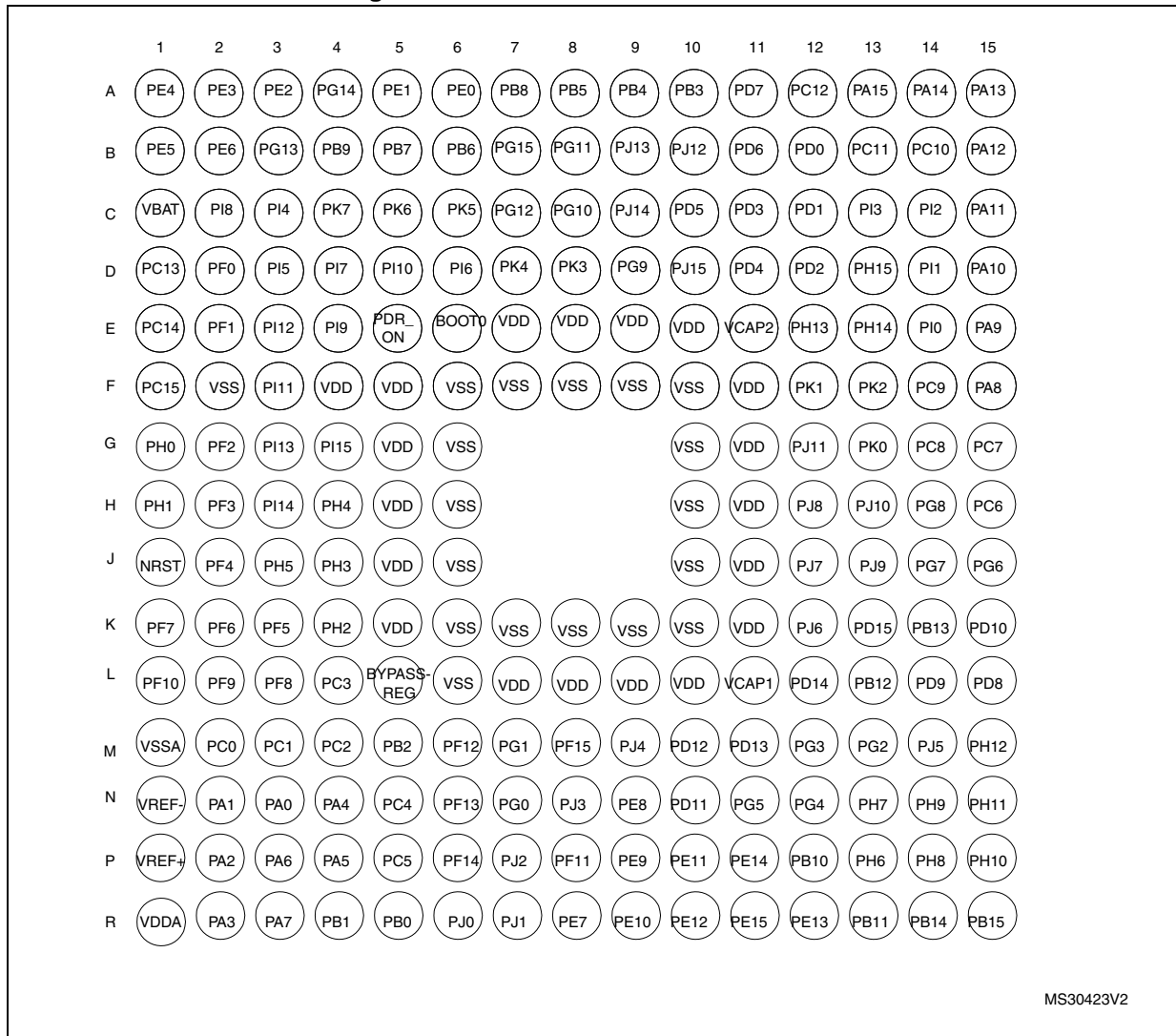
Figure 17. STM32F43x UFBGA176 ballout



ai18497c

1. The above figure shows the package top view.

Figure 18. STM32F43x TFBGA216 ballout



MS30423V2

1. The above figure shows the package top view.

**Table 9. Legend/abbreviations used in the pinout table**

Name	Abbreviation	Definition
Pin name	Unless otherwise specified in brackets below the pin name, the pin function during and after reset is the same as the actual pin name	
Pin type	S	Supply pin
	I	Input only pin
	I/O	Input / output pin
I/O structure	FT	5 V tolerant I/O
	TTa	3.3 V tolerant I/O directly connected to ADC
	B	Dedicated BOOT0 pin
	RST	Bidirectional reset pin with weak pull-up resistor
Notes	Unless otherwise specified by a note, all I/Os are set as floating inputs during and after reset	
Alternate functions	Functions selected through GPIOx_AFR registers	
Additional functions	Functions directly selected/enabled through peripheral registers	

**Table 10. STM32F437xx and STM32F439xx pin and ball definitions**

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
1	1	B2	A2	1	D8	1	A3	PE2	I/O	FT		TRACECLK, SPI4_SCK, SAI1_MCLK_A, ETH_MII_TXD3, FMC_A23, EVENTOUT	
2	2	C1	A1	2	C10	2	A2	PE3	I/O	FT		TRACED0, SAI1_SD_B, FMC_A19, EVENTOUT	
3	3	C2	B1	3	B11	3	A1	PE4	I/O	FT		TRACED1, SPI4_NSS, SAI1_FS_A, FMC_A20, DCMI_D4, LCD_B0, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
4	4	D1	B2	4	D9	4	B1	PE5	I/O	FT		TRACED2, TIM9_CH1, SPI4_MISO, SAI1_SCK_A, FMC_A21, DCMI_D6, LCD_G0, EVENTOUT	
5	5	D2	B3	5	E8	5	B2	PE6	I/O	FT		TRACED3, TIM9_CH2, SPI4_MOSI, SAI1_SD_A, FMC_A22, DCMI_D7, LCD_G1, EVENTOUT	
-	-	-	-	-	-	-	G6	V <sub>SS</sub>	S				
-	-	-	-	-	-	-	F5	V <sub>DD</sub>	S				
6	6	E5	C1	6	C11	6	C1	V <sub>BAT</sub>	S				
-	-	NC (2)	D2	7	-	7	C2	PI8	I/O	FT	(3) (4)	EVENTOUT	TAMP_2
7	7	E4	D1	8	D10	8	D1	PC13	I/O	FT	(3) (4)	EVENTOUT	TAMP_1
8	8	E1	E1	9	D11	9	E1	PC14- OSC32_IN (PC14)	I/O	FT	(3) (4)	EVENTOUT	OSC32_IN (5)
9	9	F1	F1	10	E11	10	F1	PC15- OSC32_OUT (PC15)	I/O	FT	(3) (4)	EVENTOUT	OSC32_OUT <sup>(5)</sup>
-	-	-	-	-	-	-	G5	V <sub>DD</sub>	S				
-	-	E2	D3	11	-	11	E4	PI9	I/O	FT		CAN1_RX, FMC_D30, LCD_VSYNC, EVENTOUT	
-	-	E3	E3	12	-	12	D5	PI10	I/O	FT		ETH_MII_RX_ER, FMC_D31, LCD_HSYNC, EVENTOUT	
-	-	NC (2)	E4	13	-	13	F3	PI11	I/O	FT		OTG_HS_ULPI_DIR, EVENTOUT	
-	-	F6	F2	14	E7	14	F2	V <sub>SS</sub>	S				
-	-	F4	F3	15	E10	15	F4	V <sub>DD</sub>	S				

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
-	10	F2	E2	16	F11	16	D2	PF0	I/O	FT		I2C2_SDA, FMC_A0, EVENTOUT	
-	11	F3	H3	17	E9	17	E2	PF1	I/O	FT		I2C2_SCL, FMC_A1, EVENTOUT	
-	12	G5	H2	18	F10	18	G2	PF2	I/O	FT		I2C2_SMBA, FMC_A2, EVENTOUT	
-	-	-	-	-	-	-	19	E3	I/O	FT		LCD_HSYNC, EVENTOUT	
-	-	-	-	-	-	-	20	G3	I/O	FT		LCD_VSYNC, EVENTOUT	
-	-	-	-	-	-	-	21	H3	I/O	FT		LCD_CLK, EVENTOUT	
-	13	G4	J2	19	G11	22	H2	PF3	I/O	FT	<sup>(5)</sup>	FMC_A3, EVENTOUT	ADC3_IN9
-	14	G3	J3	20	F9	23	J2	PF4	I/O	FT	<sup>(5)</sup>	FMC_A4, EVENTOUT	ADC3_IN14
-	15	H3	K3	21	F8	24	K3	PF5	I/O	FT	<sup>(5)</sup>	FMC_A5, EVENTOUT	ADC3_IN15
10	16	G7	G2	22	H7	25	H6	V <sub>SS</sub>	S				
11	17	G8	G3	23	-	26	H5	V <sub>DD</sub>	S				
-	18	NC (2)	K2	24	G10	27	K2	PF6	I/O	FT	<sup>(5)</sup>	TIM10_CH1, SPI5_NSS, SAI1_SD_B, UART7_Rx, FMC_NIORD, EVENTOUT	ADC3_IN4
-	19	NC (2)	K1	25	F7	28	K1	PF7	I/O	FT	<sup>(5)</sup>	TIM11_CH1, SPI5_SCK, SAI1_MCLK_B, UART7_Tx, FMC_NREG, EVENTOUT	ADC3_IN5
-	20	NC (2)	L3	26	H11	29	L3	PF8	I/O	FT	<sup>(5)</sup>	SPI5_MISO, SAI1_SCK_B, TIM13_CH1, FMC_NIOWR, EVENTOUT	ADC3_IN6

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
-	21	NC (2)	L2	27	G8	30	L2	PF9	I/O	FT	(5)	SPI5_MOSI, SAI1_FS_B, TIM14_CH1, FMC_CD, EVENTOUT	ADC3_IN7
-	22	H1	L1	28	G9	31	L1	PF10	I/O	FT	(5)	FMC_INTR, DCMI_D11, LCD_DE, EVENTOUT	ADC3_IN8
12	23	G2	G1	29	J11	32	G1	PH0-OSC_IN (PH0)	I/O	FT		EVENTOUT	OSC_IN <sup>(5)</sup>
13	24	G1	H1	30	H10	33	H1	PH1- OSC_OUT (PH1)	I/O	FT		EVENTOUT	OSC_OUT (5)
14	25	H2	J1	31	H9	34	J1	NRST	I/O	RS T			
15	26	G6	M2	32	H8	35	M2	PC0	I/O	FT	(5)	OTG_HS_ULPI_STP, FMC_SDNWE, EVENTOUT	ADC123_ IN10
16	27	H5	M3	33	K11	36	M3	PC1	I/O	FT	(5)	ETH_MDC, EVENTOUT	ADC123_ IN11
17	28	H6	M4	34	J10	37	M4	PC2	I/O	FT	(5)	SPI2_MISO, I2S2ext_SD, OTG_HS_ULPI_DIR, ETH_MII_TXD2, FMC_SDNE0, EVENTOUT	ADC123_ IN12
18	29	H7	M5	35	J9	38	L4	PC3	I/O	FT	(5)	SPI2_MOSI/I2S2_SD, OTG_HS_ULPI_NXT, ETH_MII_TX_CLK, FMC_SDCKE0, EVENTOUT	ADC123_ IN13
19	30	-	-	36	G7	39	J5	V <sub>DD</sub>	S				
-	-	-	-	-	-	-	J6	V <sub>SS</sub>	S				
20	31	J1	M1	37	K10	40	M1	V <sub>SSA</sub>	S				
-	-	J2	N1	-	-	-	N1	V <sub>REF-</sub>	S				
21	32	J3	P1	38	L11	41	P1	V <sub>REF+</sub>	S				

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
22	33	J4	R1	39	L10	42	R1	V <sub>DDA</sub>	S				
23	34	J5	N3	40	K9	43	N3	PA0-WKUP (PA0)	I/O	FT	(6)	TIM2_CH1/TIM2_ETR, TIM5_CH1, TIM8_ETR, USART2_CTS, UART4_TX, ETH_MII_CRCS, EVENTOUT	ADC123_ IN0/WKUP (5)
24	35	K1	N2	41	K8	44	N2	PA1	I/O	FT	(5)	TIM2_CH2, TIM5_CH2, USART2_RTS, UART4_RX, ETH_MII_RX_CLK/ETH _RMII_REF_CLK, EVENTOUT	ADC123_ IN1
25	36	K2	P2	42	L9	45	P2	PA2	I/O	FT	(5)	TIM2_CH3, TIM5_CH3, TIM9_CH1, USART2_TX, ETH_MDIO, EVENTOUT	ADC123_ IN2
-	-	L2	F4	43	-	46	K4	PH2	I/O	FT		ETH_MII_CRCS, FMC_SDCKE0, LCD_R0, EVENTOUT	
-	-	L1	G4	44	-	47	J4	PH3	I/O	FT		ETH_MII_COL, FMC_SDNE0, LCD_R1, EVENTOUT	
-	-	M2	H4	45	-	48	H4	PH4	I/O	FT		I2C2_SCL, OTG_HS_ULPI_NXT, EVENTOUT	
-	-	L3	J4	46	-	49	J3	PH5	I/O	FT		I2C2_SDA, SPI5_NSS, FMC_SDNWE, EVENTOUT	
26	37	K3	R2	47	M11	50	R2	PA3	I/O	FT	(5)	TIM2_CH4, TIM5_CH4, TIM9_CH2, USART2_RX, OTG_HS_ULPI_D0, ETH_MII_COL, LCD_B5, EVENTOUT	ADC123_ IN3
27	38	-	-	-	-	51	K6	V <sub>SS</sub>	S				

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
-	-	M1	L4	48	N11	-	L5	BYPASS_ REG	I	FT			
28	39	J11	K4	49	J8	52	K5	V <sub>DD</sub>	S				
29	40	N2	N4	50	M10	53	N4	PA4	I/O	TTa	(5)	SPI1_NSS, SPI3_NSS/I2S3_WS, USART2_CK, OTG_HS_SOF, DCMI_HSYNC, LCD_VSYNC, EVENTOUT	ADC12_ IN4 /DAC_ OUT1
30	41	M3	P4	51	M9	54	P4	PA5	I/O	TTa	(5)	TIM2_CH1/TIM2_ETR, TIM8_CH1N, SPI1_SCK, OTG_HS_ULPI_CK, EVENTOUT	ADC12_ IN5/DAC_ OUT2
31	42	N3	P3	52	N10	55	P3	PA6	I/O	FT	(5)	TIM1_BKIN, TIM3_CH1, TIM8_BKIN, SPI1_MISO, TIM13_CH1, DCMI_PIXCLK, LCD_G2, EVENTOUT	ADC12_ IN6
32	43	K4	R3	53	L8	56	R3	PA7	I/O	FT	(5)	TIM1_CH1N, TIM3_CH2, TIM8_CH1N, SPI1_MOSI, TIM14_CH1, ETH_MII_RX_DV/ETH_ RMII_CRS_DV, EVENTOUT	ADC12_ IN7
33	44	L4	N5	54	M8	57	N5	PC4	I/O	FT	(5)	ETH_MII_RXD0/ETH_ RMII_RXD0, EVENTOUT	ADC12_ IN14
34	45	M4	P5	55	N9	58	P5	PC5	I/O	FT	(5)	ETH_MII_RXD1/ETH_ RMII_RXD1, EVENTOUT	ADC12_ IN15
-	-	-	-	-	J7	59	L7	V <sub>DD</sub>	S				
-	-	-	-	-	-	60	L6	V <sub>SS</sub>	S				

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
35	46	N4	R5	56	N8	61	R5	PB0	I/O	FT	(5)	TIM1_CH2N, TIM3_CH3, TIM8_CH2N, LCD_R3, OTG_HS_ULPI_D1, ETH_MII_RXD2, EVENTOUT	ADC12_ IN8
36	47	K5	R4	57	K7	62	R4	PB1	I/O	FT	(5)	TIM1_CH3N, TIM3_CH4, TIM8_CH3N, LCD_R6, OTG_HS_ULPI_D2, ETH_MII_RXD3, EVENTOUT	ADC12_ IN9
37	48	L5	M6	58	L7	63	M5	PB2-BOOT1 (PB2)	I/O	FT		EVENTOUT	
-	-	-	-	-	-	64	G4	PI15	I/O	FT		LCD_R0, EVENTOUT	
-	-	-	-	-	-	65	R6	PJ0	I/O	FT		LCD_R1, EVENTOUT	
-	-	-	-	-	-	66	R7	PJ1	I/O	FT		LCD_R2, EVENTOUT	
-	-	-	-	-	-	67	P7	PJ2	I/O	FT		LCD_R3, EVENTOUT	
-	-	-	-	-	-	68	N8	PJ3	I/O	FT		LCD_R4, EVENTOUT	
-	-	-	-	-	-	69	M9	PJ4	I/O	FT		LCD_R5, EVENTOUT	
-	49	M5	R6	59	M7	70	P8	PF11	I/O	FT		SPI5_MOSI, FMC_SDNRAS, DCMI_D12, EVENTOUT	
-	50	N5	P6	60	N7	71	M6	PF12	I/O	FT		FMC_A6, EVENTOUT	
-	51	G9	M8	61	-	72	K7	V <sub>SS</sub>	S				
-	52	D10	N8	62	-	73	L8	V <sub>DD</sub>	S				
-	53	M6	N6	63	K6	74	N6	PF13	I/O	FT		FMC_A7, EVENTOUT	
-	54	K7	R7	64	L6	75	P6	PF14	I/O	FT		FMC_A8, EVENTOUT	
-	55	L7	P7	65	M6	76	M8	PF15	I/O	FT		FMC_A9, EVENTOUT	
-	56	N6	N7	66	N6	77	N7	PG0	I/O	FT		FMC_A10, EVENTOUT	
-	57	M7	M7	67	K5	78	M7	PG1	I/O	FT		FMC_A11, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
38	58	N7	R8	68	L5	79	R8	PE7	I/O	FT		TIM1_ETR, UART7_Rx, FMC_D4, EVENTOUT	
39	59	J8	P8	69	M5	80	N9	PE8	I/O	FT		TIM1_CH1N, UART7_Tx, FMC_D5, EVENTOUT	
40	60	K8	P9	70	N5	81	P9	PE9	I/O	FT		TIM1_CH1, FMC_D6, EVENTOUT	
-	61	J6	M9	71	H3	82	K8	V <sub>SS</sub>	S				
-	62	G10	N9	72	J5	83	L9	V <sub>DD</sub>	S				
41	63	L8	R9	73	J4	84	R9	PE10	I/O	FT		TIM1_CH2N, FMC_D7, EVENTOUT	
42	64	M8	P10	74	K4	85	P10	PE11	I/O	FT		TIM1_CH2, SPI4_NSS, FMC_D8, LCD_G3, EVENTOUT	
43	65	N8	R10	75	L4	86	R10	PE12	I/O	FT		TIM1_CH3N, SPI4_SCK, FMC_D9, LCD_B4, EVENTOUT	
44	66	H9	N11	76	N4	87	R12	PE13	I/O	FT		TIM1_CH3, SPI4_MISO, FMC_D10, LCD_DE, EVENTOUT	
45	67	J9	P11	77	M4	88	P11	PE14	I/O	FT		TIM1_CH4, SPI4_MOSI, FMC_D11, LCD_CLK, EVENTOUT	
46	68	K9	R11	78	L3	89	R11	PE15	I/O	FT		TIM1_BKIN, FMC_D12, LCD_R7, EVENTOUT	
47	69	L9	R12	79	M3	90	P12	PB10	I/O	FT		TIM2_CH3, I2C2_SCL, SPI2_SCK/I2S2_CK, USART3_TX, OTG_HS_ULPI_D3, ETH_MII_RX_ER, LCD_G4, EVENTOUT	
48	70	M9	R13	80	N3	91	R13	PB11	I/O	FT		TIM2_CH4, I2C2_SDA, USART3_RX, OTG_HS_ULPI_D4, ETH_MII_TX_EN/ETH_ RMII_TX_EN, LCD_G5, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
49	71	N9	M10	81	N2	92	L11	V <sub>CAP_1</sub>	S				
-	-	-	-	-	H2	93	K9	V <sub>SS</sub>	S				
50	72	F8	N10	82	J6	94	L10	V <sub>DD</sub>	S				
-	-	-	-	-	-	95	M14	PJ5	I/O			LCD_R6, EVENTOUT	
-	-	N10	M11	83	-	96	P13	PH6	I/O	FT		I2C2_SMBA, SPI5_SCK, TIM12_CH1, ETH_MII_RXD2, FMC_SDNE1, DCMI_D8, EVENTOUT	
-	-	M10	N12	84	-	97	N13	PH7	I/O	FT		I2C3_SCL, SPI5_MISO, ETH_MII_RXD3, FMC_SDCKE1, DCMI_D9, EVENTOUT	
-	-	L10	M12	85	-	98	P14	PH8	I/O	FT		I2C3_SDA, FMC_D16, DCMI_HSYNC, LCD_R2, EVENTOUT	
-	-	K10	M13	86	-	99	N14	PH9	I/O	FT		I2C3_SMBA, TIM12_CH2, FMC_D17, DCMI_D0, LCD_R3, EVENTOUT	
-	-	N11	L13	87	-	100	P15	PH10	I/O	FT		TIM5_CH1, FMC_D18, DCMI_D1, LCD_R4, EVENTOUT	
-	-	M11	L12	88	-	101	N15	PH11	I/O	FT		TIM5_CH2, FMC_D19, DCMI_D2, LCD_R5, EVENTOUT	
-	-	L11	K12	89	-	102	M15	PH12	I/O	FT		TIM5_CH3, FMC_D20, DCMI_D3, LCD_R6, EVENTOUT	
-	-	E7	H12	90	-	-	K10	V <sub>SS</sub>	S				
-	-	H8	J12	91	-	103	K11	V <sub>DD</sub>	S				

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
51	73	N12	P12	92	M2	104	L13	PB12	I/O	FT		TIM1_BKIN, I2C2_SMBA, SPI2_NSS/I2S2_WS, USART3_CK, CAN2_RX, OTG_HS_ULPI_D5, ETH_MII_TXD0/ETH_R MII_TXD0, OTG_HS_ID, EVENTOUT	
52	74	M12	P13	93	N1	105	K14	PB13	I/O	FT		TIM1_CH1N, SPI2_SCK/I2S2_CK, USART3_CTS, CAN2_TX, OTG_HS_ULPI_D6, ETH_MII_TXD1/ETH_R MII_TXD1, EVENTOUT	OTG_HS_­ VBUS
53	75	M13	R14	94	K3	106	R14	PB14	I/O	FT		TIM1_CH2N, TIM8_CH2N, SPI2_MISO, I2S2ext_SD, USART3_RTS, TIM12_CH1, OTG_HS_DM, EVENTOUT	
54	76	L13	R15	95	J3	107	R15	PB15	I/O	FT		RTC_REFIN, TIM1_CH3N, TIM8_CH3N, SPI2_MOSI/I2S2_SD, TIM12_CH2, OTG_HS_DP, EVENTOUT	
55	77	L12	P15	96	L2	108	L15	PD8	I/O	FT		USART3_TX, FMC_D13, EVENTOUT	
56	78	K13	P14	97	M1	109	L14	PD9	I/O	FT		USART3_RX, FMC_D14, EVENTOUT	
57	79	K11	N15	98	H4	110	K15	PD10	I/O	FT		USART3_CK, FMC_D15, LCD_B3, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
58	80	H10	N14	99	K2	111	N10	PD11	I/O	FT		USART3_CTS, FMC_A16, EVENTOUT	
59	81	J13	N13	100	H6	112	M10	PD12	I/O	FT		TIM4_CH1, USART3_RTS, FMC_A17, EVENTOUT	
60	82	K12	M15	101	H5	113	M11	PD13	I/O	FT		TIM4_CH2, FMC_A18, EVENTOUT	
-	83	-	-	102	-	114	J10	V <sub>SS</sub>	S				
-	84	F7	J13	103	L1	115	J11	V <sub>DD</sub>	S				
61	85	H11	M14	104	J2	116	L12	PD14	I/O	FT		TIM4_CH3, FMC_D0, EVENTOUT	
62	86	J12	L14	105	K1	117	K13	PD15	I/O	FT		TIM4_CH4, FMC_D1, EVENTOUT	
-	-	-	-	-	-	118	K12	PJ6	I/O	FT		LCD_R7, EVENTOUT	
-	-	-	-	-	-	119	J12	PJ7	I/O	FT		LCD_G0, EVENTOUT	
-	-	-	-	-	-	120	H12	PJ8	I/O	FT		LCD_G1, EVENTOUT	
-	-	-	-	-	-	121	J13	PJ9	I/O	FT		LCD_G2, EVENTOUT	
-	-	-	-	-	-	122	H13	PJ10	I/O	FT		LCD_G3, EVENTOUT	
-	-	-	-	-	-	123	G12	PJ11	I/O	FT		LCD_G4, EVENTOUT	
-	-	-	-	-	-	124	H11	VDD	I/O	FT			
-	-	-	-	-	-	125	H10	VSS	I/O	FT			
-	-	-	-	-	-	126	G13	PK0	I/O	FT		LCD_G5, EVENTOUT	
-	-	-	-	-	-	127	F12	PK1	I/O	FT		LCD_G6, EVENTOUT	
-	-	-	-	-	-	128	F13	PK2	I/O	FT		LCD_G7, EVENTOUT	
-	87	H13	L15	106	J1	129	M13	PG2	I/O	FT		FMC_A12, EVENTOUT	
-	88	NC (2)	K15	107	G3	130	M12	PG3	I/O	FT		FMC_A13, EVENTOUT	
-	89	H12	K14	108	G5	131	N12	PG4	I/O	FT		FMC_A14/FMC_BA0, EVENTOUT	
-	90	G13	K13	109	G6	132	N11	PG5	I/O	FT		FMC_A15/FMC_BA1, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
-	91	G11	J15	110	G4	133	J15	PG6	I/O	FT		FMC_INT2, DCMI_D12, LCD_R7, EVENTOUT	
-	92	G12	J14	111	H1	134	J14	PG7	I/O	FT		USART6_CK, FMC_INT3, DCMI_D13, LCD_CLK, EVENTOUT	
-	93	F13	H14	112	G2	135	H14	PG8	I/O	FT		SPI6_NSS, USART6_RTS, ETH_PPS_OUT, FMC_SDCLK, EVENTOUT	
-	94	J7	G12	113	D2	136	G10	V <sub>SS</sub>	S				
-	95	E6	H13	114	G1	137	G11	V <sub>DD</sub>	S				
63	96	F9	H15	115	F2	138	H15	PC6	I/O	FT		TIM3_CH1, TIM8_CH1, I2S2_MCK, USART6_TX, SDIO_D6, DCMI_D0, LCD_HSYNC, EVENTOUT	
64	97	F10	G15	116	F3	139	G15	PC7	I/O	FT		TIM3_CH2, TIM8_CH2, I2S3_MCK, USART6_RX, SDIO_D7, DCMI_D1, LCD_G6, EVENTOUT	
65	98	F11	G14	117	E4	140	G14	PC8	I/O	FT		TIM3_CH3, TIM8_CH3, USART6_CK, SDIO_D0, DCMI_D2, EVENTOUT	
66	99	F12	F14	118	E3	141	F14	PC9	I/O	FT		MCO2, TIM3_CH4, TIM8_CH4, I2C3_SDA, I2S_CKIN, SDIO_D1, DCMI_D3, EVENTOUT	
67	100	E13	F15	119	F1	142	F15	PA8	I/O	FT		MCO1, TIM1_CH1, I2C3_SCL, USART1_CK, OTG_FS_SOF, LCD_R6, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
68	101	E8	E15	120	E2	143	E15	PA9	I/O	FT		TIM1_CH2, I2C3_SMBA, USART1_TX, DCMI_D0, EVENTOUT	OTG_FS_ VBUS
69	102	E9	D15	121	D5	144	D15	PA10	I/O	FT		TIM1_CH3, USART1_RX, OTG_FS_ID, DCMI_D1, EVENTOUT	
70	103	E10	C15	122	D4	145	C15	PA11	I/O	FT		TIM1_CH4, USART1_CTS, CAN1_RX, LCD_R4, OTG_FS_DM, EVENTOUT	
71	104	E11	B15	123	E1	146	B15	PA12	I/O	FT		TIM1_ETR, USART1_RTS, CAN1_TX, LCD_R5, OTG_FS_DP, EVENTOUT	
72	105	E12	A15	124	D3	147	A15	PA13 (JTMS- SWDIO)	I/O	FT		JTMS-SWDIO, EVENTOUT	
73	106	D12	F13	125	D1	148	E11	V <sub>CAP_2</sub>	S				
74	107	J10	F12	126	D2	149	F10	V <sub>SS</sub>	S				
75	108	H4	G13	127	C1	150	F11	V <sub>DD</sub>	S				
-	-	D13	E12	128	-	151	E12	PH13	I/O	FT		TIM8_CH1N, CAN1_TX, FMC_D21, LCD_G2, EVENTOUT	
-	-	C13	E13	129	-	152	E13	PH14	I/O	FT		TIM8_CH2N, FMC_D22, DCMI_D4, LCD_G3, EVENTOUT	
-	-	C12	D13	130	-	153	D13	PH15	I/O	FT		TIM8_CH3N, FMC_D23, DCMI_D11, LCD_G4, EVENTOUT	
-	-	B13	E14	131	-	154	E14	PI0	I/O	FT		TIM5_CH4, SPI2_NSS/I2S2_WS <sup>(7)</sup> , FMC_D24, DCMI_D13, LCD_G5, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
-	-	C11	D14	132	-	155	D14	PI1	I/O	FT		SPI2_SCK/I2S2_CK <sup>(7)</sup> , FMC_D25, DCMI_D8, LCD_G6, EVENTOUT	
-	-	B12	C14	133	-	156	C14	PI2	I/O	FT		TIM8_CH4, SPI2_MISO, I2S2ext_SD, FMC_D26, DCMI_D9, LCD_G7, EVENTOUT	
-	-	A12	C13	134	-	157	C13	PI3	I/O	FT		TIM8_ETR, SPI2_MOSI/I2S2_SD, FMC_D27, DCMI_D10, EVENTOUT	
-	-	D11	D9	135	F5	-	F9	V <sub>SS</sub>	S				
-	-	D3	C9	136	A1	158	E10	V <sub>DD</sub>	S				
76	109	A11	A14	137	B1	159	A14	PA14 (JTCK- SWCLK)	I/O	FT		JTCK-SWCLK/ EVENTOUT	
77	110	B11	A13	138	C2	160	A13	PA15 (JTDI)	I/O	FT		JTDI, TIM2_CH1/TIM2_ETR, SPI1_NSS, SPI3_NSS/I2S3_WS, EVENTOUT	
78	111	C10	B14	139	A2	161	B14	PC10	I/O	FT		SPI3_SCK/I2S3_CK, USART3_TX, UART4_TX, SDIO_D2, DCMI_D8, LCD_R2, EVENTOUT	
79	112	B10	B13	140	B2	162	B13	PC11	I/O	FT		I2S3ext_SD, SPI3_MISO, USART3_RX, UART4_RX, SDIO_D3, DCMI_D4, EVENTOUT	
80	113	A10	A12	141	C3	163	A12	PC12	I/O	FT		SPI3_MOSI/I2S3_SD, USART3_CK, UART5_TX, SDIO_CK, DCMI_D9, EVENTOUT	
81	114	D9	B12	142	B3	164	B12	PD0	I/O	FT		CAN1_RX, FMC_D2, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
82	115	C9	C12	143	C4	165	C12	PD1	I/O	FT		CAN1_TX, FMC_D3, EVENTOUT	
83	116	B9	D12	144	A3	166	D12	PD2	I/O	FT		TIM3_ETR, UART5_RX, SDIO_CMD, DCMI_D11, EVENTOUT	
84	117	A9	D11	145	B4	167	C11	PD3	I/O	FT		SPI2_SCK/I2S2_CK, USART2_CTS, FMC_CLK, DCMI_D5, LCD_G7, EVENTOUT	
85	118	D8	D10	146	B5	168	D11	PD4	I/O	FT		USART2_RTS, FMC_NOE, EVENTOUT	
86	119	C8	C11	147	A4	169	C10	PD5	I/O	FT		USART2_TX, FMC_NWE, EVENTOUT	
-	120	-	D8	148	-	170	F8	V <sub>SS</sub>	S				
-	121	D6	C8	149	C5	171	E9	V <sub>DD</sub>	S				
87	122	B8	B11	150	F4	172	B11	PD6	I/O	FT		SPI3_MOSI/I2S3_SD, SAI1_SD_A, USART2_RX, FMC_NWAIT, DCMI_D10, LCD_B2, EVENTOUT	
88	123	A8	A11	151	A5	173	A11	PD7	I/O	FT		USART2_CK, FMC_NE1/FMC_NCE2, EVENTOUT	
-	-	-	-	-	-	174	B10	PJ12	I/O	FT		LCD_B0, EVENTOUT	
-	-	-	-	-	-	175	B9	PJ13	I/O	FT		LCD_B1, EVENTOUT	
-	-	-	-	-	-	176	C9	PJ14	I/O	FT		LCD_B2, EVENTOUT	
-	-	-	-	-	-	177	D10	PJ15	I/O	FT		LCD_B3, EVENTOUT	
-	124	NC (2)	C10	152	E5	178	D9	PG9	I/O	FT		USART6_RX, FMC_NE2/FMC_NCE3, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
-	125	C7	B10	153	C6	179	C8	PG10	I/O	FT		LCD_G3, FMC_NCE4_1/FMC_N E3, DCMI_D2, LCD_B2, EVENTOUT	
-	126	B7	B9	154	B6	180	B8	PG11	I/O	FT		ETH_MII_TX_EN/ETH_ RMII_TX_EN, FMC_NCE4_2, DCMI_D3, LCD_B3, EVENTOUT	
-	127	A7	B8	155	A6	181	C7	PG12	I/O	FT		SPI6_MISO, USART6_RTS, LCD_B4, FMC_NE4, LCD_B1, EVENTOUT	
-	128	NC (2)	A8	156	D6	182	B3	PG13	I/O	FT		SPI6_SCK, USART6_CTS, ETH_MII_TXD0/ETH_R MII_TXD0, FMC_A24, EVENTOUT	
-	129	NC (2)	A7	157	F6	183	A4	PG14	I/O	FT		SPI6_MOSI, USART6_TX, ETH_MII_TXD1/ETH_R MII_TXD1, FMC_A25, EVENTOUT	
-	130	D7	D7	158	-	184	F7	V <sub>SS</sub>	S				
-	131	L6	C7	159	E6	185	E8	V <sub>DD</sub>	S				
-	-	-	-	-	-	186	D8	PK3	I/O	FT		LCD_B4, EVENTOUT	
-	-	-	-	-	-	187	D7	PK4	I/O	FT		LCD_B5, EVENTOUT	
-	-	-	-	-	-	188	C6	PK5	I/O	FT		LCD_B6, EVENTOUT	
-	-	-	-	-	-	189	C5	PK6	I/O	FT		LCD_B7, EVENTOUT	
-	-	-	-	-	-	190	C4	PK7	I/O	FT		LCD_DE, EVENTOUT	
-	132	C6	B7	160	A7	191	B7	PG15	I/O	FT		USART6_CTS, FMC_SDNCAS, DCMI_D13, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
89	133	B6	A10	161	B7	192	A10	PB3 (JTDO/TRACE SWO)	I/O	FT		JTDO/TRACESWO, TIM2_CH2, SPI1_SCK, SPI3_SCK/I2S3_CK, EVENTOUT	
90	134	A6	A9	162	C7	193	A9	PB4 (NJTRST)	I/O	FT		NJTRST, TIM3_CH1, SPI1_MISO, SPI3_MISO, I2S3ext_SD, EVENTOUT	
91	135	D5	A6	163	C8	194	A8	PB5	I/O	FT		TIM3_CH2, I2C1_SMBA, SPI1_MOSI, SPI3_MOSI/I2S3_SD, CAN2_RX, OTG_HS_ULPI_D7, ETH_PPS_OUT, FMC_SDCKE1, DCMI_D10, EVENTOUT	
92	136	C5	B6	164	A8	195	B6	PB6	I/O	FT		TIM4_CH1, I2C1_SCL, USART1_TX, CAN2_TX, FMC_SDNE1, DCMI_D5, EVENTOUT	
93	137	B5	B5	165	B8	196	B5	PB7	I/O	FT		TIM4_CH2, I2C1_SDA, USART1_RX, FMC_NL, DCMI_VSYNC, EVENTOUT	
94	138	A5	D6	166	C9	197	E6	BOOT0	I	B			V <sub>PP</sub>
95	139	D4	A5	167	A9	198	A7	PB8	I/O	FT		TIM4_CH3, TIM10_CH1, I2C1_SCL, CAN1_RX, ETH_MII_TXD3, SDIO_D4, DCMI_D6, LCD_B6, EVENTOUT	

Table 10. STM32F437xx and STM32F439xx pin and ball definitions (continued)

Pin number								Pin name (function after reset) <sup>(1)</sup>	Pin type	I / O structure	Notes	Alternate functions	Additional functions
LQFP100	LQFP144	UFBGA169	UFBGA176	LQFP176	WLCSP143	LQFP208	TFBGA216						
96	140	C4	B4	168	B9	199	B4	PB9	I/O	FT		TIM4_CH4, TIM11_CH1, I2C1_SDA, SPI2_NSS/I2S2_WS, CAN1_TX, SDIO_D5, DCMI_D7, LCD_B7, EVENTOUT	
97	141	B4	A4	169	B10	200	A6	PE0	I/O	FT		TIM4_ETR, UART8_RX, FMC_NBL0, DCMI_D2, EVENTOUT	
98	142	A4	A3	170	A10	201	A5	PE1	I/O	FT		UART8_Tx, FMC_NBL1, DCMI_D3, EVENTOUT	
99	-	F5	D5	-	-	202	F6	V <sub>SS</sub>	S				
-	143	C3	C6	171	A11	203	E5	PDR_ON	S				
100	144	K6	C5	172	D7	204	E7	V <sub>DD</sub>	S				
-	-	B3	D4	173	-	205	C3	PI4	I/O	FT		TIM8_BKIN, FMC_NBL2, DCMI_D5, LCD_B4, EVENTOUT	
-	-	A3	C4	174	-	206	D3	PI5	I/O	FT		TIM8_CH1, FMC_NBL3, DCMI_VSYNC, LCD_B5, EVENTOUT	
-	-	A2	C3	175	-	207	D6	PI6	I/O	FT		TIM8_CH2, FMC_D28, DCMI_D6, LCD_B6, EVENTOUT	
-	-	B1	C2	176	-	208	D4	PI7	I/O	FT		TIM8_CH3, FMC_D29, DCMI_D7, LCD_B7, EVENTOUT	

- Function availability depends on the chosen device.
- NC (not-connected) pins are not bonded. They must be configured by software to output push-pull and forced to 0 in the output data register to avoid extra current consumption in low power modes.
- PC13, PC14, PC15 and PI8 are supplied through the power switch. Since the switch only sinks a limited amount of current (3 mA), the use of GPIOs PC13 to PC15 and PI8 in output mode is limited:
  - The speed should not exceed 2 MHz with a maximum load of 30 pF.
  - These I/Os must not be used as a current source (e.g. to drive an LED).

4. Main function after the first backup domain power-up. Later on, it depends on the contents of the RTC registers even after reset (because these registers are not reset by the main reset). For details on how to manage these I/Os, refer to the RTC register description sections in the STM32F4xx reference manual, available from the STMicroelectronics website: [www.st.com](http://www.st.com).
5. FT = 5 V tolerant except when in analog mode or oscillator mode (for PC14, PC15, PH0 and PH1).
6. If the device is delivered in an UFBGA176, LQFP176 or TFBGA216 package, and the BYPASS\_REG pin is set to V<sub>DD</sub> (Regulator OFF/internal reset ON mode), then PA0 is used as an internal Reset (active low).
7. PI0 and PI1 cannot be used for I2S2 full-duplex mode.

Table 11. FMC pin definition

Pin name	CF	NOR/PSRAM/ SRAM	NOR/PSRAM Mux	NAND16	SDRAM
PF0	A0	A0			A0
PF1	A1	A1			A1
PF2	A2	A2			A2
PF3	A3	A3			A3
PF4	A4	A4			A4
PF5	A5	A5			A5
PF12	A6	A6			A6
PF13	A7	A7			A7
PF14	A8	A8			A8
PF15	A9	A9			A9
PG0	A10	A10			A10
PG1		A11			A11
PG2		A12			A12
PG3		A13			
PG4		A14			BA0
PG5		A15			BA1
PD11		A16	A16	CLE	
PD12		A17	A17	ALE	
PD13		A18	A18		
PE3		A19	A19		
PE4		A20	A20		
PE5		A21	A21		
PE6		A22	A22		
PE2		A23	A23		
PG13		A24	A24		
PG14		A25	A25		
PD14	D0	D0	DA0	D0	D0
PD15	D1	D1	DA1	D1	D1
PD0	D2	D2	DA2	D2	D2
PD1	D3	D3	DA3	D3	D3
PE7	D4	D4	DA4	D4	D4
PE8	D5	D5	DA5	D5	D5
PE9	D6	D6	DA6	D6	D6
PE10	D7	D7	DA7	D7	D7

Table 11. FMC pin definition (continued)

Pin name	CF	NOR/PSRAM/ SRAM	NOR/PSRAM Mux	NAND16	SDRAM
PE11	D8	D8	DA8	D8	D8
PE12	D9	D9	DA9	D9	D9
PE13	D10	D10	DA10	D10	D10
PE14	D11	D11	DA11	D11	D11
PE15	D12	D12	DA12	D12	D12
PD8	D13	D13	DA13	D13	D13
PD9	D14	D14	DA14	D14	D14
PD10	D15	D15	DA15	D15	D15
PH8		D16			D16
PH9		D17			D17
PH10		D18			D18
PH11		D19			D19
PH12		D20			D20
PH13		D21			D21
PH14		D22			D22
PH15		D23			D23
PI0		D24			D24
PI1		D25			D25
PI2		D26			D26
PI3		D27			D27
PI6		D28			D28
PI7		D29			D29
PI9		D30			D30
PI10		D31			D31
PD7		NE1	NE1	NCE2	
PG9		NE2	NE2	NCE3	
PG10	NCE4_1	NE3	NE3		
PG11	NCE4_2				
PG12		NE4	NE4		
PD3		CLK	CLK		
PD4	NOE	NOE	NOE	NOE	
PD5	NWE	NWE	NWE	NWE	
PD6	NWAIT	NWAIT	NWAIT	NWAIT	
PB7		NADV	NADV		

Table 11. FMC pin definition (continued)

Pin name	CF	NOR/PSRAM/ SRAM	NOR/PSRAM Mux	NAND16	SDRAM
PF6	NIORD				
PF7	NREG				
PF8	NIOWR				
PF9	CD				
PF10	INTR				
PG6				INT2	
PG7				INT3	
PE0		NBL0	NBL0		NBL0
PE1		NBL1	NBL1		NBL1
PI4		NBL2			NBL2
PI5		NBL3			NBL3
PG8					SDCLK
PC0					SDNWE
PF11					SDNRAS
PG15					SDNCAS
PH2					SDCKE0
PH3					SDNE0
PH6					SDNE1
PH7					SDCKE1
PH5					SDNWE
PC2					SDNE0
PC3					SDCKE0
PB5					SDCKE1
PB6					SDNE1



**Table 12. STM32F437xx and STM32F439xx alternate function mapping**

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS	
Port A	PA0	-	TIM2_ CH1/TIM2_ ETR	TIM5_ CH1	TIM8_ ETR	-	-	-	USART2_ CTS	UART4_TX	-	-	ETH_MII_ CRS	-	-	-	EVEN TOUT	
	PA1	-	TIM2_ CH2	TIM5_ CH2	-	-	-	-	USART2_ RTS	UART4_RX	-	-	ETH_MII_ RX_CLK/E TH_RMII_ REF_CLK	-	-	-	EVEN TOUT	
	PA2	-	TIM2_ CH3	TIM5_ CH3	TIM9_ CH1	-	-	-	USART2_ TX	-	-	-	ETH_ MDIO	-	-	-	EVEN TOUT	
	PA3	-	TIM2_ CH4	TIM5_ CH4	TIM9_ CH2	-	-	-	USART2_ RX	-	-	OTG_HS_ ULPI_D0	ETH_MII_ COL	-	-	LCD_B5	EVEN TOUT	
	PA4	-	-	-	-	-	SPI1_ NSS	SPI3_ NSS/ I2S3_WS	USART2_ CK	-	-	-	-	OTG_HS_ SOF	DCMI_ HSYNC	LCD_ VSYNC	EVEN TOUT	
	PA5	-	TIM2_ CH1/TIM2_ ETR	-	TIM8_ CH1N	-	SPI1_ SCK	-	-	-	-	OTG_HS_ ULPI_CK	-	-	-	-	-	EVEN TOUT
	PA6	-	TIM1_ BKIN	TIM3_ CH1	TIM8_ BKIN	-	SPI1_ MISO	-	-	-	-	TIM13_CH1	-	-	DCMI_ PIXCLK	LCD_G2	EVEN TOUT	
	PA7	-	TIM1_ CH1N	TIM3_ CH2	TIM8_ CH1N	-	SPI1_ MOSI	-	-	-	-	TIM14_CH1	-	ETH_MII_ RX_DV/ ETH_RMII_ CRS_DV	-	-	-	EVEN TOUT
	PA8	MCO1	TIM1_ CH1	-	-	I2C3_ SCL	-	-	USART1_ CK	-	-	-	OTG_FS_ SOF	-	-	-	LCD_R6	EVEN TOUT
	PA9	-	TIM1_ CH2	-	-	I2C3_ SMBA	-	-	USART1_ TX	-	-	-	-	-	DCMI_ D0	-	-	EVEN TOUT
	PA10	-	TIM1_ CH3	-	-	-	-	-	USART1_ RX	-	-	-	OTG_FS_ ID	-	DCMI_ D1	-	-	EVEN TOUT
	PA11	-	TIM1_ CH4	-	-	-	-	-	USART1_ CTS	-	-	CAN1_RX	OTG_FS_ DM	-	-	-	LCD_R4	EVEN TOUT
PA12	-	TIM1_ ETR	-	-	-	-	-	USART1_ RTS	-	-	CAN1_TX	OTG_FS_ DP	-	-	-	LCD_R5	EVEN TOUT	



Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15		
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS		
Port A	PA13	JTMS- SWDI O	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT	
	PA14	JTCK- SWCL K	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT	
	PA15	JTDI	TIM2_ CH1/TIM2 _ETR	-	-	-	SPI1_ NSS	SPI3_ NSS/ I2S3_WS	-	-	-	-	-	-	-	-	-	-	EVEN TOUT
Port B	PB0	-	TIM1_ CH2N	TIM3_ CH3	TIM8_ CH2N	-	-	-	-	-	LCD_R3	OTG_HS_ ULPI_D1	ETH_MII_ RXD2	-	-	-	-	EVEN TOUT	
	PB1	-	TIM1_ CH3N	TIM3_ CH4	TIM8_ CH3N	-	-	-	-	-	LCD_R6	OTG_HS_ ULPI_D2	ETH_MII_ RXD3	-	-	-	-	EVEN TOUT	
	PB2	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT	
	PB3	JTDO/ TRAC ESWO	TIM2_ CH2	-	-	-	SPI1_ SCK	SPI3_ SCK/ I2S3_CK	-	-	-	-	-	-	-	-	-	-	EVEN TOUT
	PB4	NJTR ST	-	TIM3_ CH1	-	-	SPI1_ MISO	SPI3_ MISO	I2S3ext_ SD	-	-	-	-	-	-	-	-	-	EVEN TOUT
	PB5	-	-	TIM3_ CH2	-	I2C1_ SMBA	SPI1_ MOSI	SPI3_ MOSI/ I2S3_SD	-	-	-	CAN2_RX	OTG_HS_ ULPI_D7	ETH_PPS_ _OUT	FMC_ SDCKE1	DCMI_ D10	-	-	EVEN TOUT
	PB6	-	-	TIM4_ CH1	-	I2C1_ SCL	-	-	USART1_ TX	-	-	CAN2_TX	-	-	FMC_ SDNE1	DCMI_ D5	-	-	EVEN TOUT
	PB7	-	-	TIM4_ CH2	-	I2C1_ SDA	-	-	USART1_ RX	-	-	-	-	-	FMC_NL	DCMI_ VSYNC	-	-	EVEN TOUT
	PB8	-	-	TIM4_ CH3	TIM10_ CH1	I2C1_ SCL	-	-	-	-	-	CAN1_RX	-	ETH_MII_ TXD3	SDIO_D4	DCMI_ D6	LCD_B6	-	EVEN TOUT
	PB9	-	-	TIM4_ CH4	TIM11_ CH1	I2C1_ SDA	SPI2_ NSS/I2 S2_WS	-	-	-	-	CAN1_TX	-	-	SDIO_D5	DCMI_ D7	LCD_B7	-	EVEN TOUT
PB10	-	TIM2_ CH3	-	-	I2C2_ SCL	SPI2_ SCK/I2 S2_CK	-	USART3_ TX	-	-	-	OTG_HS_ ULPI_D3	ETH_MII_ RX_ER	-	-	LCD_G4	-	EVEN TOUT	



**Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)**

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS
Port B	PB11	-	TIM2_CH4	-	-	I2C2_SDA	-	-	USART3_RX	-	-	OTG_HS_ULPI_D4	ETH_MII_TX_EN/ ETH_RMII_TX_EN	-	-	LCD_G5	EVEN TOUT
	PB12	-	TIM1_BKIN	-	-	I2C2_SMBA	SPI2_NSS/I2 S2_WS	-	USART3_CK	-	CAN2_RX	OTG_HS_ULPI_D5	ETH_MII_TXD0/ETH RMII_TXD0	OTG_HS_ID	-	-	EVEN TOUT
	PB13	-	TIM1_CH1N	-	-	-	SPI2_SCK/I2 S2_CK	-	USART3_CTS	-	CAN2_TX	OTG_HS_ULPI_D6	ETH_MII_TXD1/ETH RMII_TXD1	-	-	-	EVEN TOUT
	PB14	-	TIM1_CH2N	-	TIM8_CH2N	-	SPI2_MISO	I2S2ext_SD	USART3_RTS	-	TIM12_CH1	-	-	OTG_HS_DM	-	-	EVEN TOUT
	PB15	RTC_REFIN	TIM1_CH3N	-	TIM8_CH3N	-	SPI2_MOSI/I2 S2_SD	-	-	-	TIM12_CH2	-	-	OTG_HS_DP	-	-	EVEN TOUT
Port C	PC0	-	-	-	-	-	-	-	-	-	-	OTG_HS_ULPI_STP	-	FMC_SDNWE	-	-	EVEN TOUT
	PC1	-	-	-	-	-	-	-	-	-	-	-	ETH_MDC	-	-	-	EVEN TOUT
	PC2	-	-	-	-	-	SPI2_MISO	I2S2ext_SD	-	-	-	OTG_HS_ULPI_DIR	ETH_MII_TXD2	FMC_SDNE0	-	-	EVEN TOUT
	PC3	-	-	-	-	-	SPI2_MOSI/I2 S2_SD	-	-	-	-	OTG_HS_ULPI_NXT	ETH_MII_TX_CLK	FMC_SDCKE0	-	-	EVEN TOUT
	PC4	-	-	-	-	-	-	-	-	-	-	-	ETH_MII_RXD0/ETH RMII_RXD0	-	-	-	EVEN TOUT
	PC5	-	-	-	-	-	-	-	-	-	-	-	ETH_MII_RXD1/ETH RMII_RXD1	-	-	-	EVEN TOUT
	PC6	-	-	TIM3_CH1	TIM8_CH1	-	I2S2_MCK	-	-	USART6_TX	-	-	-	SDIO_D6	DCMI_D0	LCD_HSYNC	EVEN TOUT
	PC7	-	-	TIM3_CH2	TIM8_CH2	-	-	I2S3_MCK	-	USART6_RX	-	-	-	SDIO_D7	DCMI_D1	LCD_G6	EVEN TOUT



Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS	
Port C	PC8	-	-	TIM3_ CH3	TIM8_ CH3	-	-	-	-	USART6_ CK	-	-	-	SDIO_D0	DCMI_ D2	-	EVEN TOUT	
	PC9	MCO2	-	TIM3_ CH4	TIM8_ CH4	I2C3_ SDA	I2S_ CKIN	-	-	-	-	-	-	SDIO_D1	DCMI_ D3	-	EVEN TOUT	
	PC10	-	-	-	-	-	-	SPI3_ SCK/I2S 3_CK	USART3_ TX	UART4_TX	-	-	-	SDIO_D2	DCMI_ D8	LCD_R2	EVEN TOUT	
	PC11	-	-	-	-	-	I2S3ext _SD	SPI3_ MISO	USART3_ RX	UART4_RX	-	-	-	SDIO_D3	DCMI_ D4	-	EVEN TOUT	
	PC12	-	-	-	-	-	-	SPI3_ MOSI/I2 S3_SD	USART3_ CK	UART5_TX	-	-	-	SDIO_CK	DCMI_ D9	-	EVEN TOUT	
	PC13	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT
	PC14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT
	PC15	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT
Port D	PD0	-	-	-	-	-	-	-	-	-	-	CAN1_RX	-	FMC_D2	-	-	EVEN TOUT	
	PD1	-	-	-	-	-	-	-	-	-	-	CAN1_TX	-	FMC_D3	-	-	EVEN TOUT	
	PD2	-	-	TIM3_ ETR	-	-	-	-	-	UART5_RX	-	-	-	SDIO_ CMD	DCMI_ D11	-	EVEN TOUT	
	PD3	-	-	-	-	-	-	SPI2_S CK/I 2S2_CK	-	USART2_ CTS	-	-	-	FMC_CLK	DCMI_ D5	LCD_G7	EVEN TOUT	
	PD4	-	-	-	-	-	-	-	USART2_ RTS	-	-	-	-	FMC_NOE	-	-	EVEN TOUT	
	PD5	-	-	-	-	-	-	-	USART2_ TX	-	-	-	-	FMC_NWE	-	-	EVEN TOUT	
	PD6	-	-	-	-	-	-	SPI3_ MOSI/I2 S3_SD	SAI1_ SD_A	USART2_ RX	-	-	-	-	FMC_ NWAIT	DCMI_ D10	LCD_B2	EVEN TOUT



**Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)**

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS
Port D	PD7	-	-	-	-	-	-	-	USART2_ CK	-	-	-	-	FMC_NE1/ FMC NCE2	-	-	EVEN TOUT
	PD8	-	-	-	-	-	-	-	USART3_ TX	-	-	-	-	FMC_D13	-	-	EVEN TOUT
	PD9	-	-	-	-	-	-	-	USART3_ RX	-	-	-	-	FMC_D14	-	-	EVEN TOUT
	PD10	-	-	-	-	-	-	-	USART3_ CK	-	-	-	-	FMC_D15	-	LCD_B3	EVEN TOUT
	PD11	-	-	-	-	-	-	-	USART3_ CTS	-	-	-	-	FMC_A16	-	-	EVEN TOUT
	PD12	-	-	TIM4_ CH1	-	-	-	-	USART3_ RTS	-	-	-	-	FMC_A17	-	-	EVEN TOUT
	PD13	-	-	TIM4_ CH2	-	-	-	-	-	-	-	-	-	FMC_A18	-	-	EVEN TOUT
	PD14	-	-	TIM4_ CH3	-	-	-	-	-	-	-	-	-	FMC_D0	-	-	EVEN TOUT
	PD15	-	-	TIM4_ CH4	-	-	-	-	-	-	-	-	-	FMC_D1	-	-	EVEN TOUT
Port E	PE0	-	-	TIM4_ ETR	-	-	-	-	-	UART8_Rx	-	-	-	FMC_ NBL0	DCMI_ D2	-	EVEN TOUT
	PE1	-	-	-	-	-	-	-	-	UART8_Tx	-	-	-	FMC_ NBL1	DCMI_ D3	-	EVEN TOUT
	PE2	TRAC ECLK	-	-	-	-	SPI4_ SCK	SAI1_ MCLK_A	-	-	-	-	ETH_MII_ TXD3	FMC_A23	-	-	EVEN TOUT
	PE3	TRAC ED0	-	-	-	-	-	SAI1_ SD_B	-	-	-	-	-	FMC_A19	-	-	EVEN TOUT
	PE4	TRAC ED1	-	-	-	-	SPI4_ NSS	SAI1_ FS_A	-	-	-	-	-	FMC_A20	DCMI_ D4	LCD_B0	EVEN TOUT
	PE5	TRAC ED2	-	-	TIM9_ CH1	-	SPI4_M ISO	SAI1_ SCK_A	-	-	-	-	-	FMC_A21	DCMI_ D6	LCD_G0	EVEN TOUT
	PE6	TRAC ED3	-	-	TIM9_ CH2	-	SPI4_ MOSI	SAI1_ SD_A	-	-	-	-	-	FMC_A22	DCMI_ D7	LCD_G1	EVEN TOUT



Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS	
Port E	PE7	-	TIM1_ETR	-	-	-	-	-	-	UART7_Rx	-	-	-	FMC_D4	-	-	EVEN TOUT	
	PE8	-	TIM1_CH1N	-	-	-	-	-	-	UART7_Tx	-	-	-	FMC_D5	-	-	EVEN TOUT	
	PE9	-	TIM1_CH1	-	-	-	-	-	-	-	-	-	-	FMC_D6	-	-	EVEN TOUT	
	PE10	-	TIM1_CH2N	-	-	-	-	-	-	-	-	-	-	FMC_D7	-	-	EVEN TOUT	
	PE11	-	TIM1_CH2	-	-	-	SPI4_NSS	-	-	-	-	-	-	FMC_D8	-	LCD_G3	EVEN TOUT	
	PE12	-	TIM1_CH3N	-	-	-	SPI4_SCK	-	-	-	-	-	-	FMC_D9	-	LCD_B4	EVEN TOUT	
	PE13	-	TIM1_CH3	-	-	-	SPI4_MISO	-	-	-	-	-	-	FMC_D10	-	LCD_DE	EVEN TOUT	
	PE14	-	TIM1_CH4	-	-	-	SPI4_MOSI	-	-	-	-	-	-	FMC_D11	-	LCD_CLK	EVEN TOUT	
	PE15	-	TIM1_BKIN	-	-	-	-	-	-	-	-	-	-	FMC_D12	-	LCD_R7	EVEN TOUT	
Port F	PF0	-	-	-	-	I2C2_SDA	-	-	-	-	-	-	-	FMC_A0	-	-	EVEN TOUT	
	PF1	-	-	-	-	I2C2_SCL	-	-	-	-	-	-	-	FMC_A1	-	-	EVEN TOUT	
	PF2	-	-	-	-	I2C2_SMBA	-	-	-	-	-	-	-	FMC_A2	-	-	EVEN TOUT	
	PF3	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A3	-	-	EVEN TOUT	
	PF4	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A4	-	-	EVEN TOUT	
	PF5	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A5	-	-	EVEN TOUT	
	PF6	-	-	-	TIM10_CH1	-	SPI5_NSS	SAI1_SD_B	-	UART7_Rx	-	-	-	-	FMC_NIORD	-	-	EVEN TOUT
	PF7	-	-	-	TIM11_CH1	-	SPI5_SCK	SAI1_MCLK_B	-	UART7_Tx	-	-	-	-	FMC_NREG	-	-	EVEN TOUT



**Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)**

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS
Port F	PF8	-	-	-	-	-	SPI5_ MISO	SAI1_ SCK_B	-	-	TIM13_CH1	-	-	FMC_ NIOWR	-	-	EVEN TOUT
	PF9	-	-	-	-	-	SPI5_ MOSI	SAI1_ FS_B	-	-	TIM14_CH1	-	-	FMC_CD	-	-	EVEN TOUT
	PF10	-	-	-	-	-	-	-	-	-	-	-	-	FMC_INTR	DCMI_ D11	LCD_DE	EVEN TOUT
	PF11	-	-	-	-	-	SPI5_ MOSI	-	-	-	-	-	-	FMC_ SDNRAS	DCMI_ D12	-	EVEN TOUT
	PF12	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A6	-	-	EVEN TOUT
	PF13	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A7	-	-	EVEN TOUT
	PF14	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A8	-	-	EVEN TOUT
	PF15	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A9	-	-	EVEN TOUT
Port G	PG0	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A10	-	-	EVEN TOUT
	PG1	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A11	-	-	EVEN TOUT
	PG2	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A12	-	-	EVEN TOUT
	PG3	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A13	-	-	EVEN TOUT
	PG4	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A14/ FMC_BA0	-	-	EVEN TOUT
	PG5	-	-	-	-	-	-	-	-	-	-	-	-	FMC_A15/ FMC_BA1	-	-	EVEN TOUT
	PG6	-	-	-	-	-	-	-	-	-	-	-	-	FMC_INT2	DCMI_ D12	LCD_R7	EVEN TOUT
	PG7	-	-	-	-	-	-	-	-	USART6_ CK	-	-	-	FMC_INT3	DCMI_ D13	LCD_ CLK	EVEN TOUT
	PG8	-	-	-	-	-	SPI6_ NSS	-	-	USART6_ RTS	-	-	-	ETH_PPS_ OUT	FMC_SDC LK	-	-



Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15	
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS	
Port G	PG9	-	-	-	-	-	-	-	-	USART6_ RX	-	-	-	FMC_NE2/ FMC_ NCE3	-	-	EVEN TOUT	
	PG10	-	-	-	-	-	-	-	-	-	LCD_G3	-	-	FMC_ NCE4_1/ FMC_NE3	DCMI_ D2	LCD_B2	EVEN TOUT	
	PG11	-	-	-	-	-	-	-	-	-	-	-	ETH_MII_ TX_EN/ ETH_RMII_ TX_EN	FMC_ NCE4_2	DCMI_ D3	LCD_B3	EVEN TOUT	
	PG12	-	-	-	-	-	SPI6_ MISO	-	-	USART6_ RTS	LCD_B4	-	-	FMC_NE4	-	LCD_B1	EVEN TOUT	
	PG13	-	-	-	-	-	SPI6_ SCK	-	-	USART6_ CTS	-	-	-	ETH_MII_ TXD0/ ETH_RMII_ TXD0	FMC_A24	-	-	EVEN TOUT
	PG14	-	-	-	-	-	SPI6_ MOSI	-	-	USART6_ TX	-	-	-	ETH_MII_ TXD1/ ETH_RMII_ TXD1	FMC_A25	-	-	EVEN TOUT
	PG15	-	-	-	-	-	-	-	-	USART6_ CTS	-	-	-	-	FMC_ SDNCAS	DCMI_ D13	-	EVEN TOUT
Port H	PH0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT	
	PH1	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT	
	PH2	-	-	-	-	-	-	-	-	-	-	-	ETH_MII_ CRS	FMC_ SDCKE0	-	LCD_R0	EVEN TOUT	
	PH3	-	-	-	-	-	-	-	-	-	-	-	ETH_MII_ COL	FMC_SDN E0	-	LCD_R1	EVEN TOUT	
	PH4	-	-	-	-	I2C2_ SCL	-	-	-	-	-	OTG_HS_ ULPI_NXT	-	-	-	-	EVEN TOUT	
	PH5	-	-	-	-	I2C2_ SDA	SPI5_N SS	-	-	-	-	-	-	FMC_SDN WE	-	-	EVEN TOUT	
	PH6	-	-	-	-	I2C2_ SMBA	SPI5_ SCK	-	-	-	-	TIM12_CH1	-	-	FMC_ SDNE1	DCMI_ D8	-	-



**Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)**

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS
Port H	PH7	-	-	-	-	I2C3_ SCL	SPI5_ MISO	-	-	-	-	-	ETH_MII_ RXD3	FMC_ SDCKE1	DCMI_ D9	-	-
	PH8	-	-	-	-	I2C3_ SDA	-	-	-	-	-	-	-	FMC_D16	DCMI_ HSYNC	LCD_R2	EVEN TOUT
	PH9	-	-	-	-	I2C3_ SMBA	-	-	-	-	TIM12_CH2	-	-	FMC_D17	DCMI_ D0	LCD_R3	EVEN TOUT
	PH10	-	-	TIM5_ CH1	-	-	-	-	-	-	-	-	-	FMC_D18	DCMI_ D1	LCD_R4	EVEN TOUT
	PH11	-	-	TIM5_ CH2	-	-	-	-	-	-	-	-	-	FMC_D19	DCMI_ D2	LCD_R5	EVEN TOUT
	PH12	-	-	TIM5_ CH3	-	-	-	-	-	-	-	-	-	FMC_D20	DCMI_ D3	LCD_R6	EVEN TOUT
	PH13	-	-	-	TIM8_ CH1N	-	-	-	-	-	-	CAN1_TX	-	FMC_D21	-	LCD_G2	EVEN TOUT
	PH14	-	-	-	TIM8_ CH2N	-	-	-	-	-	-	-	-	FMC_D22	DCMI_ D4	LCD_G3	EVEN TOUT
	PH15	-	-	-	TIM8_ CH3N	-	-	-	-	-	-	-	-	FMC_D23	DCMI_ D11	LCD_G4	EVEN TOUT
Port I	PI0	-	-	TIM5_ CH4	-	-	SPI2_ NSS/I2 S2_WS	-	-	-	-	-	-	FMC_D24	DCMI_ D13	LCD_G5	EVEN TOUT
	PI1	-	-	-	-	-	SPI2_ SCK/I2 S2_CK	-	-	-	-	-	-	FMC_D25	DCMI_ D8	LCD_G6	EVEN TOUT
	PI2	-	-	-	TIM8_ CH4	-	SPI2_ MISO	I2S2ext_ SD	-	-	-	-	-	FMC_D26	DCMI_ D9	LCD_G7	EVEN TOUT
	PI3	-	-	-	TIM8_ ETR	-	SPI2_M OSI/I2S 2_SD	-	-	-	-	-	-	FMC_D27	DCMI_D 10	-	EVEN TOUT
	PI4	-	-	-	TIM8_ BKIN	-	-	-	-	-	-	-	-	FMC_ NBL2	DCMI_D 5	LCD_B4	EVEN TOUT
	PI5	-	-	-	TIM8_ CH1	-	-	-	-	-	-	-	-	FMC_ NBL3	DCMI_ VSYNC	LCD_B5	EVEN TOUT
	PI6	-	-	-	TIM8_ CH2	-	-	-	-	-	-	-	-	FMC_D28	DCMI_ D6	LCD_B6	EVEN TOUT



Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS
Port I	PI7	-	-	-	TIM8_ CH3	-	-	-	-	-	-	-	-	FMC_D29	DCMI_ D7	LCD_B7	EVEN TOUT
	PI8	-	-	-	-	-	-	-	-	-	-	-	-	-	-	-	EVEN TOUT
	PI9	-	-	-	-	-	-	-	-	-	CAN1_RX	-	-	FMC_D30	-	LCD_ VSYNC	EVEN TOUT
	PI10	-	-	-	-	-	-	-	-	-	-	-	ETH_MII_ RX_ER	FMC_D31	-	LCD_ HSYNC	EVEN TOUT
	PI11	-	-	-	-	-	-	-	-	-	-	OTG_HS_ ULPI_DIR	-	-	-	-	EVEN TOUT
	PI12	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_ HSYNC	EVEN TOUT
	PI13	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_ VSYNC	EVEN TOUT
	PI14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_ CLK	EVEN TOUT
	PI15	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_R0	EVEN TOUT
Port J	PJ0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_R1	EVEN TOUT
	PJ1	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_R2	EVEN TOUT
	PJ2	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_R3	EVEN TOUT
	PJ3	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_R4	EVEN TOUT
	PJ4	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_R5	EVEN TOUT
	PJ5	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_R6	EVEN TOUT
	PJ6	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_R7	EVEN TOUT
	PJ7	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G0	EVEN TOUT



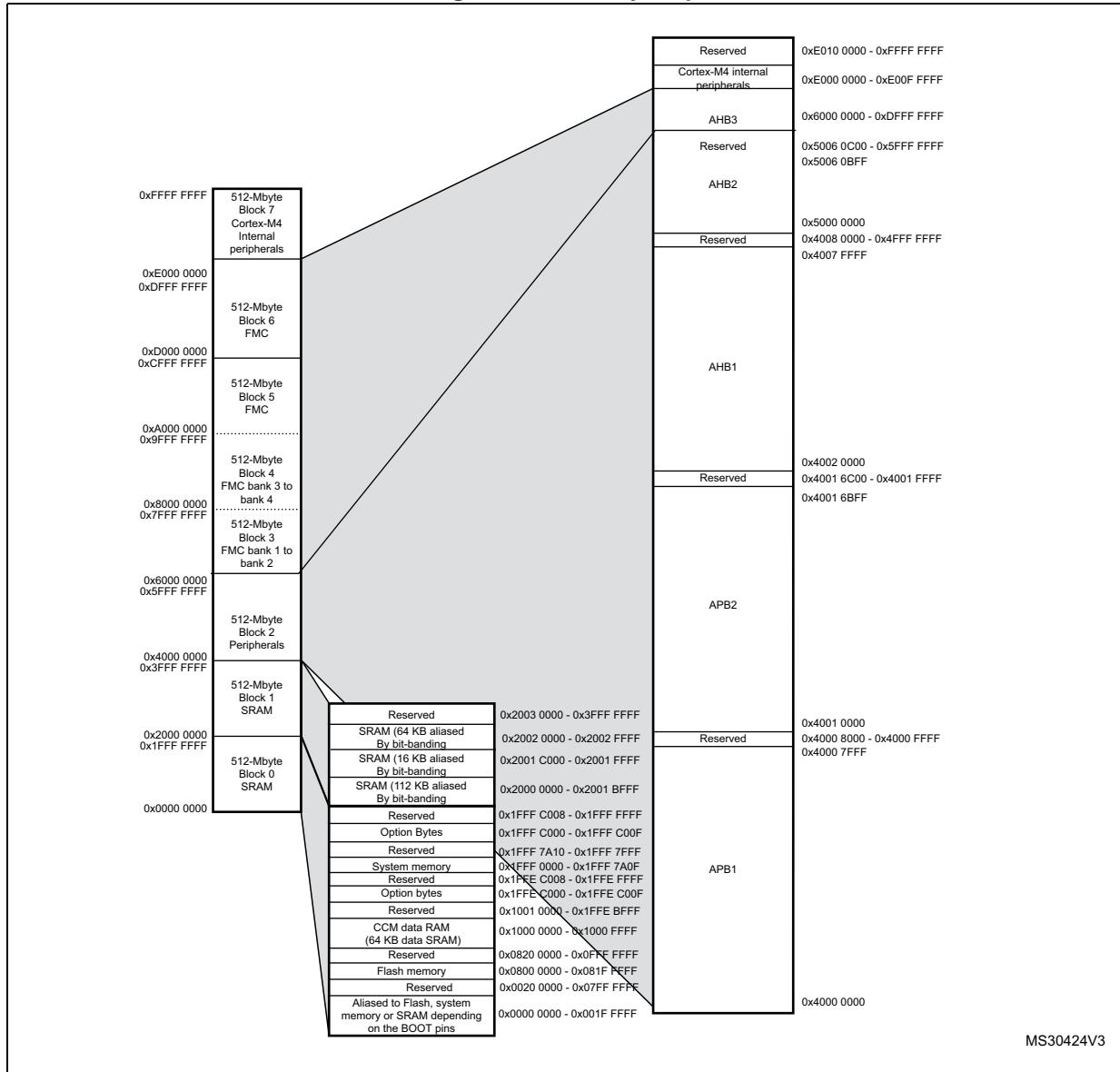
Table 12. STM32F437xx and STM32F439xx alternate function mapping (continued)

Port		AF0	AF1	AF2	AF3	AF4	AF5	AF6	AF7	AF8	AF9	AF10	AF11	AF12	AF13	AF14	AF15
		SYS	TIM1/2	TIM3/4/5	TIM8/9/ 10/11	I2C1/ 2/3	SPI1/2/ 3/4/5/6	SPI2/3/S AI1	SPI3/US ART1/2/3	USART6/U ART4/5/7/8	CAN1/2/TIM 12/13/14/ LCD	OTG2_HS /OTG1_ FS	ETH	FMC/SDIO /OTG2_FS	DCMI	LCD	SYS
Port J	PJ8	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G1	EVEN TOUT
	PJ9	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G2	EVEN TOUT
	PJ10	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G3	EVEN TOUT
	PJ11	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G4	EVEN TOUT
	PJ12	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B0	EVEN TOUT
	PJ13	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B1	EVEN TOUT
	PJ14	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B2	EVEN TOUT
	PJ15	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B3	EVEN TOUT
Port K	PK0	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G5	EVEN TOUT
	PK1	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G6	EVEN TOUT
	PK2	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_G7	EVEN TOUT
	PK3	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B4	EVEN TOUT
	PK4	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B5	EVEN TOUT
	PK5	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B6	EVEN TOUT
	PK6	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_B7	EVEN TOUT
	PK7	-	-	-	-	-	-	-	-	-	-	-	-	-	-	LCD_DE	EVEN TOUT

# 5 Memory mapping

The memory map is shown in *Figure 19*.

Figure 19. Memory map



MS30424V3

**Table 13. STM32F437xx and STM32F439xx register boundary addresses**

Bus	Boundary address	Peripheral
	0xE00F FFFF - 0xFFFF FFFF	Reserved
Cortex-M4	0xE000 0000 - 0xE00F FFFF	Cortex-M4 internal peripherals
AHB3	0xD000 0000 - 0xDFFF FFFF	FMC bank 6
	0xC000 0000 - 0xCFFF FFFF	FMC bank 5
	0xA000 1000 - 0xBFFF FFFF	Reserved
	0xA000 0000 - 0xA000 0FFF	FMC control register
	0x9000 0000 - 0x9FFF FFFF	FMC bank 4
	0x8000 0000 - 0x8FFF FFFF	FMC bank 3
	0x7000 0000 - 0x7FFF FFFF	FMC bank 2
	0x6000 0000 - 0x6FFF FFFF	FMC bank 1
	0x5006 0C00 - 0x5FFF FFFF	Reserved
AHB2	0x5006 0800 - 0x5006 0BFF	RNG
	0x5006 0400 - 0x5006 07FF	HASH
	0x5006 0000 - 0x5006 03FF	CRYP
	0x5005 0400 - 0x5006 07FF	Reserved
	0x5005 0000 - 0x5005 03FF	DCMI
	0x5004 0000 - 0x5004 FFFF	Reserved
	0x5000 0000 - 0x5003 FFFF	USB OTG FS

**Table 13. STM32F437xx and STM32F439xx register boundary addresses (continued)**

Bus	Boundary address	Peripheral
	0x4008 0000- 0x4FFF FFFF	Reserved
AHB1	0x4004 0000 - 0x4007 FFFF	USB OTG HS
	0x4002 BC00- 0x4003 FFFF	Reserved
	0x4002 B000 - 0x4002 BBFF	DMA2D
	0x4002 9400 - 0x4002 AFFF	Reserved
	0x4002 9000 - 0x4002 93FF	ETHERNET MAC
	0x4002 8C00 - 0x4002 8FFF	
	0x4002 8800 - 0x4002 8BFF	
	0x4002 8400 - 0x4002 87FF	
	0x4002 8000 - 0x4002 83FF	
	0x4002 6800 - 0x4002 7FFF	Reserved
	0x4002 6400 - 0x4002 67FF	DMA2
	0x4002 6000 - 0x4002 63FF	DMA1
	0x4002 5000 - 0x4002 5FFF	Reserved
	0x4002 4000 - 0x4002 4FFF	BKPSRAM
	0x4002 3C00 - 0x4002 3FFF	Flash interface register
	0x4002 3800 - 0x4002 3BFF	RCC
	0x4002 3400 - 0x4002 37FF	Reserved
	0x4002 3000 - 0x4002 33FF	CRC
	0x4002 2C00 - 0x4002 2FFF	Reserved
	0x4002 2800 - 0x4002 2BFF	GPIOK
	0x4002 2400 - 0x4002 27FF	GPIOJ
	0x4002 2000 - 0x4002 23FF	GPIOI
	0x4002 1C00 - 0x4002 1FFF	GPIOH
	0x4002 1800 - 0x4002 1BFF	GPIOG
	0x4002 1400 - 0x4002 17FF	GPIOF
	0x4002 1000 - 0x4002 13FF	GPIOE
	0x4002 0C00 - 0x4002 0FFF	GPIOD
	0x4002 0800 - 0x4002 0BFF	GPIOC
	0x4002 0400 - 0x4002 07FF	GPIOB
	0x4002 0000 - 0x4002 03FF	GPIOA

**Table 13. STM32F437xx and STM32F439xx register boundary addresses (continued)**

Bus	Boundary address	Peripheral
	0x4001 6C00 - 0x4001 FFFF	Reserved
APB2	0x4001 6800 - 0x4001 6BFF	LCD-TFT
	0x4001 5C00 - 0x4001 67FF	Reserved
	0x4001 5800 - 0x4001 5BFF	SAI1
	0x4001 5400 - 0x4001 57FF	SPI6
	0x4001 5000 - 0x4001 53FF	SPI5
	0x4001 4C00 - 0x4001 4FFF	Reserved
	0x4001 4800 - 0x4001 4BFF	TIM11
	0x4001 4400 - 0x4001 47FF	TIM10
	0x4001 4000 - 0x4001 43FF	TIM9
	0x4001 3C00 - 0x4001 3FFF	EXTI
	0x4001 3800 - 0x4001 3BFF	SYSCFG
	0x4001 3400 - 0x4001 37FF	SPI4
	0x4001 3000 - 0x4001 33FF	SPI1
	0x4001 2C00 - 0x4001 2FFF	SDIO
	0x4001 2400 - 0x4001 2BFF	Reserved
	0x4001 2000 - 0x4001 23FF	ADC1 - ADC2 - ADC3
	0x4001 1800 - 0x4001 1FFF	Reserved
	0x4001 1400 - 0x4001 17FF	USART6
	0x4001 1000 - 0x4001 13FF	USART1
	0x4001 0800 - 0x4001 0FFF	Reserved
0x4001 0400 - 0x4001 07FF	TIM8	
0x4001 0000 - 0x4001 03FF	TIM1	

Table 13. STM32F437xx and STM32F439xx register boundary addresses (continued)

Bus	Boundary address	Peripheral
	0x4000 8000 - 0x4000 FFFF	Reserved
APB1	0x4000 7C00 - 0x4000 7FFF	UART8
	0x4000 7800 - 0x4000 7BFF	UART7
	0x4000 7400 - 0x4000 77FF	DAC
	0x4000 7000 - 0x4000 73FF	PWR
	0x4000 6C00 - 0x4000 6FFF	Reserved
	0x4000 6800 - 0x4000 6BFF	CAN2
	0x4000 6400 - 0x4000 67FF	CAN1
	0x4000 6000 - 0x4000 63FF	Reserved
	0x4000 5C00 - 0x4000 5FFF	I2C3
	0x4000 5800 - 0x4000 5BFF	I2C2
	0x4000 5400 - 0x4000 57FF	I2C1
	0x4000 5000 - 0x4000 53FF	UART5
	0x4000 4C00 - 0x4000 4FFF	UART4
	0x4000 4800 - 0x4000 4BFF	USART3
	0x4000 4400 - 0x4000 47FF	USART2
	0x4000 4000 - 0x4000 43FF	I2S3ext
	0x4000 3C00 - 0x4000 3FFF	SPI3 / I2S3
	0x4000 3800 - 0x4000 3BFF	SPI2 / I2S2
	0x4000 3400 - 0x4000 37FF	I2S2ext
	0x4000 3000 - 0x4000 33FF	IWDG
	0x4000 2C00 - 0x4000 2FFF	WWDG
	0x4000 2800 - 0x4000 2BFF	RTC & BKP Registers
	0x4000 2400 - 0x4000 27FF	Reserved
	0x4000 2000 - 0x4000 23FF	TIM14
	0x4000 1C00 - 0x4000 1FFF	TIM13
	0x4000 1800 - 0x4000 1BFF	TIM12
	0x4000 1400 - 0x4000 17FF	TIM7
	0x4000 1000 - 0x4000 13FF	TIM6
	0x4000 0C00 - 0x4000 0FFF	TIM5
	0x4000 0800 - 0x4000 0BFF	TIM4
0x4000 0400 - 0x4000 07FF	TIM3	
0x4000 0000 - 0x4000 03FF	TIM2	

## 6 Electrical characteristics

### 6.1 Parameter conditions

Unless otherwise specified, all voltages are referenced to  $V_{SS}$ .

#### 6.1.1 Minimum and maximum values

Unless otherwise specified the minimum and maximum values are guaranteed in the worst conditions of ambient temperature, supply voltage and frequencies by tests in production on 100% of the devices with an ambient temperature at  $T_A = 25\text{ °C}$  and  $T_A = T_{Amax}$  (given by the selected temperature range).

Data based on characterization results, design simulation and/or technology characteristics are indicated in the table footnotes and are not tested in production. Based on characterization, the minimum and maximum values refer to sample tests and represent the mean value plus or minus three times the standard deviation ( $\text{mean} \pm 3\sigma$ ).

#### 6.1.2 Typical values

Unless otherwise specified, typical data are based on  $T_A = 25\text{ °C}$ ,  $V_{DD} = 3.3\text{ V}$  (for the  $1.8\text{ V} \leq V_{DD} \leq 3.6\text{ V}$  voltage range). They are given only as design guidelines and are not tested.

Typical ADC accuracy values are determined by characterization of a batch of samples from a standard diffusion lot over the full temperature range, where 95% of the devices have an error less than or equal to the value indicated ( $\text{mean} \pm 2\sigma$ ).

#### 6.1.3 Typical curves

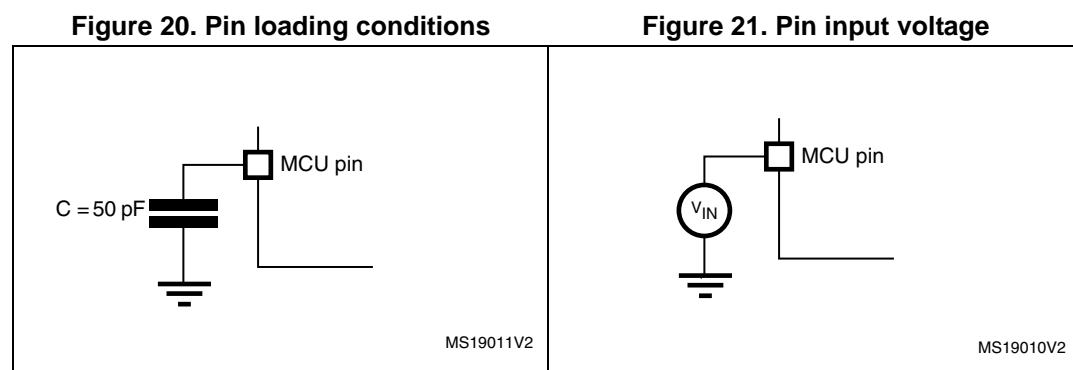
Unless otherwise specified, all typical curves are given only as design guidelines and are not tested.

#### 6.1.4 Loading capacitor

The loading conditions used for pin parameter measurement are shown in [Figure 20](#).

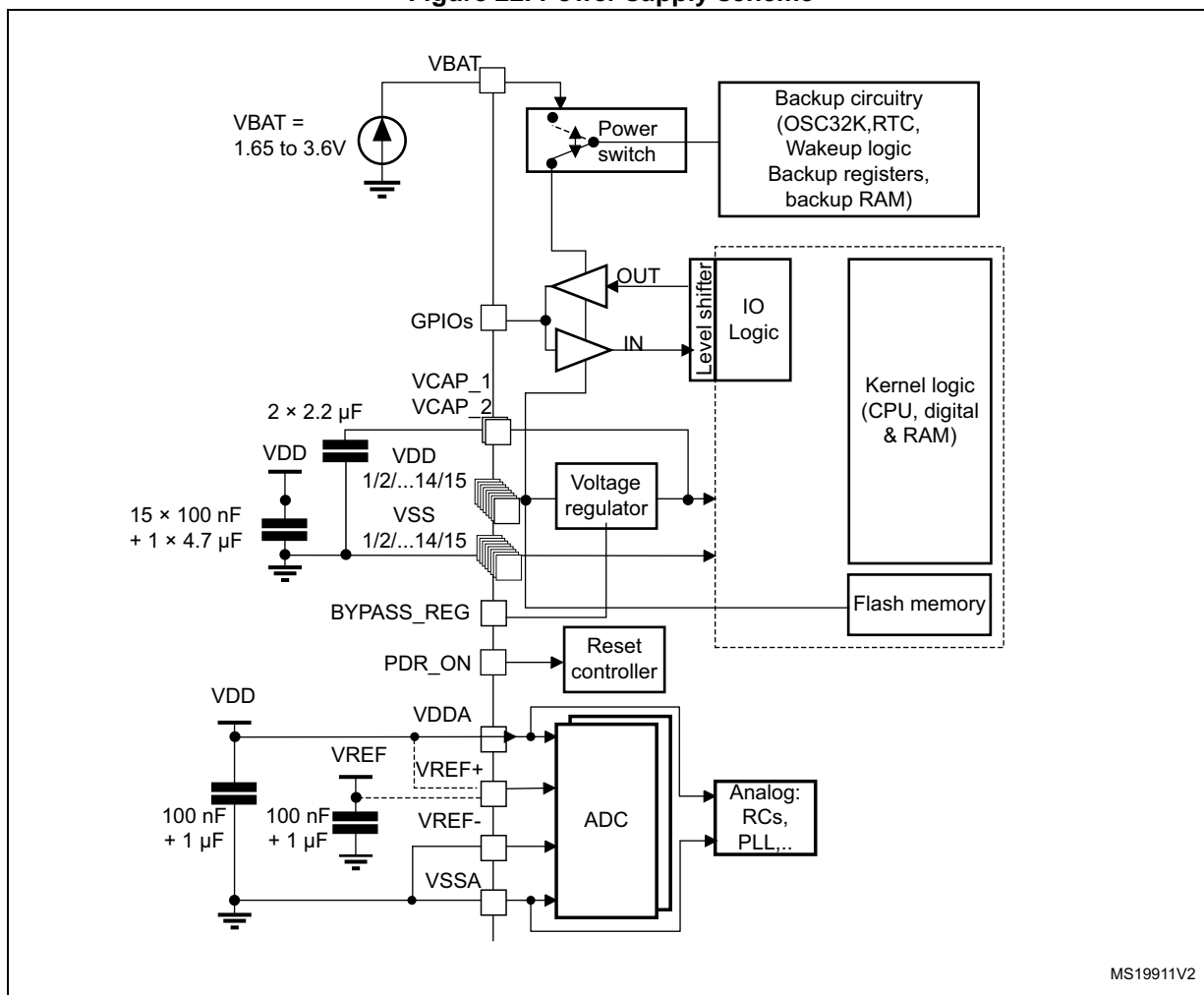
#### 6.1.5 Pin input voltage

The input voltage measurement on a pin of the device is described in [Figure 21](#).



### 6.1.6 Power supply scheme

Figure 22. Power supply scheme

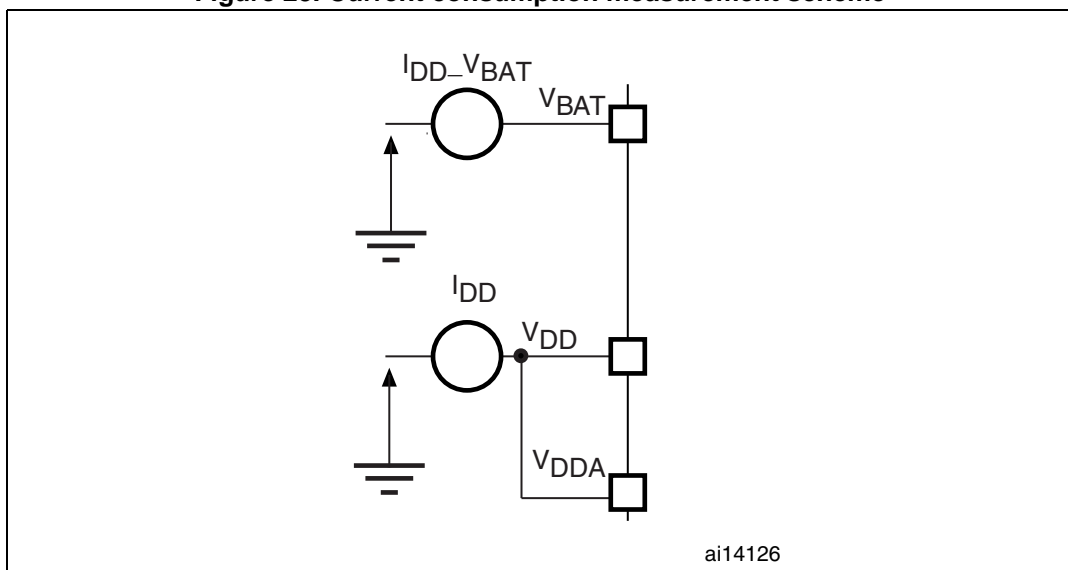


1. To connect BYPASS\_REG and PDR\_ON pins, refer to [Section 3.17: Power supply supervisor](#) and [Section 3.18: Voltage regulator](#)
2. The two 2.2 μF ceramic capacitors should be replaced by two 100 nF decoupling capacitors when the voltage regulator is OFF.
3. The 4.7 μF ceramic capacitor must be connected to one of the V<sub>DD</sub> pins.
4. V<sub>DDA</sub>=V<sub>DD</sub> and V<sub>SSA</sub>=V<sub>SS</sub>.

**Caution:** Each power supply pair (V<sub>DD</sub>/V<sub>SS</sub>, V<sub>DDA</sub>/V<sub>SSA</sub> ...) must be decoupled with filtering ceramic capacitors as shown above. These capacitors must be placed as close as possible to, or below, the appropriate pins on the underside of the PCB to ensure good operation of the device. It is not recommended to remove filtering capacitors to reduce PCB size or cost. This might cause incorrect operation of the device.

### 6.1.7 Current consumption measurement

Figure 23. Current consumption measurement scheme



## 6.2 Absolute maximum ratings

Stresses above the absolute maximum ratings listed in [Table 14: Voltage characteristics](#), [Table 15: Current characteristics](#), and [Table 16: Thermal characteristics](#) may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these conditions is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

Table 14. Voltage characteristics

Symbol	Ratings	Min	Max	Unit
$V_{DD-VSS}$	External main supply voltage (including $V_{DDA}$ , $V_{DD}$ and $V_{BAT}$ ) <sup>(1)</sup>	-0.3	4.0	V
$V_{IN}$	Input voltage on FT pins <sup>(2)</sup>	$V_{SS}-0.3$	$V_{DD}+4.0$	
	Input voltage on TTa pins	$V_{SS}-0.3$	4.0	
	Input voltage on any other pin	$V_{SS}-0.3$	4.0	
	Input voltage on BOOT0 pin	$V_{SS}$	9.0	
$ \Delta V_{DDx} $	Variations between different $V_{DD}$ power pins	-	50	mV
$ V_{SSx}-V_{SS} $	Variations between all the different ground pins	-	50	
$V_{ESD(HBM)}$	Electrostatic discharge voltage (human body model)	see <a href="#">Section 6.3.15: Absolute maximum ratings (electrical sensitivity)</a>		

1. All main power ( $V_{DD}$ ,  $V_{DDA}$ ) and ground ( $V_{SS}$ ,  $V_{SSA}$ ) pins must always be connected to the external power supply, in the permitted range.
2.  $V_{IN}$  maximum value must always be respected. Refer to [Table 15](#) for the values of the maximum allowed injected current.

**Table 15. Current characteristics**

Symbol	Ratings	Max.	Unit
$\Sigma I_{VDD}$	Total current into sum of all $V_{DD\_x}$ power lines (source) <sup>(1)</sup>	270	mA
$\Sigma I_{VSS}$	Total current out of sum of all $V_{SS\_x}$ ground lines (sink) <sup>(1)</sup>	-270	
$I_{VDD}$	Maximum current into each $V_{DD\_x}$ power line (source) <sup>(1)</sup>	100	
$I_{VSS}$	Maximum current out of each $V_{SS\_x}$ ground line (sink) <sup>(1)</sup>	-100	
$I_{IO}$	Output current sunk by any I/O and control pin	25	
	Output current sourced by any I/Os and control pin	-25	
$\Sigma I_{IO}$	Total output current sunk by sum of all I/O and control pins <sup>(2)</sup>	120	
	Total output current sourced by sum of all I/Os and control pins <sup>(2)</sup>	-120	
$I_{INJ(PIN)}^{(3)}$	Injected current on FT pins <sup>(4)</sup>	-5/+0	
	Injected current on NRST and BOOT0 pins <sup>(4)</sup>		
	Injected current on TTa pins <sup>(5)</sup>	±5	
$\Sigma I_{INJ(PIN)}^{(5)}$	Total injected current (sum of all I/O and control pins) <sup>(6)</sup>	±25	

1. All main power ( $V_{DD}$ ,  $V_{DDA}$ ) and ground ( $V_{SS}$ ,  $V_{SSA}$ ) pins must always be connected to the external power supply, in the permitted range.
2. This current consumption must be correctly distributed over all I/Os and control pins. The total output current must not be sunk/sourced between two consecutive power supply pins referring to high pin count LQFP packages.
3. Negative injection disturbs the analog performance of the device. See note in [Section 6.3.21: 12-bit ADC characteristics](#).
4. Positive injection is not possible on these I/Os and does not occur for input voltages lower than the specified maximum value.
5. A positive injection is induced by  $V_{IN} > V_{DDA}$  while a negative injection is induced by  $V_{IN} < V_{SS}$ .  $I_{INJ(PIN)}$  must never be exceeded. Refer to [Table 14](#) for the values of the maximum allowed input voltage.
6. When several inputs are submitted to a current injection, the maximum  $\Sigma I_{INJ(PIN)}$  is the absolute sum of the positive and negative injected currents (instantaneous values).

**Table 16. Thermal characteristics**

Symbol	Ratings	Value	Unit
$T_{STG}$	Storage temperature range	-65 to +150	°C
$T_J$	Maximum junction temperature	125	°C

### 6.3 Operating conditions

#### 6.3.1 General operating conditions

Table 17. General operating conditions

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Typ	Max	Unit	
f <sub>HCLK</sub>	Internal AHB clock frequency	Power Scale 3 (VOS[1:0] bits in PWR_CR register = 0x01), Regulator ON, over-drive OFF	0	-	120	MHz	
		Power Scale 2 (VOS[1:0] bits in PWR_CR register = 0x10), Regulator ON	Over-drive OFF	0	-		144
			Over-drive ON		-		168
		Power Scale 1 (VOS[1:0] bits in PWR_CR register= 0x11), Regulator ON	Over-drive OFF	0	-		168
			Over-drive ON		-		180
		f <sub>PCLK1</sub>	Internal APB1 clock frequency	Over-drive OFF	0		-
Over-drive ON	0			-	45		
f <sub>PCLK2</sub>	Internal APB2 clock frequency	Over-drive OFF	0	-	84		
		Over-drive ON	0	-	90		

Table 17. General operating conditions (continued)

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Typ	Max	Unit
V <sub>DD</sub>	Standard operating voltage		1.7 <sup>(2)</sup>	-	3.6	V
V <sub>DDA</sub> <sup>(3)</sup> (4)	Analog operating voltage (ADC limited to 1.2 M samples)	Must be the same potential as V <sub>DD</sub> <sup>(5)</sup>	1.8 <sup>(2)</sup>	-	2.4	
	Analog operating voltage (ADC limited to 2.4 M samples)		2.4	-	3.6	
V <sub>BAT</sub>	Backup operating voltage		1.65	-	3.6	
V <sub>12</sub>	Regulator ON: 1.2 V internal voltage on V <sub>CAP_1</sub> /V <sub>CAP_2</sub> pins	Power Scale 3 ((VOS[1:0] bits in PWR_CR register = 0x01), 120 MHz HCLK max frequency	1.08	1.14	1.20	
		Power Scale 2 ((VOS[1:0] bits in PWR_CR register = 0x10), 144 MHz HCLK max frequency with over-drive OFF or 168 MHz with over-drive ON	1.20	1.26	1.32	
		Power Scale 1 ((VOS[1:0] bits in PWR_CR register = 0x11), 168 MHz HCLK max frequency with over-drive OFF or 180 MHz with over-drive ON	1.26	1.32	1.40	
	Regulator OFF: 1.2 V external voltage must be supplied from external regulator on V <sub>CAP_1</sub> /V <sub>CAP_2</sub> pins <sup>(6)</sup>	Max frequency 120 MHz	1.10	1.14	1.20	
		Max frequency 144 MHz	1.20	1.26	1.32	
		Max frequency 168 MHz	1.26	1.32	1.38	
	V <sub>IN</sub>	Input voltage on RST and FT pins <sup>(7)</sup>	2 V ≤ V <sub>DD</sub> ≤ 3.6 V	-0.3	-	5.5
V <sub>DD</sub> ≤ 2 V			-0.3	-	5.2	
Input voltage on TTA pins		-0.3	-	V <sub>DDA</sub> + 0.3		
	Input voltage on BOOT0 pin	0	-	9		
P <sub>D</sub>	Power dissipation at T <sub>A</sub> = 85 °C for suffix 6 or T <sub>A</sub> = 105 °C for suffix 7 <sup>(8)</sup>	LQFP100	-	-	465	
		WLCSP143	-	-	641	
		LQFP144	-	-	500	
		UFBGA169	-	-	385	
		LQFP176	-	-	526	
		UFBGA176	-	-	513	
		LQFP208	-	-	1053	
TFBGA216	-	-	690			
T <sub>A</sub>	Ambient temperature for 6 suffix version	Maximum power dissipation	-40		85	
		Low power dissipation <sup>(9)</sup>	-40		105	
	Ambient temperature for 7 suffix version	Maximum power dissipation	-40		105	
		Low power dissipation <sup>(9)</sup>	-40		125	

**Table 17. General operating conditions (continued)**

Symbol	Parameter	Conditions <sup>(1)</sup>	Min	Typ	Max	Unit
T <sub>J</sub>	Junction temperature range	6 suffix version	-40		105	°C
		7 suffix version	-40		125	

1. The over-drive mode is not supported at the voltage ranges from 1.8 to 2.1 V.
2. V<sub>DD</sub>/V<sub>DDA</sub> minimum value of 1.7 V is obtained with the use of an external power supply supervisor (refer to [Section 3.17.2: Internal reset OFF](#)).
3. When the ADC is used, refer to [Table 76: ADC characteristics](#).
4. If V<sub>REF+</sub> pin is present, it must respect the following condition: V<sub>DDA</sub>-V<sub>REF+</sub> < 1.2 V.
5. It is recommended to power V<sub>DD</sub> and V<sub>DDA</sub> from the same source. A maximum difference of 300 mV between V<sub>DD</sub> and V<sub>DDA</sub> can be tolerated during power-up and power-down operation.
6. The over-drive mode is not supported when the internal regulator is OFF.
7. To sustain a voltage higher than VDD+0.3, the internal Pull-up and Pull-Down resistors must be disabled
8. If T<sub>A</sub> is lower, higher P<sub>D</sub> values are allowed as long as T<sub>J</sub> does not exceed T<sub>Jmax</sub>.
9. In low power dissipation state, T<sub>A</sub> can be extended to this range as long as T<sub>J</sub> does not exceed T<sub>Jmax</sub>.

**Table 18. Limitations depending on the operating power supply range**

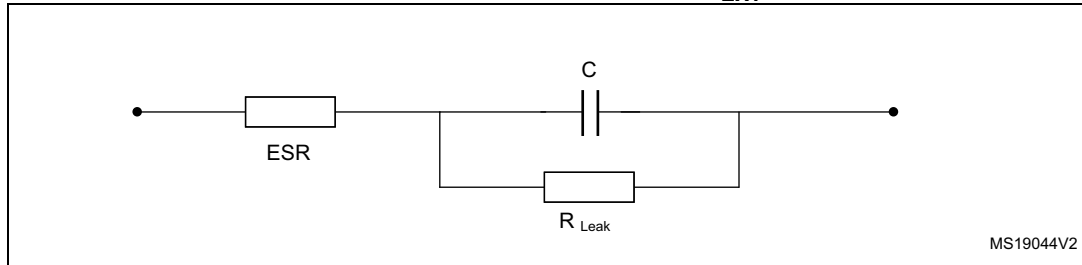
Operating power supply range	ADC operation	Maximum Flash memory access frequency with no wait states (f <sub>Flashmax</sub> )	Maximum HCLK frequency vs Flash memory wait states (1)(2)	I/O operation	Possible Flash memory operations
V <sub>DD</sub> = 1.8 to 2.1 V <sup>(3)</sup>	Conversion time up to 1.2 Msps	20 MHz <sup>(4)</sup>	168 MHz with 8 wait states and over-drive OFF	- No I/O compensation	8-bit erase and program operations only
V <sub>DD</sub> = 2.1 to 2.4 V	Conversion time up to 1.2 Msps	22 MHz	180 MHz with 8 wait states and over-drive ON	- No I/O compensation	16-bit erase and program operations
V <sub>DD</sub> = 2.4 to 2.7 V	Conversion time up to 2.4 Msps	24 MHz	180 MHz with 7 wait states and over-drive ON	- I/O compensation works	16-bit erase and program operations
V <sub>DD</sub> = 2.7 to 3.6 V <sup>(5)</sup>	Conversion time up to 2.4 Msps	30 MHz	180 MHz with 5 wait states and over-drive ON	- I/O compensation works	32-bit erase and program operations

1. Applicable only when the code is executed from Flash memory. When the code is executed from RAM, no wait state is required.
2. Thanks to the ART accelerator and the 128-bit Flash memory, the number of wait states given here does not impact the execution speed from Flash memory since the ART accelerator allows to achieve a performance equivalent to 0 wait state program execution.
3. V<sub>DD</sub>/V<sub>DDA</sub> minimum value of 1.7 V is obtained with the use of an external power supply supervisor (refer to [Section 3.17.2: Internal reset OFF](#)).
4. Prefetch is not available.
5. The voltage range for USB full speed PHYs can drop down to 2.7 V. However the electrical characteristics of D- and D+ pins will be degraded between 2.7 and 3 V.

### 6.3.2 VCAP1/VCAP2 external capacitor

Stabilization for the main regulator is achieved by connecting an external capacitor  $C_{EXT}$  to the VCAP1/VCAP2 pins.  $C_{EXT}$  is specified in [Table 19](#).

**Figure 24. External capacitor  $C_{EXT}$**



- Legend: ESR is the equivalent series resistance.

**Table 19. VCAP1/VCAP2 operating conditions<sup>(1)</sup>**

Symbol	Parameter	Conditions
C <sub>EXT</sub>	Capacitance of external capacitor	2.2 μF
ESR	ESR of external capacitor	< 2 Ω

- When bypassing the voltage regulator, the two 2.2 μF  $V_{CAP}$  capacitors are not required and should be replaced by two 100 nF decoupling capacitors.

### 6.3.3 Operating conditions at power-up / power-down (regulator ON)

Subject to general operating conditions for  $T_A$ .

**Table 20. Operating conditions at power-up / power-down (regulator ON)**

Symbol	Parameter	Min	Max	Unit
$t_{VDD}$	$V_{DD}$ rise time rate	20	∞	μs/V
	$V_{DD}$ fall time rate	20	∞	

### 6.3.4 Operating conditions at power-up / power-down (regulator OFF)

Subject to general operating conditions for  $T_A$ .

**Table 21. Operating conditions at power-up / power-down (regulator OFF)<sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Max	Unit
$t_{VDD}$	$V_{DD}$ rise time rate	Power-up	20	∞	μs/V
	$V_{DD}$ fall time rate	Power-down	20	∞	
$t_{VCAP}$	$V_{CAP\_1}$ and $V_{CAP\_2}$ rise time rate	Power-up	20	∞	
	$V_{CAP\_1}$ and $V_{CAP\_2}$ fall time rate	Power-down	20	∞	

- To reset the internal logic at power-down, a reset must be applied on pin PA0 when  $V_{DD}$  reach below 1.08 V.

6.3.5 reset and power control block characteristics

The parameters given in [Table 22](#) are derived from tests performed under ambient temperature and V<sub>DD</sub> supply voltage conditions summarized in [Table 17](#).

Table 22. reset and power control block characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V <sub>PVD</sub>	Programmable voltage detector level selection	PLS[2:0]=000 (rising edge)	2.09	2.14	2.19	V
		PLS[2:0]=000 (falling edge)	1.98	2.04	2.08	V
		PLS[2:0]=001 (rising edge)	2.23	2.30	2.37	V
		PLS[2:0]=001 (falling edge)	2.13	2.19	2.25	V
		PLS[2:0]=010 (rising edge)	2.39	2.45	2.51	V
		PLS[2:0]=010 (falling edge)	2.29	2.35	2.39	V
		PLS[2:0]=011 (rising edge)	2.54	2.60	2.65	V
		PLS[2:0]=011 (falling edge)	2.44	2.51	2.56	V
		PLS[2:0]=100 (rising edge)	2.70	2.76	2.82	V
		PLS[2:0]=100 (falling edge)	2.59	2.66	2.71	V
		PLS[2:0]=101 (rising edge)	2.86	2.93	2.99	V
		PLS[2:0]=101 (falling edge)	2.65	2.84	3.02	V
		PLS[2:0]=110 (rising edge)	2.96	3.03	3.10	V
		PLS[2:0]=110 (falling edge)	2.85	2.93	2.99	V
		PLS[2:0]=111 (rising edge)	3.07	3.14	3.21	V
PLS[2:0]=111 (falling edge)	2.95	3.03	3.09	V		
V <sub>PVDhyst</sub> <sup>(1)</sup>	PVD hysteresis		-	100	-	mV
V <sub>POR/PDR</sub>	Power-on/power-down reset threshold	Falling edge	1.60	1.68	1.76	V
		Rising edge	1.64	1.72	1.80	V
V <sub>PDRhyst</sub> <sup>(1)</sup>	PDR hysteresis		-	40	-	mV
V <sub>BOR1</sub>	Brownout level 1 threshold	Falling edge	2.13	2.19	2.24	V
		Rising edge	2.23	2.29	2.33	V
V <sub>BOR2</sub>	Brownout level 2 threshold	Falling edge	2.44	2.50	2.56	V
		Rising edge	2.53	2.59	2.63	V
V <sub>BOR3</sub>	Brownout level 3 threshold	Falling edge	2.75	2.83	2.88	V
		Rising edge	2.85	2.92	2.97	V
V <sub>BORhyst</sub> <sup>(1)</sup>	BOR hysteresis		-	100	-	mV
T <sub>RSTTEMPO</sub> <sup>(1)(2)</sup>	POR reset temporization		0.5	1.5	3.0	ms

**Table 22. reset and power control block characteristics (continued)**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{RUSH}^{(1)}$	InRush current on voltage regulator power-on (POR or wakeup from Standby)		-	160	200	mA
$E_{RUSH}^{(1)}$	InRush energy on voltage regulator power-on (POR or wakeup from Standby)	$V_{DD} = 1.8\text{ V}$ , $T_A = 105\text{ °C}$ , $I_{RUSH} = 171\text{ mA}$ for $31\text{ }\mu\text{s}$	-	-	5.4	$\mu\text{C}$

1. Guaranteed by design, not tested in production.
2. The reset temporization is measured from the power-on (POR reset or wakeup from  $V_{BAT}$ ) to the instant when first instruction is read by the user application code.

### 6.3.6 Over-drive switching characteristics

When the over-drive mode switches from enabled to disabled or disabled to enabled, the system clock is stalled during the internal voltage set-up.

The over-drive switching characteristics are given in [Table 23](#). They are subject to general operating conditions for  $T_A$ .

**Table 23. Over-drive switching characteristics<sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
Tod_swen	Over_drive switch enable time	HSI	-	45	-	$\mu\text{s}$
		HSE max for 4 MHz and min for 26 MHz	45	-	100	
		External HSE 50 MHz	-	40	-	
Tod_swdis	Over_drive switch disable time	HSI	-	20	-	
		HSE max for 4 MHz and min for 26 MHz.	20	-	80	
		External HSE 50 MHz	-	15	-	

1. Guaranteed by design, not tested in production.

### 6.3.7 Supply current characteristics

The current consumption is a function of several parameters and factors such as the operating voltage, ambient temperature, I/O pin loading, device software configuration, operating frequencies, I/O pin switching rate, program location in memory and executed binary code.

The current consumption is measured as described in [Figure 23: Current consumption measurement scheme](#).

All the run-mode current consumption measurements given in this section are performed with a reduced code that gives a consumption equivalent to CoreMark code.

### Typical and maximum current consumption

The MCU is placed under the following conditions:

- All I/O pins are in input mode with a static value at  $V_{DD}$  or  $V_{SS}$  (no load).
- All peripherals are disabled except if it is explicitly mentioned.
- The Flash memory access time is adjusted both to  $f_{HCLK}$  frequency and  $V_{DD}$  range (see [Table 18: Limitations depending on the operating power supply range](#)).
- Regulator ON
- The voltage scaling and over-drive mode are adjusted to  $f_{HCLK}$  frequency as follows:
  - Scale 3 for  $f_{HCLK} \leq 120$  MHz
  - Scale 2 for  $120 \text{ MHz} < f_{HCLK} \leq 144$  MHz
  - Scale 1 for  $144 \text{ MHz} < f_{HCLK} \leq 180$  MHz. The over-drive is only ON at 180 MHz.
- The system clock is HCLK,  $f_{PCLK1} = f_{HCLK}/4$ , and  $f_{PCLK2} = f_{HCLK}/2$ .
- External clock frequency is 4 MHz and PLL is ON when  $f_{HCLK}$  is higher than 25 MHz.
- The maximum values are obtained for  $V_{DD} = 3.6$  V and a maximum ambient temperature ( $T_A$ ), and the typical values for  $T_A = 25$  °C and  $V_{DD} = 3.3$  V unless otherwise specified.

**Table 24. Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator enabled except prefetch) or RAM<sup>(1)</sup>**

Symbol	Parameter	Conditions	f <sub>HCLK</sub> (MHz)	Typ	Max <sup>(2)</sup>			Unit
					T <sub>A</sub> = 25 °C	T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	
I <sub>DD</sub>	Supply current in RUN mode	All Peripherals enabled <sup>(3)(4)</sup>	180	98	104 <sup>(5)</sup>	123 <sup>(5)</sup>	141 <sup>(5)</sup>	mA
			168	89	98 <sup>(5)</sup>	116 <sup>(5)</sup>	133 <sup>(5)</sup>	
			150	75	84	100	115	
			144	72	81	96	112	
			120	54	58	72	85	
			90	43	45	56	66	
			60	29	30	38	45	
			30	16	20	34	46	
			25	13	16	30	43	
			16	11	13	27	39	
			8	5	9	23	36	
			4	4	8	21	34	
		2	2	7	20	33		
		All Peripherals disabled <sup>(3)</sup>	180	44	47 <sup>(5)</sup>	69 <sup>(5)</sup>	87 <sup>(5)</sup>	
			168	41	45 <sup>(5)</sup>	66 <sup>(5)</sup>	83 <sup>(5)</sup>	
			150	36	39	57	73	
			144	33	37	56	72	
			120	25	29	43	56	
			90	20	21	32	41	
			60	14	15	22	28	
			30	8	8	12	26	
			25	7	7	10	24	
			16	7	6.5	9	22	
			8	3	3.4	7	21	
4	3		2.7	6	20			
2	2	2.4	6	20				

- Code and data processing running from SRAM1 using boot pins.
- Based on characterization, not tested in production.
- When analog peripheral blocks such as ADCs, DACs, HSE, LSE, HSI, or LSI are ON, an additional power consumption should be considered.
- When the ADC is ON (ADON bit set in the ADC\_CR2 register), add an additional power consumption of 1.6 mA per ADC for the analog part.
- Based on characterization, tested in production.



**Table 25. Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator disabled)**

Symbol	Parameter	Conditions	f <sub>HCLK</sub> (MHz)	Typ	Max <sup>(1)</sup>			Unit
					TA=25 °C	TA=85 °C	TA=105 °C	
I <sub>DD</sub>	Supply current in RUN mode	All Peripherals enabled <sup>(2)(3)</sup>	180	103	112	140	151	mA
			168	98	107	126	144	
			150	87	95	112	128	
			144	85	92	108	124	
			120	66	71	85	99	
			90	54	58	69	80	
			60	37	39	47	55	
			30	20	24	39	51	
			25	17	21	35	48	
			16	12	16	30	42	
			8	7	11	24	37	
			4	5	8	22	35	
		2	3	7	21	34		
		All Peripherals disabled <sup>(3)</sup>	180	57	62	87	106	
			168	50	54	76	93	
			150	46	50	70	86	
			144	45	49	68	84	
			120	36	41	56	69	
			90	29	34	46	57	
			60	21	24	33	41	
			30	13	17	31	44	
			25	11	15	28	41	
			16	8	12	25	38	
			8	5	9	23	35	
4	4		7	21	34			
2	3	6.5	20	33				

1. Based on characterization, not tested in production unless otherwise specified.
2. When analog peripheral blocks such as ADCs, DACs, HSE, LSE, HSI, or LSI are ON, an additional power consumption should be considered.
3. When the ADC is ON (ADON bit set in the ADC\_CR2 register), add an additional power consumption of 1.6 mA per ADC for the analog part.

Table 26. Typical and maximum current consumption in Sleep mode

Symbol	Parameter	Conditions	f <sub>HCLK</sub> (MHz)	Typ	Max <sup>(1)</sup>			Unit
					T <sub>A</sub> = 25 °C	T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	
I <sub>DD</sub>	Supply current in Sleep mode	All Peripherals enabled <sup>(2)</sup>	180	78	89 <sup>(3)</sup>	110 <sup>(3)</sup>	130 <sup>(3)</sup>	mA
			168	66	75 <sup>(3)</sup>	93 <sup>(3)</sup>	110 <sup>(3)</sup>	
			150	56	61	80	96	
			144	54	58	78	94	
			120	40	44	59	72	
			90	32	34	46	56	
			60	22	23	31	38	
			30	10	16	30	43	
			25	9	14	28	40	
			16	5	12	25	40	
			8	3	8	22	35	
			4	3	7	21	34	
		2	2	6.5	20	33		
		All Peripherals disabled	180	21	26 <sup>(3)</sup>	54 <sup>(3)</sup>	76 <sup>(3)</sup>	
			168	16	20 <sup>(3)</sup>	41 <sup>(3)</sup>	58 <sup>(3)</sup>	
			150	14	17	36	52	
			144	13	16.5	35	51	
			120	10	14	28	41	
			90	8	13	26	37	
			60	6	9	17	25	
			30	5	8	22	35	
			25	3	7	21	34	
			16	3	7	21	34	
			8	2	6	20	33	
4	2		6	20	33			
2	2	6	20	33				

1. Based on characterization, not tested in production unless otherwise specified.
2. When analog peripheral blocks such as ADCs, DACs, HSE, LSE, HSI, or LSI are ON, an additional power consumption should be considered.
3. Based on characterization, tested in production.

**Table 27. Typical and maximum current consumptions in Stop mode**

Symbol	Parameter	Conditions	Max <sup>(1)</sup>				Unit
			V <sub>DD</sub> = 3.6 V				
			Typ	T <sub>A</sub> = 25 °C	T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	
I <sub>DD_STOP_NM</sub> (normal mode)	Supply current in Stop mode with voltage regulator in main regulator mode	Flash memory in Stop mode, all oscillators OFF, no independent watchdog	0.40	1.50	14.00	25.00	mA
		Flash memory in Deep power down mode, all oscillators OFF, no independent watchdog	0.35	1.50	14.00	25.00	
	Supply current in Stop mode with voltage regulator in Low Power regulator mode	Flash memory in Stop mode, all oscillators OFF, no independent watchdog	0.29	1.10	10.00	18.00	
		Flash memory in Deep power down mode, all oscillators OFF, no independent watchdog	0.23	1.10	10.00	18.00	
I <sub>DD_STOP_UDM</sub> (under-drive mode)	Supply current in Stop mode with voltage regulator in main regulator and under-drive mode	Flash memory in Deep power down mode, main regulator in under-drive mode, all oscillators OFF, no independent watchdog	0.19	0.50	6.00	9.00	mA
	Supply current in Stop mode with voltage regulator in Low Power regulator and under-drive mode	Flash memory in Deep power down mode, Low Power regulator in under-drive mode, all oscillators OFF, no independent watchdog	0.12	0.40	4.00	7.00	

1. Data based on characterization, tested in production.

**Table 28. Typical and maximum current consumptions in Standby mode**

Symbol	Parameter	Conditions	Typ <sup>(1)</sup>			Max <sup>(2)</sup>			Unit
			T <sub>A</sub> = 25 °C			T <sub>A</sub> = 25 °C	T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	
			V <sub>DD</sub> = 1.8 V	V <sub>DD</sub> = 2.4 V	V <sub>DD</sub> = 3.3 V	V <sub>DD</sub> = 3.6 V			
I <sub>DD_STBY</sub>	Supply current in Standby mode	Backup SRAM ON, low-speed oscillator (LSE) and RTC ON	2.80	3.00	3.60	7.00	19.00	36.00	µA
		Backup SRAM OFF, low-speed oscillator (LSE) and RTC ON	2.30	2.60	3.10	6.00	16.00	31.00	
		Backup SRAM ON, RTC and LSE OFF	2.30	2.50	2.90	6.00 <sup>(3)</sup>	18.00 <sup>(3)</sup>	35.00 <sup>(3)</sup>	
		Backup SRAM OFF, RTC and LSE OFF	1.70	1.90	2.20	5.00 <sup>(3)</sup>	15.00 <sup>(3)</sup>	30.00 <sup>(3)</sup>	

1. When the PDR is OFF (internal reset is OFF), the typical current consumption is reduced by 1.2 µA.



- 2. Based on characterization, not tested in production unless otherwise specified.
- 3. Based on characterization, tested in production.

**Table 29. Typical and maximum current consumptions in V<sub>BAT</sub> mode**

Symbol	Parameter	Conditions <sup>(1)</sup>	Typ			Max <sup>(2)</sup>		Unit
			T <sub>A</sub> = 25 °C			T <sub>A</sub> = 85 °C	T <sub>A</sub> = 105 °C	
			V <sub>BAT</sub> = 1.8 V	V <sub>BAT</sub> = 2.4 V	V <sub>BAT</sub> = 3.3 V	V <sub>BAT</sub> = 3.6 V		
I <sub>DD_VBAT</sub>	Backup domain supply current	Backup SRAM ON, low-speed oscillator (LSE) and RTC ON	1.28	1.40	1.62	6	11	µA
		Backup SRAM OFF, low-speed oscillator (LSE) and RTC ON	0.66	0.76	0.97	3	5	
		Backup SRAM ON, RTC and LSE OFF	0.70	0.72	0.74	5	10	
		Backup SRAM OFF, RTC and LSE OFF	0.10	0.10	0.10	2	4	

- 1. Crystal used: Abracon ABS07-120-32.768 kHz-T with a C<sub>L</sub> of 6 pF for typical values.
- 2. Based on characterization, not tested in production.

**Figure 25. Typical V<sub>BAT</sub> current consumption (LSE and RTC ON/backup RAM OFF)**

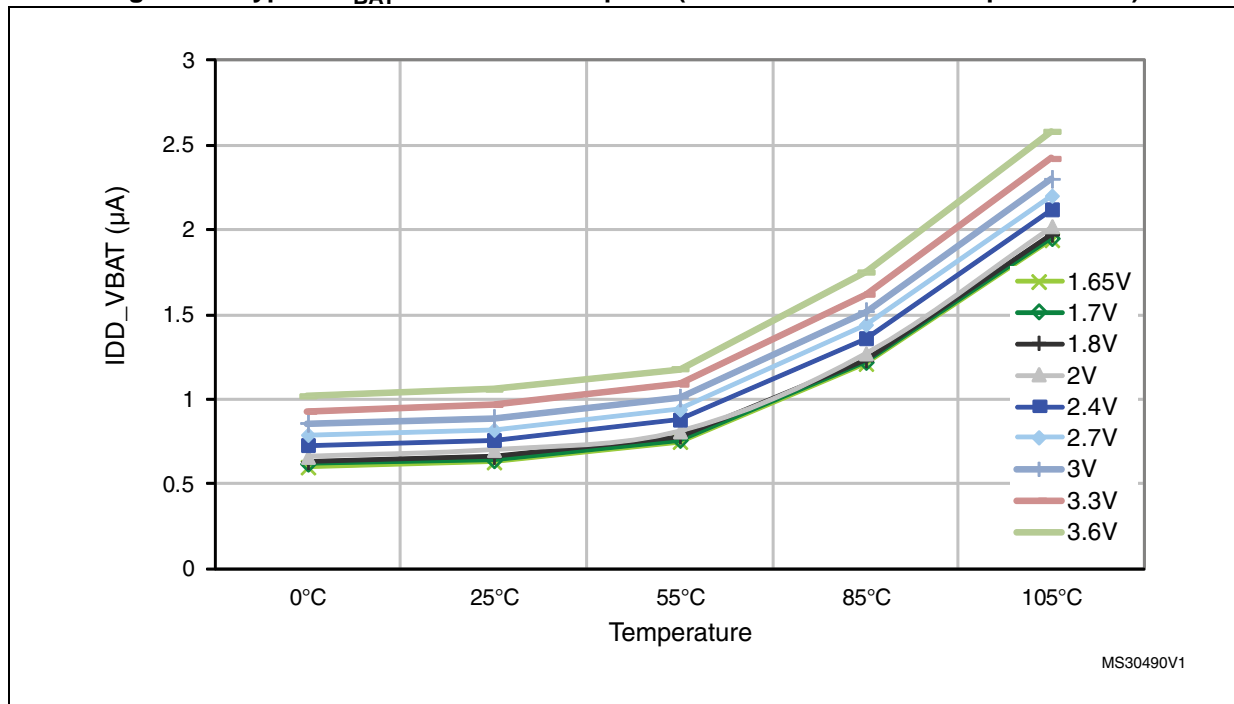
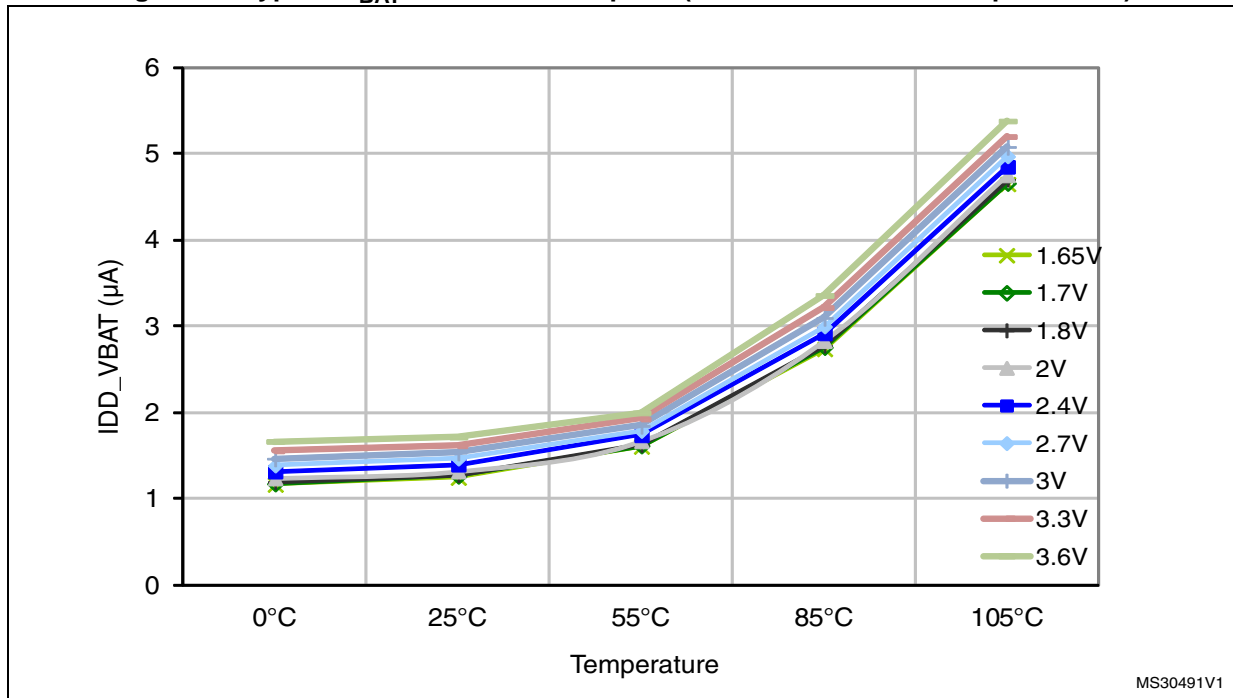


Figure 26. Typical  $V_{BAT}$  current consumption (LSE and RTC ON/backup RAM ON)



### Additional current consumption

The MCU is placed under the following conditions:

- All I/O pins are configured in analog mode.
- The Flash memory access time is adjusted to  $f_{HCLK}$  frequency.
- The voltage scaling is adjusted to  $f_{HCLK}$  frequency as follows:
  - Scale 3 for  $f_{HCLK} \leq 120$  MHz,
  - Scale 2 for  $120 \text{ MHz} < f_{HCLK} \leq 144$  MHz
  - Scale 1 for  $144 \text{ MHz} < f_{HCLK} \leq 180$  MHz. The over-drive is only ON at 180 MHz.
- The system clock is HCLK,  $f_{PCLK1} = f_{HCLK}/4$ , and  $f_{PCLK2} = f_{HCLK}/2$ .
- HSE crystal clock frequency is 25 MHz.
- When the regulator is OFF, V12 is provided externally as described in [Table 17: General operating conditions](#)
- $T_A = 25^\circ\text{C}$ .

**Table 30. Typical current consumption in Run mode, code with data processing running from Flash memory or RAM, regulator ON (ART accelerator enabled except prefetch), V<sub>DD</sub>=1.8 V<sup>(1)</sup>**

Symbol	Parameter	Conditions	f <sub>HCLK</sub> (MHz)	Typ	Unit
I <sub>DD</sub>	Supply current in RUN mode from V <sub>DD</sub> supply	All Peripheral enabled	168	88.2	mA
			150	74.3	
			144	71.3	
			120	52.9	
			90	42.6	
			60	28.6	
			30	15.7	
		All Peripheral disabled	25	12.3	
			168	40.6	
			150	30.6	
			144	32.6	
			120	24.7	
			90	19.7	
			60	13.6	
30	7.7				
25	6.7				

1. When peripherals are enabled, the power consumption corresponding to the analog part of the peripherals (such as ADC, or DAC) is not included.

**Table 31. Typical current consumption in Run mode, code with data processing running from Flash memory, regulator OFF (ART accelerator enabled except prefetch)<sup>(1)</sup>**

Symbol	Parameter	Conditions	f <sub>HCLK</sub> (MHz)	VDD=3.3 V		VDD=1.8 V		Unit
				I <sub>DD12</sub>	I <sub>DD</sub>	I <sub>DD12</sub>	I <sub>DD</sub>	
I <sub>DD12</sub> / I <sub>DD</sub>	Supply current in RUN mode from V <sub>12</sub> and V <sub>DD</sub> supply	All Peripherals enabled	168	77.8	1.3	76.8	1.0	mA
			150	70.8	1.3	69.8	1.0	
			144	64.5	1.3	63.6	1.0	
			120	49.9	1.2	49.3	0.9	
			90	39.2	1.3	38.7	1.0	
			60	27.2	1.2	26.8	0.9	
			30	15.6	1.2	15.4	0.9	
			25	13.6	1.2	13.5	0.9	
		All Peripherals disabled	168	38.2	1.3	37.0	1.0	
			150	34.6	1.3	33.4	1.0	
			144	31.3	1.3	30.3	1.0	
			120	24.0	1.2	23.2	0.9	
			90	18.1	1.4	18.0	1.0	
			60	12.9	1.2	12.5	0.9	
			30	7.2	1.2	6.9	0.9	
			25	6.3	1.2	6.1	0.9	

1. When peripherals are enabled, the power consumption corresponding to the analog part of the peripherals (such as ADC, or DAC) is not included.

**Table 32. Typical current consumption in Sleep mode, regulator ON, V<sub>DD</sub>=1.8 V<sup>(1)</sup>**

Symbol	Parameter	Conditions	f <sub>HCLK</sub> (MHz)	Typ	Unit
I <sub>DD</sub>	Supply current in Sleep mode from V <sub>DD</sub> supply	All Peripherals enabled	168	65.5	mA
			150	55.5	
			144	53.5	
			120	39.0	
			90	31.6	
			60	21.7	
			30	9.8	
			25	8.8	
		All Peripherals disabled	168	15.7	
			150	13.7	
			144	12.7	
			120	9.7	
			90	7.7	
			60	5.7	
			30	4.7	
			25	2.8	

1. When peripherals are enabled, the power consumption corresponding to the analog part of the peripherals (such as ADC, or DAC) is not included.

**Table 33. Typical current consumption in Sleep mode, regulator OFF<sup>(1)</sup>**

Symbol	Parameter	Conditions	f <sub>HCLK</sub> (MHz)	VDD=3.3 V		VDD=1.8 V		Unit
				I <sub>DD12</sub>	I <sub>DD</sub>	I <sub>DD12</sub>	I <sub>DD</sub>	
I <sub>DD12</sub> /I <sub>DD</sub>	Supply current in Sleep mode from V <sub>12</sub> and V <sub>DD</sub> supply	All Peripherals enabled	180	61.5	1.4	-	-	mA
			168	59.4	1.3	59.4	1.0	
			150	53.9	1.3	53.9	1.0	
			144	49.0	1.3	49.0	1.0	
			120	38.0	1.2	38.0	0.9	
			90	29.3	1.4	29.3	1.1	
			60	20.2	1.2	20.2	0.9	
			30	11.9	1.2	11.9	0.9	
		25	10.4	1.2	10.4	0.9		
		All Peripherals disabled	180	14.9	1.4	-	-	
			168	14.0	1.3	14.0	1.0	
			150	12.6	1.3	12.6	1.0	
			144	11.5	1.3	11.5	1.0	
			120	8.7	1.2	8.7	0.9	
			90	7.1	1.4	7.1	1.1	
			60	5.0	1.2	5.0	0.9	
30	3.1		1.2	3.1	0.9			
25	2.8	1.2	2.8	0.9				

1. When peripherals are enabled, the power consumption corresponding to the analog part of the peripherals (such as ADC, or DAC) is not included.

**I/O system current consumption**

The current consumption of the I/O system has two components: static and dynamic.

**I/O static current consumption**

All the I/Os used as inputs with pull-up generate current consumption when the pin is externally held low. The value of this current consumption can be simply computed by using the pull-up/pull-down resistors values given in [Table 56: I/O static characteristics](#).

For the output pins, any external pull-down or external load must also be considered to estimate the current consumption.

Additional I/O current consumption is due to I/Os configured as inputs if an intermediate voltage level is externally applied. This current consumption is caused by the input Schmitt trigger circuits used to discriminate the input value. Unless this specific configuration is required by the application, this supply current consumption can be avoided by configuring these I/Os in analog mode. This is notably the case of ADC input pins which should be configured as analog inputs.

**Caution:** Any floating input pin can also settle to an intermediate voltage level or switch inadvertently, as a result of external electromagnetic noise. To avoid current consumption related to floating pins, they must either be configured in analog mode, or forced internally to a definite digital value. This can be done either by using pull-up/down resistors or by configuring the pins in output mode.

I/O dynamic current consumption

In addition to the internal peripheral current consumption (see [Table 35: Peripheral current consumption](#)), the I/Os used by an application also contribute to the current consumption. When an I/O pin switches, it uses the current from the MCU supply voltage to supply the I/O pin circuitry and to charge/discharge the capacitive load (internal or external) connected to the pin:

$$I_{SW} = V_{DD} \times f_{SW} \times C$$

where

$I_{SW}$  is the current sunk by a switching I/O to charge/discharge the capacitive load

$V_{DD}$  is the MCU supply voltage

$f_{SW}$  is the I/O switching frequency

$C$  is the total capacitance seen by the I/O pin:  $C = C_{INT} + C_{EXT}$

The test pin is configured in push-pull output mode and is toggled by software at a fixed frequency.

**Table 34. Switching output I/O current consumption<sup>(1)</sup>**

Symbol	Parameter	Conditions	I/O toggling frequency (fsw)	Typ	Unit
I <sub>DDIO</sub>	I/O switching Current	V <sub>DD</sub> = 3.3 V C = C <sub>INT</sub> <sup>(2)</sup>	2 MHz	0.0	mA
			8 MHz	0.2	
			25 MHz	0.6	
			50 MHz	1.1	
			60 MHz	1.3	
			84 MHz	1.8	
			90 MHz	1.9	
		V <sub>DD</sub> = 3.3 V C <sub>EXT</sub> = 0 pF C = C <sub>INT</sub> + C <sub>EXT</sub> + C <sub>S</sub>	2 MHz	0.1	
			8 MHz	0.4	
			25 MHz	1.23	
			50 MHz	2.43	
			60 MHz	2.93	
			84 MHz	3.86	
			90 MHz	4.07	

**Table 34. Switching output I/O current consumption<sup>(1)</sup> (continued)**

Symbol	Parameter	Conditions	I/O toggling frequency (fsw)	Typ	Unit
I <sub>DDIO</sub>	I/O switching Current	$V_{DD} = 3.3\text{ V}$ $C_{EXT} = 10\text{ pF}$ $C = C_{INT} + C_{EXT} + C_S$	2 MHz	0.18	mA
			8 MHz	0.67	
			25 MHz	2.09	
			50 MHz	3.6	
			60 MHz	4.5	
			84 MHz	7.8	
			90 MHz	9.8	
		$V_{DD} = 3.3\text{ V}$ $C_{EXT} = 22\text{ pF}$ $C = C_{INT} + C_{EXT} + C_S$	2 MHz	0.26	
			8 MHz	1.01	
			25 MHz	3.14	
			50 MHz	6.39	
		$V_{DD} = 3.3\text{ V}$ $C_{EXT} = 33\text{ pF}$ $C = C_{INT} + C_{EXT} + C_S$	60 MHz	10.68	
			2 MHz	0.33	
			8 MHz	1.29	
			25 MHz	4.23	
			50 MHz	11.02	

1. C<sub>S</sub> is the PCB board capacitance including the pad pin. C<sub>S</sub> = 7 pF (estimated value).
2. This test is performed by cutting the LQFP176 package pin (pad removal).

**On-chip peripheral current consumption**

The MCU is placed under the following conditions:

- At startup, all I/O pins are in analog input configuration.
- All peripherals are disabled unless otherwise mentioned.
- I/O compensation cell enabled.
- The ART accelerator is ON.
- Scale 1 mode selected, internal digital voltage V12 = 1.32 V.
- HCLK is the system clock. f<sub>PCLK1</sub> = f<sub>HCLK</sub>/4, and f<sub>PCLK2</sub> = f<sub>HCLK</sub>/2.  
 The given value is calculated by measuring the difference of current consumption
  - with all peripherals clocked off
  - with only one peripheral clocked on
  - f<sub>HCLK</sub> = 180 MHz (Scale1 + over-drive ON), f<sub>HCLK</sub> = 144 MHz (Scale 2), f<sub>HCLK</sub> = 120 MHz (Scale 3)"
- Ambient operating temperature is 25 °C and V<sub>DD</sub>=3.3 V.

Table 35. Peripheral current consumption

Peripheral		I <sub>DD</sub> ( Typ) <sup>(1)</sup>			Unit
		Scale 1	Scale 2	Scale 3	
AHB1 (up to 180 MHz)	GPIOA	2.50	2.36	2.08	μA/MHz
	GPIOB	2.56	2.36	2.08	
	GPIOC	2.44	2.29	2.00	
	GPIOD	2.50	2.36	2.08	
	GPIOE	2.44	2.29	2.00	
	GPIOF	2.44	2.29	2.00	
	GPIOG	2.39	2.22	2.00	
	GPIOH	2.33	2.15	1.92	
	GPIOI	2.39	2.22	2.00	
	GPIOJ	2.33	2.15	1.92	
	GPIOK	2.33	2.15	1.92	
	OTG_HS+ULPI	27.00	24.86	21.92	
	CRC	0.44	0.42	0.33	
	BKPSRAM	0.78	0.69	0.58	
	DMA1	25.33	23.26	20.50	
	DMA2	24.72	22.71	20.00	
DMA2D	28.50	26.32	23.33		
ETH_MAC ETH_MAC_TX ETH_MAC_RX ETH_MAC_PTP	21.56	20.07	17.75		
AHB2 (up to 180 MHz)	OTG_FS	25.67	26.67	23.58	μA/MHz
	DCMI	3.72	3.40	3.00	
	RNG	2.28	2.36	2.17	
	Hash	4.39	4.03	3.58	
	Crypto	3.00	2.78	2.42	
AHB3 (up to 180 MHz)	FMC	21.39	19.79	17.50	μA/MHz
Bus matrix <sup>(2)</sup>		14.06	13.19	11.75	μA/MHz

Table 35. Peripheral current consumption (continued)

Peripheral		I <sub>DD</sub> ( Typ) <sup>(1)</sup>			Unit
		Scale 1	Scale 2	Scale 3	
APB1 (up to 45 MHz)	TIM2	17.56	16.42	14.47	μA/MHz
	TIM3	14.22	13.36	11.80	
	TIM4	14.89	13.64	12.13	
	TIM5	17.33	16.42	14.47	
	TIM6	2.89	2.53	2.47	
	TIM7	3.11	2.81	2.47	
	TIM12	7.33	6.97	6.13	
	TIM13	4.89	4.47	4.13	
	TIM14	5.56	5.31	4.80	
	PWR	11.11	10.31	9.13	
	USART2	4.22	3.92	3.47	
	USART3	4.44	4.19	3.80	
	UART4	4.00	3.92	3.47	
	UART5	4.00	3.92	3.47	
	UART7	4.00	3.92	3.47	
	UART8	3.78	3.92	3.47	
	I2C1	4.00	3.92	3.47	
	I2C2	4.00	3.92	3.47	
	I2C3	4.00	3.92	3.47	
	SPI2 <sup>(3)</sup>	3.11	3.08	2.80	
	SPI3 <sup>(3)</sup>	3.56	3.36	3.13	
	I2S2	2.89	2.81	2.47	
	I2S3	3.33	3.08	2.80	
	CAN1	6.89	6.42	5.80	
	CAN2	6.67	6.14	5.47	
	DAC <sup>(4)</sup>	2.89	2.25	2.13	
WWDG	0.89	0.86	0.80		

**Table 35. Peripheral current consumption (continued)**

Peripheral		I <sub>DD</sub> ( Typ) <sup>(1)</sup>			Unit
		Scale 1	Scale 2	Scale 3	
APB2 (up to 90 MHz)	SDIO	8.11	8.75	7.83	μA/MHz
	TIM1	17.11	15.97	14.17	
	TIM8	17.33	16.11	14.33	
	TIM9	7.22	6.67	6.00	
	TIM10	4.56	4.31	3.83	
	TIM11	4.78	4.44	4.00	
	ADC1 <sup>(5)</sup>	4.67	4.31	3.83	
	ADC2 <sup>(5)</sup>	4.78	4.44	4.00	
	ADC3 <sup>(5)</sup>	4.56	4.17	3.67	
	SPI1	1.44	1.39	1.17	
	USART1	4.00	3.75	3.33	
	USART6	4.00	3.75	3.33	
	SPI4	1.44	1.39	1.17	
	SPI5	1.44	1.39	1.17	
	SPI6	1.44	1.39	1.17	
	SYSCFG	0.78	0.69	0.67	
	LCD_TFT	39.89	37.22	33.17	
SAI1	3.78	3.47	3.17		

1. When the I/O compensation cell is ON, I<sub>DD</sub> typical value increases by 0.22 mA.
2. The BusMatrix is automatically active when at least one master is ON.
3. To enable an I2S peripheral, first set the I2SMOD bit and then the I2SE bit in the SPI\_I2SCFGR register.
4. When the DAC is ON and EN1/2 bits are set in DAC\_CR register, add an additional power consumption of 0.8 mA per DAC channel for the analog part.
5. When the ADC is ON (ADON bit set in the ADC\_CR2 register), add an additional power consumption of 1.6 mA per ADC for the analog part.

### 6.3.8 Wakeup time from low-power modes

The wakeup times given in [Table 36](#) are measured starting from the wakeup event trigger up to the first instruction executed by the CPU:

- For Stop or Sleep modes: the wakeup event is WFE.
- WKUP (PA0) pin is used to wakeup from Standby, Stop and Sleep modes.

All timings are derived from tests performed under ambient temperature and V<sub>DD</sub>=3.3 V.

**Table 36. Low-power mode wakeup timings**

Symbol	Parameter	Conditions	Typ <sup>(1)</sup>	Max <sup>(1)</sup>	Unit
t <sub>WUSLEEP</sub> <sup>(2)</sup>	Wakeup from Sleep	-	6	-	CPU clock cycle
t <sub>WUSTOP</sub> <sup>(2)</sup>	Wakeup from Stop mode with MR/LP regulator in normal mode	Main regulator is ON	13.6	-	μs
		Main regulator is ON and Flash memory in Deep power down mode	93	111	
		Low power regulator is ON	22	32	
		Low power regulator is ON and Flash memory in Deep power down mode	103	126	
t <sub>WUSTOP</sub> <sup>(2)</sup>	Wakeup from Stop mode with MR/LP regulator in Under-drive mode	Main regulator in under-drive mode (Flash memory in Deep power-down mode)	125	155	
		Low power regulator in under-drive mode (Flash memory in Deep power-down mode)	105	128	
t <sub>WUSTDBY</sub> <sup>(2)(3)</sup>	Wakeup from Standby mode		318	412	

1. Based on characterization, not tested in production.

2. The wakeup times are measured from the wakeup event to the point in which the application code reads the first

3. t<sub>WUSTDBY</sub> maximum value is given at -40 °C.

### 6.3.9 External clock source characteristics

#### High-speed external user clock generated from an external source

In bypass mode the HSE oscillator is switched off and the input pin is a standard I/O. The external clock signal has to respect the [Table 56: I/O static characteristics](#). However, the recommended clock input waveform is shown in [Figure 27](#).

The characteristics given in [Table 37](#) result from tests performed using an high-speed external clock source, and under ambient temperature and supply voltage conditions summarized in [Table 17](#).

**Table 37. High-speed external user clock characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f <sub>HSE_ext</sub>	External user clock source frequency <sup>(1)</sup>		1	-	50	MHz
V <sub>HSEH</sub>	OSC_IN input pin high level voltage		0.7V <sub>DD</sub>	-	V <sub>DD</sub>	V
V <sub>HSEL</sub>	OSC_IN input pin low level voltage		V <sub>SS</sub>	-	0.3V <sub>DD</sub>	
t <sub>w(HSE)</sub> t <sub>w(HSE)</sub>	OSC_IN high or low time <sup>(1)</sup>		5	-	-	ns
t <sub>r(HSE)</sub> t <sub>f(HSE)</sub>	OSC_IN rise or fall time <sup>(1)</sup>		-	-	10	
C <sub>in(HSE)</sub>	OSC_IN input capacitance <sup>(1)</sup>		-	5	-	pF
DuCy <sub>(HSE)</sub>	Duty cycle		45	-	55	%
I <sub>L</sub>	OSC_IN Input leakage current	V <sub>SS</sub> ≤ V <sub>IN</sub> ≤ V <sub>DD</sub>	-	-	±1	µA

1. Guaranteed by design, not tested in production.

**Low-speed external user clock generated from an external source**

In bypass mode the LSE oscillator is switched off and the input pin is a standard I/O. The external clock signal has to respect the [Table 56: I/O static characteristics](#). However, the recommended clock input waveform is shown in [Figure 28](#).

The characteristics given in [Table 38](#) result from tests performed using an low-speed external clock source, and under ambient temperature and supply voltage conditions summarized in [Table 17](#).

**Table 38. Low-speed external user clock characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f <sub>LSE_ext</sub>	User External clock source frequency <sup>(1)</sup>		-	32.768	1000	kHz
V <sub>LSEH</sub>	OSC32_IN input pin high level voltage		0.7V <sub>DD</sub>	-	V <sub>DD</sub>	V
V <sub>LSEL</sub>	OSC32_IN input pin low level voltage		V <sub>SS</sub>	-	0.3V <sub>DD</sub>	
t <sub>w(LSE)</sub> t <sub>f(LSE)</sub>	OSC32_IN high or low time <sup>(1)</sup>		450	-	-	ns
t <sub>r(LSE)</sub> t <sub>f(LSE)</sub>	OSC32_IN rise or fall time <sup>(1)</sup>		-	-	50	
C <sub>in(LSE)</sub>	OSC32_IN input capacitance <sup>(1)</sup>		-	5	-	pF
DuCy <sub>(LSE)</sub>	Duty cycle		30	-	70	%
I <sub>L</sub>	OSC32_IN Input leakage current	V <sub>SS</sub> ≤ V <sub>IN</sub> ≤ V <sub>DD</sub>	-	-	±1	µA

1. Guaranteed by design, not tested in production.

Figure 27. High-speed external clock source AC timing diagram

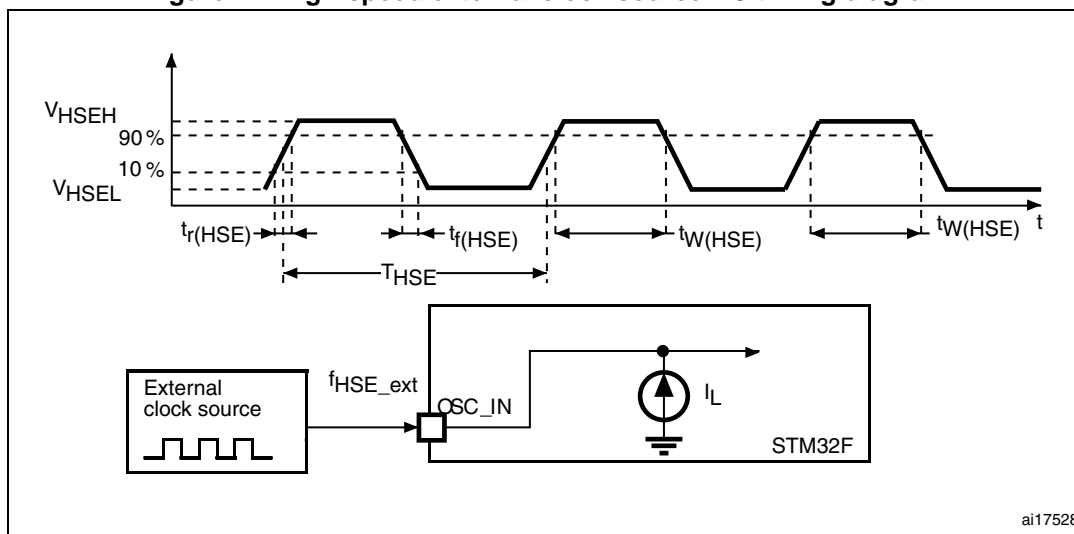
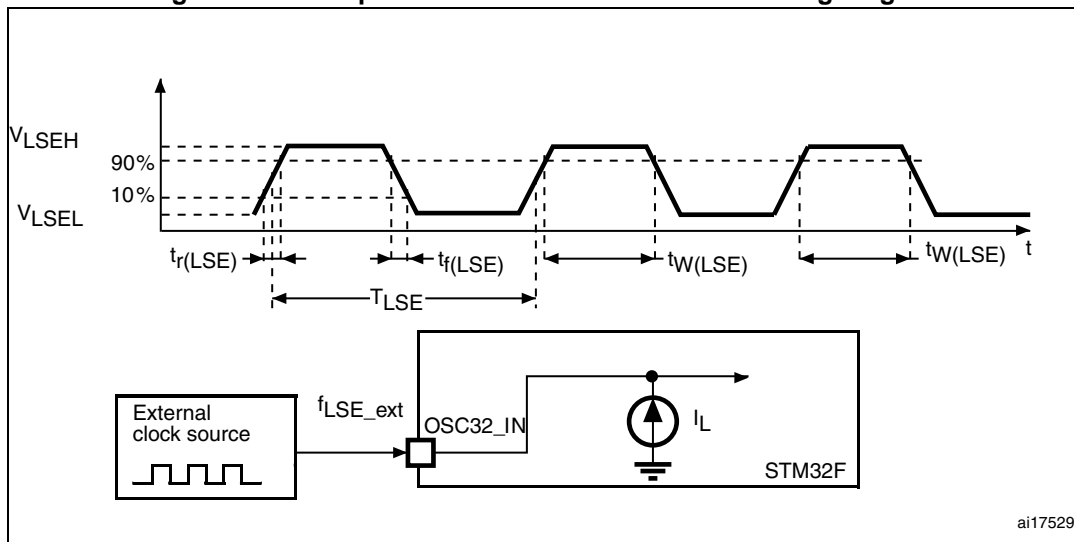


Figure 28. Low-speed external clock source AC timing diagram



**High-speed external clock generated from a crystal/ceramic resonator**

The high-speed external (HSE) clock can be supplied with a 4 to 26 MHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in [Table 39](#). In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

Table 39. HSE 4-26 MHz oscillator characteristics (1)

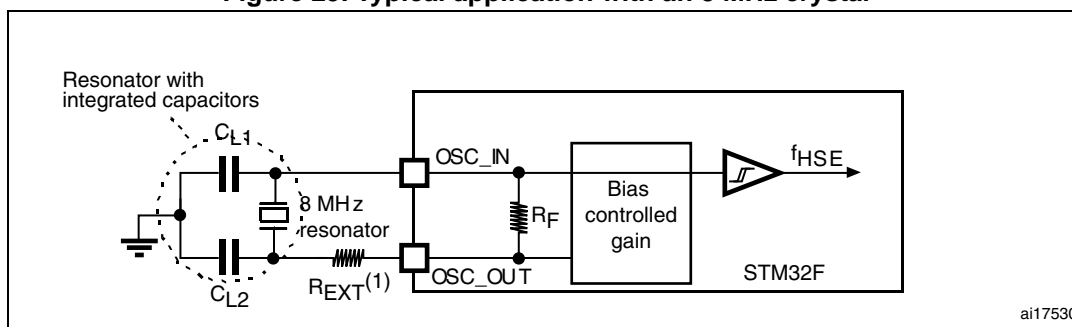
Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{OSC\_IN}$	Oscillator frequency		4	-	26	MHz
$R_F$	Feedback resistor		-	200	-	k $\Omega$
$I_{DD}$	HSE current consumption	$V_{DD}=3.3\text{ V}$ , ESR= 30 $\Omega$ , $C_L=5\text{ pF}@25\text{ MHz}$	-	450	-	$\mu\text{A}$
		$V_{DD}=3.3\text{ V}$ , ESR= 30 $\Omega$ , $C_L=10\text{ pF}@25\text{ MHz}$	-	530	-	
$G_{m\_crit\_max}$	Maximum critical crystal $g_m$	Startup	-	-	1	mA/V
$t_{SU(HSE)}^{(2)}$	Startup time	$V_{DD}$ is stabilized	-	2	-	ms

1. Guaranteed by design, not tested in production.
2.  $t_{SU(HSE)}$  is the startup time measured from the moment it is enabled (by software) to a stabilized 8 MHz oscillation is reached. This value is based on characterization and not tested in production. It is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

For  $C_{L1}$  and  $C_{L2}$ , it is recommended to use high-quality external ceramic capacitors in the 5 pF to 25 pF range (typ.), designed for high-frequency applications, and selected to match the requirements of the crystal or resonator (see Figure 29).  $C_{L1}$  and  $C_{L2}$  are usually the same size. The crystal manufacturer typically specifies a load capacitance which is the series combination of  $C_{L1}$  and  $C_{L2}$ . PCB and MCU pin capacitance must be included (10 pF can be used as a rough estimate of the combined pin and board capacitance) when sizing  $C_{L1}$  and  $C_{L2}$ .

Note: For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website [www.st.com](http://www.st.com).

Figure 29. Typical application with an 8 MHz crystal



1.  $R_{EXT}$  value depends on the crystal characteristics.

### Low-speed external clock generated from a crystal/ceramic resonator

The low-speed external (LSE) clock can be supplied with a 32.768 kHz crystal/ceramic resonator oscillator. All the information given in this paragraph are based on characterization results obtained with typical external components specified in Table 40. In the application, the resonator and the load capacitors have to be placed as close as possible to the oscillator pins in order to minimize output distortion and startup stabilization time. Refer to the crystal resonator manufacturer for more details on the resonator characteristics (frequency, package, accuracy).

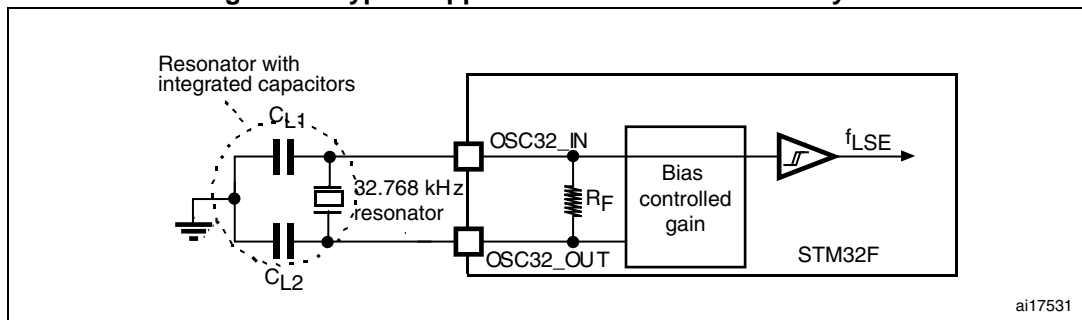
**Table 40. LSE oscillator characteristics ( $f_{LSE} = 32.768 \text{ kHz}$ ) <sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$R_F$	Feedback resistor		-	18.4	-	$M\Omega$
$I_{DD}$	LSE current consumption		-	-	1	$\mu A$
$G_{m\_crit\_max}$	Maximum critical crystal $g_m$	Startup	-	-	0.56	$\mu A/V$
$t_{SU(LSE)}^{(2)}$	startup time	$V_{DD}$ is stabilized	-	2	-	s

1. Guaranteed by design, not tested in production.
2.  $t_{SU(LSE)}$  is the startup time measured from the moment it is enabled (by software) to a stabilized 32.768 kHz oscillation is reached. This value is based on characterization and not tested in production. It is measured for a standard crystal resonator and it can vary significantly with the crystal manufacturer.

*Note:* For information on selecting the crystal, refer to the application note AN2867 "Oscillator design guide for ST microcontrollers" available from the ST website [www.st.com](http://www.st.com).

**Figure 30. Typical application with a 32.768 kHz crystal**



### 6.3.10 Internal clock source characteristics

The parameters given in [Table 41](#) and [Table 42](#) are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#).

High-speed internal (HSI) RC oscillator

Figure 31. LACC<sub>HSI</sub> versus temperature



Table 41. HSI oscillator characteristics (1)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
f <sub>HSI</sub>	Frequency		-	16	-	MHz	
ACC <sub>HSI</sub>	Accuracy of the HSI oscillator	User-trimmed with the RCC_CR register <sup>(2)</sup>	-	-	1	%	
		Factory-calibrated	T <sub>A</sub> = -40 to 105 °C <sup>(3)</sup>	-8	-	4.5	%
			T <sub>A</sub> = -10 to 85 °C <sup>(3)</sup>	-4	-	4	%
		T <sub>A</sub> = 25 °C	-1	-	1	%	
t <sub>su(HSI)</sub> <sup>(2)</sup>	HSI oscillator startup time		-	2.2	4	µs	
I <sub>DD(HSI)</sub> <sup>(2)</sup>	HSI oscillator power consumption		-	60	80	µA	

- V<sub>DD</sub> = 3.3 V, T<sub>A</sub> = -40 to 105 °C unless otherwise specified.
- Guaranteed by design, not tested in production
- Based on characterization, not tested in production.

- Based on characterisation results, not tested in production.

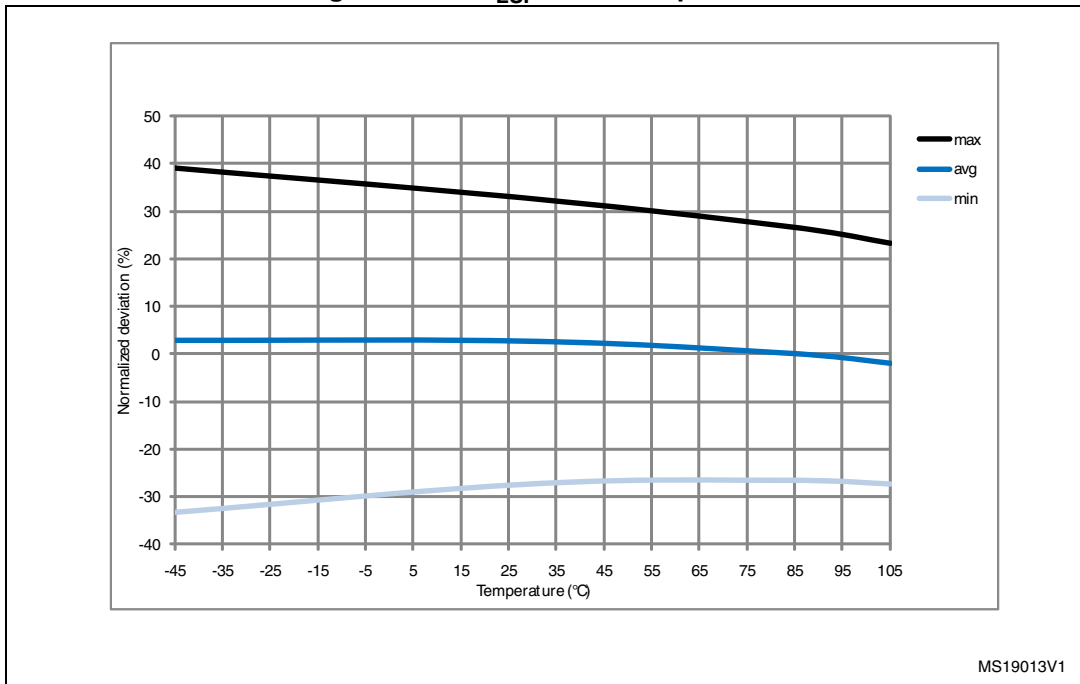
Low-speed internal (LSI) RC oscillator

Table 42. LSI oscillator characteristics (1)

Symbol	Parameter	Min	Typ	Max	Unit
f <sub>LSI</sub> <sup>(2)</sup>	Frequency	17	32	47	kHz
t <sub>su(LSI)</sub> <sup>(3)</sup>	LSI oscillator startup time	-	15	40	µs
I <sub>DD(LSI)</sub> <sup>(3)</sup>	LSI oscillator power consumption	-	0.4	0.6	µA

- V<sub>DD</sub> = 3 V, T<sub>A</sub> = -40 to 105 °C unless otherwise specified.
- Based on characterization, not tested in production.
- Guaranteed by design, not tested in production.

Figure 32. ACC<sub>LSI</sub> versus temperature



MS19013V1

### 6.3.11 PLL characteristics

The parameters given in [Table 43](#) and [Table 44](#) are derived from tests performed under temperature and V<sub>DD</sub> supply voltage conditions summarized in [Table 17](#).

Table 43. Main PLL characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f <sub>PLL_IN</sub>	PLL input clock <sup>(1)</sup>		0.95 <sup>(2)</sup>	1	2.10	MHz
f <sub>PLL_OUT</sub>	PLL multiplier output clock		24	-	180	MHz
f <sub>PLL48_OUT</sub>	48 MHz PLL multiplier output clock		-	48	75	MHz
f <sub>VCO_OUT</sub>	PLL VCO output		192	-	432	MHz
t <sub>LOCK</sub>	PLL lock time	VCO freq = 192 MHz	75	-	200	μs
		VCO freq = 432 MHz	100	-	300	

**Table 43. Main PLL characteristics (continued)**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
Jitter <sup>(3)</sup>	Cycle-to-cycle jitter	System clock 120 MHz	RMS	-	25	-	ps
			peak to peak	-	±150	-	
	Period Jitter		RMS	-	15	-	
			peak to peak	-	±200	-	
	Main clock output (MCO) for RMI Ethernet	Cycle to cycle at 50 MHz on 1000 samples	-	32	-		
	Main clock output (MCO) for MII Ethernet	Cycle to cycle at 25 MHz on 1000 samples	-	40	-		
	Bit Time CAN jitter	Cycle to cycle at 1 MHz on 1000 samples	-	330	-		
I <sub>DD(PLL)</sub> <sup>(4)</sup>	PLL power consumption on VDD	VCO freq = 192 MHz VCO freq = 432 MHz	0.15 0.45	-	0.40 0.75	mA	
I <sub>DDA(PLL)</sub> <sup>(4)</sup>	PLL power consumption on VDDA	VCO freq = 192 MHz VCO freq = 432 MHz	0.30 0.55	-	0.40 0.85	mA	

1. Take care of using the appropriate division factor M to obtain the specified PLL input clock values. The M factor is shared between PLL and PLLI2S.
2. Guaranteed by design, not tested in production.
3. The use of 2 PLLs in parallel could degraded the Jitter up to +30%.
4. Based on characterization, not tested in production.

**Table 44. PLLI2S (audio PLL) characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit	
f <sub>PLLI2S_IN</sub>	PLLI2S input clock <sup>(1)</sup>		0.95 <sup>(2)</sup>	1	2.10	MHz	
f <sub>PLLI2S_OUT</sub>	PLLI2S multiplier output clock		-	-	216	MHz	
f <sub>VCO_OUT</sub>	PLLI2S VCO output		192	-	432	MHz	
t <sub>LOCK</sub>	PLLI2S lock time	VCO freq = 192 MHz	75	-	200	µs	
		VCO freq = 432 MHz	100	-	300		
Jitter <sup>(3)</sup>	Master I2S clock jitter	Cycle to cycle at 12.288 MHz on 48KHz period, N=432, R=5	RMS	-	90	-	ps
			peak to peak	-	±280	-	
	Average frequency of 12.288 MHz N = 432, R = 5 on 1000 samples	-	90	-			
	WS I2S clock jitter	Cycle to cycle at 48 KHz on 1000 samples	-	400	-	ps	

Table 44. PLLI2S (audio PLL) characteristics (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{DD(PLLI2S)}^{(4)}$	PLLI2S power consumption on $V_{DD}$	VCO freq = 192 MHz VCO freq = 432 MHz	0.15 0.45	-	0.40 0.75	mA
$I_{DDA(PLLI2S)}^{(4)}$	PLLI2S power consumption on $V_{DDA}$	VCO freq = 192 MHz VCO freq = 432 MHz	0.30 0.55	-	0.40 0.85	mA

1. Take care of using the appropriate division factor M to have the specified PLL input clock values.
2. Guaranteed by design, not tested in production.
3. Value given with main PLL running.
4. Based on characterization, not tested in production.

Table 45. PLLISAI (audio and LCD-TFT PLL) characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{PLLSAI\_IN}$	PLLSAI input clock <sup>(1)</sup>		0.95 <sup>(2)</sup>	1	2.10	MHz
$f_{PLLSAI\_OUT}$	PLLSAI multiplier output clock		-	-	216	MHz
$f_{VCO\_OUT}$	PLLSAI VCO output		192	-	432	MHz
$t_{LOCK}$	PLLSAI lock time	VCO freq = 192 MHz	75	-	200	$\mu$ s
		VCO freq = 432 MHz	100	-	300	
Jitter <sup>(3)</sup>	Main SAI clock jitter	Cycle to cycle at 12.288 MHz on 48KHz period, N=432, R=5	RMS	-	90	-
			peak to peak	-	$\pm$ 280	-
		Average frequency of 12.288 MHz N = 432, R = 5 on 1000 samples		-	90	-
	FS clock jitter	Cycle to cycle at 48 KHz on 1000 samples	-	400	-	ps
$I_{DD(PLLSAI)}^{(4)}$	PLLSAI power consumption on $V_{DD}$	VCO freq = 192 MHz VCO freq = 432 MHz	0.15 0.45	-	0.40 0.75	mA
$I_{DDA(PLLSAI)}^{(4)}$	PLLSAI power consumption on $V_{DDA}$	VCO freq = 192 MHz VCO freq = 432 MHz	0.30 0.55	-	0.40 0.85	mA

1. Take care of using the appropriate division factor M to have the specified PLL input clock values.
2. Guaranteed by design, not tested in production.
3. Value given with main PLL running.
4. Based on characterization, not tested in production.

### 6.3.12 PLL spread spectrum clock generation (SSCG) characteristics

The spread spectrum clock generation (SSCG) feature allows to reduce electromagnetic interferences (see [Table 52: EMI characteristics](#)). It is available only on the main PLL.

**Table 46. SSCG parameters constraint**

Symbol	Parameter	Min	Typ	Max <sup>(1)</sup>	Unit
$f_{Mod}$	Modulation frequency	-	-	10	KHz
md	Peak modulation depth	0.25	-	2	%
MODEPER * INCSTEP		-	-	$2^{15}-1$	-

1. Guaranteed by design, not tested in production.

#### Equation 1

The frequency modulation period (MODEPER) is given by the equation below:

$$\text{MODEPER} = \text{round}[f_{\text{PLL\_IN}} / (4 \times f_{\text{Mod}})]$$

$f_{\text{PLL\_IN}}$  and  $f_{\text{Mod}}$  must be expressed in Hz.

As an example:

If  $f_{\text{PLL\_IN}} = 1 \text{ MHz}$ , and  $f_{\text{MOD}} = 1 \text{ kHz}$ , the modulation depth (MODEPER) is given by equation 1:

$$\text{MODEPER} = \text{round}[10^6 / (4 \times 10^3)] = 250$$

#### Equation 2

Equation 2 allows to calculate the increment step (INCSTEP):

$$\text{INCSTEP} = \text{round}[(2^{15} - 1) \times \text{md} \times \text{PLL}_N / (100 \times 5 \times \text{MODEPER})]$$

$f_{\text{VCO\_OUT}}$  must be expressed in MHz.

With a modulation depth (md) =  $\pm 2 \%$  (4 % peak to peak), and  $\text{PLL}_N = 240$  (in MHz):

$$\text{INCSTEP} = \text{round}[(2^{15} - 1) \times 2 \times 240 / (100 \times 5 \times 250)] = 126 \text{md}(\text{quantitized})\%$$

An amplitude quantization error may be generated because the linear modulation profile is obtained by taking the quantized values (rounded to the nearest integer) of MODPER and INCSTEP. As a result, the achieved modulation depth is quantized. The percentage quantized modulation depth is given by the following formula:

$$\text{md}_{\text{quantized}}\% = (\text{MODEPER} \times \text{INCSTEP} \times 100 \times 5) / ((2^{15} - 1) \times \text{PLL}_N)$$

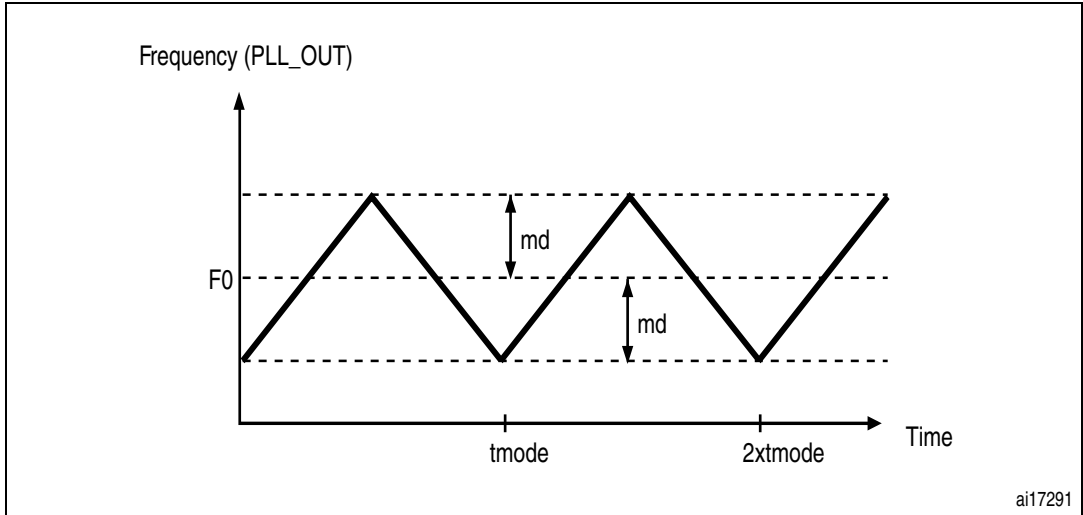
As a result:

$$\text{md}_{\text{quantized}}\% = (250 \times 126 \times 100 \times 5) / ((2^{15} - 1) \times 240) = 2.002\%(\text{peak})$$

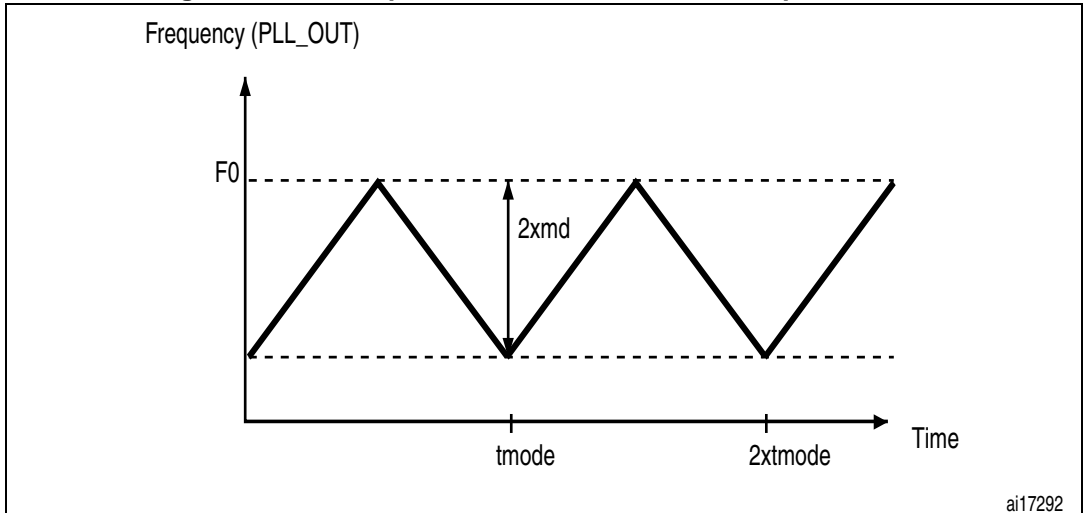
Figure 33 and Figure 34 show the main PLL output clock waveforms in center spread and down spread modes, where:

- F0 is  $f_{PLL\_OUT}$  nominal.
- $T_{mode}$  is the modulation period.
- md is the modulation depth.

**Figure 33. PLL output clock waveforms in center spread mode**



**Figure 34. PLL output clock waveforms in down spread mode**



### 6.3.13 Memory characteristics

#### Flash memory

The characteristics are given at  $T_A = -40$  to  $105$  °C unless otherwise specified.

The devices are shipped to customers with the Flash memory erased.

**Table 47. Flash memory characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$I_{DD}$	Supply current	Write / Erase 8-bit mode, $V_{DD} = 1.8$ V	-	5	-	mA
		Write / Erase 16-bit mode, $V_{DD} = 2.1$ V	-	8	-	
		Write / Erase 32-bit mode, $V_{DD} = 3.3$ V	-	12	-	

**Table 48. Flash memory programming**

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Typ	Max <sup>(1)</sup>	Unit
$t_{prog}$	Word programming time	Program/erase parallelism (PSIZE) = x 8/16/32	-	16	100 <sup>(2)</sup>	$\mu$ s
$t_{ERASE16KB}$	Sector (16 KB) erase time	Program/erase parallelism (PSIZE) = x 8	-	400	800	ms
		Program/erase parallelism (PSIZE) = x 16	-	300	600	
		Program/erase parallelism (PSIZE) = x 32	-	250	500	
$t_{ERASE64KB}$	Sector (64 KB) erase time	Program/erase parallelism (PSIZE) = x 8	-	1200	2400	ms
		Program/erase parallelism (PSIZE) = x 16	-	700	1400	
		Program/erase parallelism (PSIZE) = x 32	-	550	1100	
$t_{ERASE128KB}$	Sector (128 KB) erase time	Program/erase parallelism (PSIZE) = x 8	-	2	4	s
		Program/erase parallelism (PSIZE) = x 16	-	1.3	2.6	
		Program/erase parallelism (PSIZE) = x 32	-	1	2	
$t_{ME}$	Mass erase time	Program/erase parallelism (PSIZE) = x 8	-	16	32	s
		Program/erase parallelism (PSIZE) = x 16	-	11	22	
		Program/erase parallelism (PSIZE) = x 32	-	8	16	

**Table 48. Flash memory programming (continued)**

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Typ	Max <sup>(1)</sup>	Unit
t <sub>BE</sub>	Bank erase time	Program/erase parallelism (PSIZE) = x 8	-	16	32	s
		Program/erase parallelism (PSIZE) = x 16	-	11	22	
		Program/erase parallelism (PSIZE) = x 32	-	8	16	
V <sub>prog</sub>	Programming voltage	32-bit program operation	2.7	-	3.6	V
		16-bit program operation	2.1	-	3.6	V
		8-bit program operation	1.8	-	3.6	V

1. Based on characterization, not tested in production.
2. The maximum programming time is measured after 100K erase operations.

**Table 49. Flash memory programming with V<sub>PP</sub>**

Symbol	Parameter	Conditions	Min <sup>(1)</sup>	Typ	Max <sup>(1)</sup>	Unit
t <sub>prog</sub>	Double word programming	T <sub>A</sub> = 0 to +40 °C V <sub>DD</sub> = 3.3 V V <sub>PP</sub> = 8.5 V	-	16	100 <sup>(2)</sup>	μs
t <sub>ERASE16KB</sub>	Sector (16 KB) erase time		-	230	-	ms
t <sub>ERASE64KB</sub>	Sector (64 KB) erase time		-	490	-	
t <sub>ERASE128KB</sub>	Sector (128 KB) erase time		-	875	-	
t <sub>ME</sub>	Mass erase time		-	6.9	-	s
t <sub>BE</sub>	Bank erase time		-	6.9	-	s
V <sub>prog</sub>	Programming voltage		2.7	-	3.6	V
V <sub>PP</sub>	V <sub>PP</sub> voltage range		7	-	9	V
I <sub>PP</sub>	Minimum current sunk on the V <sub>PP</sub> pin		10	-	-	mA
t <sub>VPP</sub> <sup>(3)</sup>	Cumulative time during which V <sub>PP</sub> is applied		-	-	1	hour

1. Guaranteed by design, not tested in production.
2. The maximum programming time is measured after 100K erase operations.
3. V<sub>PP</sub> should only be connected during programming/erasing.

**Table 50. Flash memory endurance and data retention**

Symbol	Parameter	Conditions	Value	Unit
			Min <sup>(1)</sup>	
N <sub>END</sub>	Endurance	T <sub>A</sub> = -40 to +85 °C (6 suffix versions) T <sub>A</sub> = -40 to +105 °C (7 suffix versions)	10	kcycles
t <sub>RET</sub>	Data retention	1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 85 °C	30	Years
		1 kcycle <sup>(2)</sup> at T <sub>A</sub> = 105 °C	10	
		10 kcycles <sup>(2)</sup> at T <sub>A</sub> = 55 °C	20	

1. Based on characterization, not tested in production.
2. Cycling performed over the whole temperature range.

### 6.3.14 EMC characteristics

Susceptibility tests are performed on a sample basis during device characterization.

#### Functional EMS (electromagnetic susceptibility)

While a simple application is executed on the device (toggling 2 LEDs through I/O ports), the device is stressed by two electromagnetic events until a failure occurs. The failure is indicated by the LEDs:

- **Electrostatic discharge (ESD)** (positive and negative) is applied to all device pins until a functional disturbance occurs. This test is compliant with the IEC 61000-4-2 standard.
- **FTB**: A burst of fast transient voltage (positive and negative) is applied to V<sub>DD</sub> and V<sub>SS</sub> through a 100 pF capacitor, until a functional disturbance occurs. This test is compliant with the IEC 61000-4-4 standard.

A device reset allows normal operations to be resumed.

The test results are given in [Table 51](#). They are based on the EMS levels and classes defined in application note AN1709.

**Table 51. EMS characteristics**

Symbol	Parameter	Conditions	Level/Class
V <sub>FESD</sub>	Voltage limits to be applied on any I/O pin to induce a functional disturbance	V <sub>DD</sub> = 3.3 V, LQFP176, T <sub>A</sub> = +25 °C, f <sub>HCLK</sub> = 168 MHz, conforms to IEC 61000-4-2	2B
V <sub>EFTB</sub>	Fast transient voltage burst limits to be applied through 100 pF on V <sub>DD</sub> and V <sub>SS</sub> pins to induce a functional disturbance	V <sub>DD</sub> = 3.3 V, LQFP176, T <sub>A</sub> = +25 °C, f <sub>HCLK</sub> = 168 MHz, conforms to IEC 61000-4-2	4A

When the application is exposed to a noisy environment, it is recommended to avoid pin exposition to disturbances. The pins showing a middle range robustness are: PA0, PA1, PA2, PH2, PH3, PH4, PH5, PA3, PA4, PA5, PA6, PA7, PC4, and PC5.

As a consequence, it is recommended to add a serial resistor (1 kΩ) located as close as possible to the MCU to the pins exposed to noise (connected to tracks longer than 50 mm on PCB).

**Designing hardened software to avoid noise problems**

EMC characterization and optimization are performed at component level with a typical application environment and simplified MCU software. It should be noted that good EMC performance is highly dependent on the user application and the software in particular.

Therefore it is recommended that the user applies EMC software optimization and prequalification tests in relation with the EMC level requested for his application.

Software recommendations

The software flowchart must include the management of runaway conditions such as:

- Corrupted program counter
- Unexpected reset
- Critical Data corruption (control registers...)

Prequalification trials

Most of the common failures (unexpected reset and program counter corruption) can be reproduced by manually forcing a low state on the NRST pin or the Oscillator pins for 1 second.

To complete these trials, ESD stress can be applied directly on the device, over the range of specification values. When unexpected behavior is detected, the software can be hardened to prevent unrecoverable errors occurring (see application note AN1015).

**Electromagnetic Interference (EMI)**

The electromagnetic field emitted by the device are monitored while a simple application, executing EEMBC<sup>?</sup> code, is running. This emission test is compliant with SAE IEC61967-2 standard which specifies the test board and the pin loading.

**Table 52. EMI characteristics**

Symbol	Parameter	Conditions	Monitored frequency band	Max vs. [f <sub>HSE</sub> /f <sub>CPU</sub> ]	Max vs. [f <sub>HSE</sub> /f <sub>CPU</sub> ]	Unit
				25/168 MHz	25/180 MHz	
S <sub>EMI</sub>	Peak level	V <sub>DD</sub> = 3.3 V, T <sub>A</sub> = 25 °C, LQFP176 package, conforming to SAE J1752/3 EEMBC, ART ON, all peripheral clocks enabled, clock dithering disabled.	0.1 to 30 MHz	16	19	dBµV
			30 to 130 MHz	23	23	
			130 MHz to 1GHz	25	22	
			SAE EMI Level	4	4	-
		V <sub>DD</sub> = 3.3 V, T <sub>A</sub> = 25 °C, LQFP176 package, conforming to SAE J1752/3 EEMBC, ART ON, all peripheral clocks enabled, clock dithering enabled	0.1 to 30 MHz	17	16	dBµV
			30 to 130 MHz	8	10	
			130 MHz to 1GHz	11	16	
			SAE EMI level	3.5	3.5	-

### 6.3.15 Absolute maximum ratings (electrical sensitivity)

Based on three different tests (ESD, LU) using specific measurement methods, the device is stressed in order to determine its performance in terms of electrical sensitivity.

#### Electrostatic discharge (ESD)

Electrostatic discharges (a positive then a negative pulse separated by 1 second) are applied to the pins of each sample according to each pin combination. The sample size depends on the number of supply pins in the device (3 parts × (n+1) supply pins). This test conforms to the JESD22-A114/C101 standard.

**Table 53. ESD absolute maximum ratings**

Symbol	Ratings	Conditions	Class	Maximum value <sup>(1)</sup>	Unit
V <sub>ESD(HBM)</sub>	Electrostatic discharge voltage (human body model)	T <sub>A</sub> = +25 °C conforming to JESD22-A114	2	2000	V
V <sub>ESD(CDM)</sub>	Electrostatic discharge voltage (charge device model)	T <sub>A</sub> = +25 °C conforming to JESD22-C101	II	500	

1. Based on characterization results, not tested in production.

#### Static latchup

Two complementary static tests are required on six parts to assess the latchup performance:

- A supply overvoltage is applied to each power supply pin
- A current injection is applied to each input, output and configurable I/O pin

These tests are compliant with EIA/JESD 78A IC latchup standard.

**Table 54. Electrical sensitivities**

Symbol	Parameter	Conditions	Class
LU	Static latch-up class	T <sub>A</sub> = +105 °C conforming to JESD78A	II level A

### 6.3.16 I/O current injection characteristics

As a general rule, current injection to the I/O pins, due to external voltage below V<sub>SS</sub> or above V<sub>DD</sub> (for standard, 3 V-capable I/O pins) should be avoided during normal product operation. However, in order to give an indication of the robustness of the microcontroller in cases when abnormal injection accidentally happens, susceptibility tests are performed on a sample basis during device characterization.

#### Functional susceptibility to I/O current injection

While a simple application is executed on the device, the device is stressed by injecting current into the I/O pins programmed in floating input mode. While current is injected into the I/O pin, one at a time, the device is checked for functional failures.

The failure is indicated by an out of range parameter: ADC error above a certain limit (>5 LSB TUE), out of conventional limits of induced leakage current on adjacent pins (out of – 5 μA/+0 μA range), or other functional failure (for example reset, oscillator frequency deviation).

Negative induced leakage current is caused by negative injection and positive induced leakage current by positive injection.

The test results are given in [Table 55](#).

**Table 55. I/O current injection susceptibility<sup>(1)</sup>**

Symbol	Description	Functional susceptibility		Unit
		Negative injection	Positive injection	
I <sub>INJ</sub>	Injected current on BOOT0 pin	-0	NA	mA
	Injected current on NRST pin	-0	NA	
	Injected current on PA0, PA1, PA2, PA3, PA6, PA7, PB0, PC0, PC1, PC2, PC3, PC4, PC5, PH1, PH2, PH3, PH4, PH5	-0	NA	
	Injected current on TTa pins: PA4 and PA5	-0	+5	
	Injected current on any other FT pin	-5	NA	

1. NA = not applicable.

*Note:* It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.

### 6.3.17 I/O port characteristics

#### General input/output characteristics

Unless otherwise specified, the parameters given in [Table 56: I/O static characteristics](#) are derived from tests performed under the conditions summarized in [Table 17](#). All I/Os are CMOS and TTL compliant.

**Table 56. I/O static characteristics**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
V <sub>IL</sub>	FT, TTa and NRST I/O input low level voltage	1.7 V ≤ V <sub>DD</sub> ≤ 3.6 V	-	-	0.35V <sub>DD</sub> -0.04 <sup>(1)</sup>	V
					0.3V <sub>DD</sub> <sup>(2)</sup>	
	BOOT0 I/O input low level voltage	1.75 V ≤ V <sub>DD</sub> ≤ 3.6 V, -40 °C ≤ T <sub>A</sub> ≤ 105 °C	-	-	0.1V <sub>DD</sub> +0.1 <sup>(1)</sup>	
	1.7 V ≤ V <sub>DD</sub> ≤ 3.6 V, 0 °C ≤ T <sub>A</sub> ≤ 105 °C	-	-			

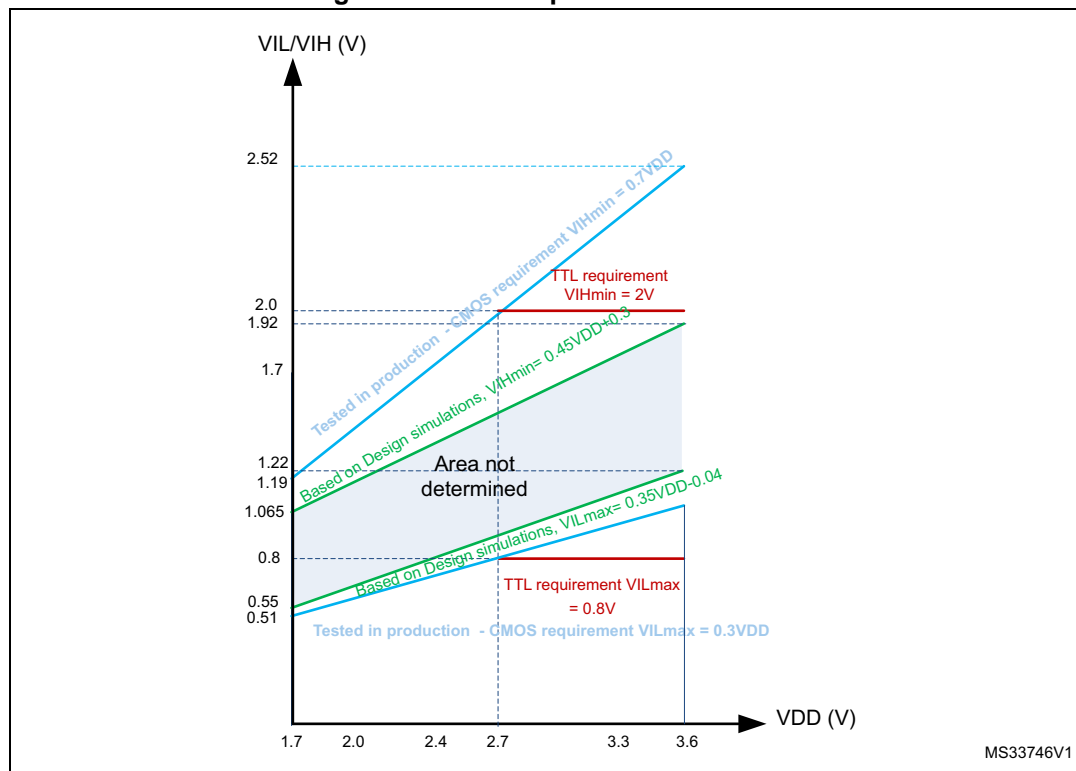
Table 56. I/O static characteristics (continued)

Symbol	Parameter		Conditions	Min	Typ	Max	Unit
$V_{IH}$	FT, TTA and NRST I/O input high level voltage <sup>(5)</sup>		$1.7 V \leq V_{DD} \leq 3.6 V$	$0.45V_{DD} + 0.3$ <sup>(1)</sup>	-	-	V
				$0.4V_{DD}$ <sup>(2)</sup>			
$V_{IH}$	BOOT0 I/O input high level voltage		$1.75 V \leq V_{DD} \leq 3.6 V$ , $-40\text{ }^{\circ}\text{C} \leq T_A \leq 105\text{ }^{\circ}\text{C}$	$0.17V_{DD} + 0.7$ <sup>(1)</sup>	-	-	V
			$1.7 V \leq V_{DD} \leq 3.6 V$ , $0\text{ }^{\circ}\text{C} \leq T_A \leq 105\text{ }^{\circ}\text{C}$				
$V_{HYS}$	FT, TTA and NRST I/O input hysteresis		$1.7 V \leq V_{DD} \leq 3.6 V$	$0.45V_{DD} + 0.3$ <sup>(1)</sup>	-	-	V
	BOOT0 I/O input hysteresis		$1.75 V \leq V_{DD} \leq 3.6 V$ , $-40\text{ }^{\circ}\text{C} \leq T_A \leq 105\text{ }^{\circ}\text{C}$	$10\%V_{DDIO}$ <sup>(1)</sup> <sup>(3)</sup>	-	-	
$I_{lkg}$	I/O input leakage current <sup>(4)</sup>		$V_{SS} \leq V_{IN} \leq V_{DD}$	-	-	$\pm 1$	$\mu\text{A}$
	I/O FT input leakage current <sup>(5)</sup>		$V_{IN} = 5 V$	-	-	3	
$R_{PU}$	Weak pull-up equivalent resistor <sup>(6)</sup>	All pins except for PA10/PB12 (OTG_FS_ID, OTG_HS_ID)	$V_{IN} = V_{SS}$	30	40	50	k $\Omega$
		PA10/PB12 (OTG_FS_ID, OTG_HS_ID)	-	7	10	14	
$R_{PD}$	Weak pull-down equivalent resistor <sup>(7)</sup>	All pins except for PA10/PB12 (OTG_FS_ID, OTG_HS_ID)	$V_{IN} = V_{DD}$	30	40	50	k $\Omega$
		PA10/PB12 (OTG_FS_ID, OTG_HS_ID)	-	7	10	14	
$C_{IO}$ <sup>(8)</sup>	I/O pin capacitance		-	-	5	-	pF

1. Data based on design simulation.
2. Tested in production.
3. With a minimum of 200 mV.
4. Leakage could be higher than the maximum value, if negative current is injected on adjacent pins, Refer to [Table 55: I/O current injection susceptibility](#)
5. To sustain a voltage higher than VDD +0.3 V, the internal pull-up/pull-down resistors must be disabled. Leakage could be higher than the maximum value, if negative current is injected on adjacent pins. Refer to [Table 55: I/O current injection susceptibility](#)
6. Pull-up resistors are designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance is minimum (~10% order).
7. Pull-down resistors are designed with a true resistance in series with a switchable NMOS. This NMOS contribution to the series resistance is minimum (~10% order).
8. Hysteresis voltage between Schmitt trigger switching levels. Based on characterization, not tested in production.

All I/Os are CMOS and TTL compliant (no software configuration required). Their characteristics cover more than the strict CMOS-technology or TTL parameters. The coverage of these requirements for FT I/Os is shown in [Figure 35](#).

Figure 35. FT I/O input characteristics



### Output driving current

The GPIOs (general purpose input/outputs) can sink or source up to  $\pm 8$  mA, and sink or source up to  $\pm 20$  mA (with a relaxed  $V_{OL}/V_{OH}$ ) except PC13, PC14, PC15 and PI8 which can sink or source up to  $\pm 3$  mA. When using the PC13 to PC15 and PI8 GPIOs in output mode, the speed should not exceed 2 MHz with a maximum load of 30 pF.

In the user application, the number of I/O pins which can drive current must be limited to respect the absolute maximum rating specified in [Section 6.2](#). In particular:

- The sum of the currents sourced by all the I/Os on  $V_{DD}$ , plus the maximum Run consumption of the MCU sourced on  $V_{DD}$ , cannot exceed the absolute maximum rating  $\Sigma I_{VDD}$  (see [Table 15](#)).
- The sum of the currents sunk by all the I/Os on  $V_{SS}$  plus the maximum Run consumption of the MCU sunk on  $V_{SS}$  cannot exceed the absolute maximum rating  $\Sigma I_{VSS}$  (see [Table 15](#)).

### Output voltage levels

Unless otherwise specified, the parameters given in [Table 57](#) are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#). All I/Os are CMOS and TTL compliant.

**Table 57. Output voltage characteristics**

Symbol	Parameter	Conditions	Min	Max	Unit
$V_{OL}^{(1)}$	Output low level voltage for an I/O pin	CMOS port <sup>(2)</sup> $I_{IO} = +8 \text{ mA}$ $2.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	0.4	V
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$V_{DD}-0.4$	-	
$V_{OL}^{(1)}$	Output low level voltage for an I/O pin	TTL port <sup>(2)</sup> $I_{IO} = +8 \text{ mA}$ $2.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	0.4	V
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		2.4	-	
$V_{OL}^{(1)}$	Output low level voltage for an I/O pin	$I_{IO} = +20 \text{ mA}$ $2.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	1.3 <sup>(4)</sup>	V
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$V_{DD}-1.3^{(4)}$	-	
$V_{OL}^{(1)}$	Output low level voltage for an I/O pin	$I_{IO} = +6 \text{ mA}$ $1.8 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	0.4 <sup>(4)</sup>	V
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$V_{DD}-0.4^{(4)}$	-	
$V_{OL}^{(1)}$	Output low level voltage for an I/O pin	$I_{IO} = +4 \text{ mA}$ $1.7 \text{ V} \leq V_{DD} \leq 3.6 \text{ V}$	-	0.4 <sup>(5)</sup>	V
$V_{OH}^{(3)}$	Output high level voltage for an I/O pin		$V_{DD}-0.4^{(5)}$	-	

1. The  $I_{IO}$  current sunk by the device must always respect the absolute maximum rating specified in [Table 15](#) and the sum of  $I_{IO}$  (I/O ports and control pins) must not exceed  $I_{VSS}$ .
2. TTL and CMOS outputs are compatible with JEDEC standards JESD36 and JESD52.
3. The  $I_{IO}$  current sourced by the device must always respect the absolute maximum rating specified in [Table 15](#) and the sum of  $I_{IO}$  (I/O ports and control pins) must not exceed  $I_{VDD}$ .
4. Based on characterization data.
5. Data based on design simulation.

Input/output AC characteristics

The definition and values of input/output AC characteristics are given in [Figure 36](#) and [Table 58](#), respectively.

Unless otherwise specified, the parameters given in [Table 58](#) are derived from tests performed under the ambient temperature and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#).

**Table 58. I/O AC characteristics<sup>(1)(2)</sup>**

OSPEEDRy [1:0] bit value <sup>(1)</sup>	Symbol	Parameter	Conditions	Min	Typ	Max	Unit
00	$f_{\text{max(I/O)out}}$	Maximum frequency <sup>(3)</sup>	$C_L = 50 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	4	MHz
			$C_L = 50 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	2	
			$C_L = 10 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	8	
			$C_L = 10 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	4	
	$t_{f(I/O)out}/t_{r(I/O)out}$	Output high to low level fall time and output low to high level rise time	$C_L = 50 \text{ pF}, V_{DD} = 1.8 \text{ V to } 3.6 \text{ V}$	-	-	100	ns

Table 58. I/O AC characteristics<sup>(1)(2)</sup> (continued)

OSPEEDRy [1:0] bit value <sup>(1)</sup>	Symbol	Parameter	Conditions	Min	Typ	Max	Unit
01	$f_{\max(\text{IO})\text{out}}$	Maximum frequency <sup>(3)</sup>	$C_L = 50 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	25	MHz
			$C_L = 50 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	12.5	
			$C_L = 10 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	50	
			$C_L = 10 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	20	
	$t_{f(\text{IO})\text{out}}/$ $t_{r(\text{IO})\text{out}}$	Output high to low level fall time and output low to high level rise time	$C_L = 50 \text{ pF}, V_{DD} \geq 2.7 \text{ V}$	-	-	10	ns
			$C_L = 50 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	20	
			$C_L = 10 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	6	
			$C_L = 10 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	10	
10	$f_{\max(\text{IO})\text{out}}$	Maximum frequency <sup>(3)</sup>	$C_L = 40 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	50 <sup>(4)</sup>	MHz
			$C_L = 40 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	25	
			$C_L = 10 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	100 <sup>(4)</sup>	
			$C_L = 10 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	50 <sup>(4)</sup>	
	$t_{f(\text{IO})\text{out}}/$ $t_{r(\text{IO})\text{out}}$	Output high to low level fall time and output low to high level rise time	$C_L = 40 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	6	ns
			$C_L = 40 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	10	
			$C_L = 10 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	4	
			$C_L = 10 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	6	
11	$f_{\max(\text{IO})\text{out}}$	Maximum frequency <sup>(3)</sup>	$C_L = 30 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	100 <sup>(4)</sup>	MHz
			$C_L = 30 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	50 <sup>(4)</sup>	
			$C_L = 10 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	180 <sup>(4)</sup>	
			$C_L = 10 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	100 <sup>(4)</sup>	
	$t_{f(\text{IO})\text{out}}/$ $t_{r(\text{IO})\text{out}}$	Output high to low level fall time and output low to high level rise time	$C_L = 30 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	4	ns
			$C_L = 30 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	6	
			$C_L = 10 \text{ pF}, V_{DD} \geq 2.70 \text{ V}$	-	-	2.5	
			$C_L = 10 \text{ pF}, V_{DD} \geq 1.8 \text{ V}$	-	-	4	
-	$t_{\text{EXTI}pw}$	Pulse width of external signals detected by the EXTI controller		10	-	-	ns

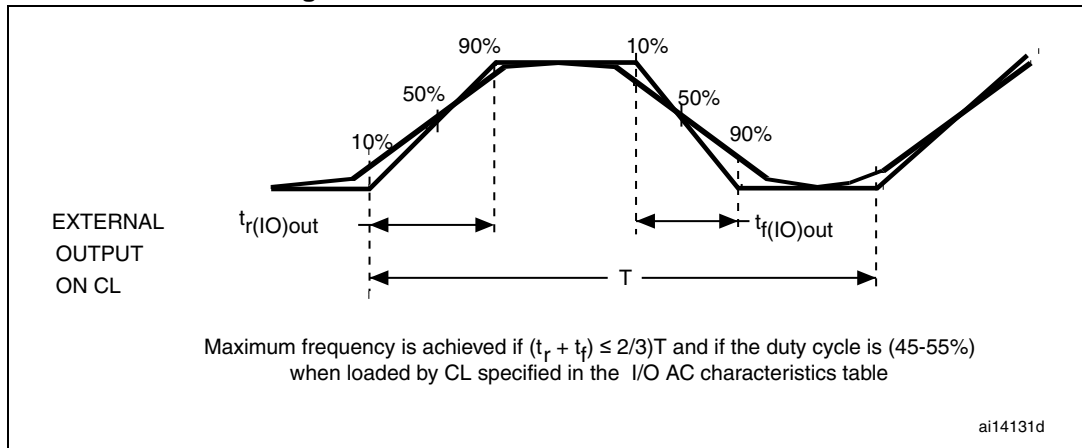
1. Based on characterization data, not tested in production.

2. The I/O speed is configured using the OSPEEDRy[1:0] bits. Refer to the STM32F4xx reference manual for a description of the GPIOx\_SPEEDR GPIO port output speed register.

3. The maximum frequency is defined in [Figure 36](#).

4. For maximum frequencies above 50 MHz and  $V_{DD} > 2.4 \text{ V}$ , the compensation cell should be used.

Figure 36. I/O AC characteristics definition



### 6.3.18 NRST pin characteristics

The NRST pin input driver uses CMOS technology. It is connected to a permanent pull-up resistor,  $R_{PU}$  (see [Table 56: I/O static characteristics](#)).

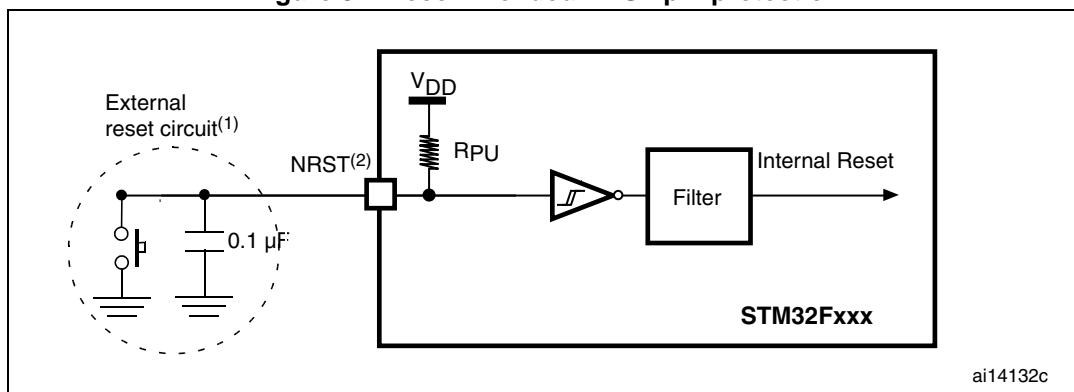
Unless otherwise specified, the parameters given in [Table 59](#) are derived from tests performed under the ambient temperature and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#).

Table 59. NRST pin characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$R_{PU}$	Weak pull-up equivalent resistor <sup>(1)</sup>	$V_{IN} = V_{SS}$	30	40	50	k $\Omega$
$V_{F(NRST)}$ <sup>(2)</sup>	NRST Input filtered pulse		-	-	100	ns
$V_{NF(NRST)}$ <sup>(2)</sup>	NRST Input not filtered pulse	$V_{DD} > 2.7 V$	300	-	-	ns
$T_{NRST\_OUT}$	Generated reset pulse duration	Internal Reset source	20	-	-	$\mu s$

1. The pull-up is designed with a true resistance in series with a switchable PMOS. This PMOS contribution to the series resistance must be minimum (~10% order).
2. Guaranteed by design, not tested in production.

Figure 37. Recommended NRST pin protection



1. The reset network protects the device against parasitic resets.
2. The user must ensure that the level on the NRST pin can go below the  $V_{IL(NRST)}$  max level specified in [Table 59](#). Otherwise the reset is not taken into account by the device.

### 6.3.19 TIM timer characteristics

The parameters given in [Table 60](#) are guaranteed by design.

Refer to [Section 6.3.17: I/O port characteristics](#) for details on the input/output alternate function characteristics (output compare, input capture, external clock, PWM output).

Table 60. TIMx characteristics<sup>(1)(2)</sup>

Symbol	Parameter	Conditions <sup>(3)</sup>	Min	Max	Unit
$t_{res(TIM)}$	Timer resolution time	AHB/APBx prescaler=1 or 2 or 4, $f_{TIMxCLK} = 180$ MHz	1	-	$t_{TIMxCLK}$
		AHB/APBx prescaler>4, $f_{TIMxCLK} = 90$ MHz	1	-	$t_{TIMxCLK}$
$f_{EXT}$	Timer external clock frequency on CH1 to CH4	$f_{TIMxCLK} = 180$ MHz	0	$f_{TIMxCLK}/2$	MHZ
$Res_{TIM}$	Timer resolution		-	16/32	bit
$t_{MAX\_COUNT}$	Maximum possible count with 32-bit counter		-	$65536 \times 65536$	$t_{TIMxCLK}$

1. TIMx is used as a general term to refer to the TIM1 to TIM12 timers.
2. Guaranteed by design, not tested in production.
3. The maximum timer frequency on APB1 or APB2 is up to 180 MHz, by setting the TIMPRE bit in the RCC\_DCKCFGR register, if APBx prescaler is 1 or 2 or 4, then  $TIMxCLK = HCKL$ , otherwise  $TIMxCLK = 4x PCLKx$ .

### 6.3.20 Communications interfaces

#### I<sup>2</sup>C interface characteristics

The I<sup>2</sup>C interface meets the requirements of the standard I<sup>2</sup>C communication protocol with the following restrictions: the I/O pins SDA and SCL are mapped to are not “true” open-drain. When configured as open-drain, the PMOS connected between the I/O pin and  $V_{DD}$  is disabled, but is still present.

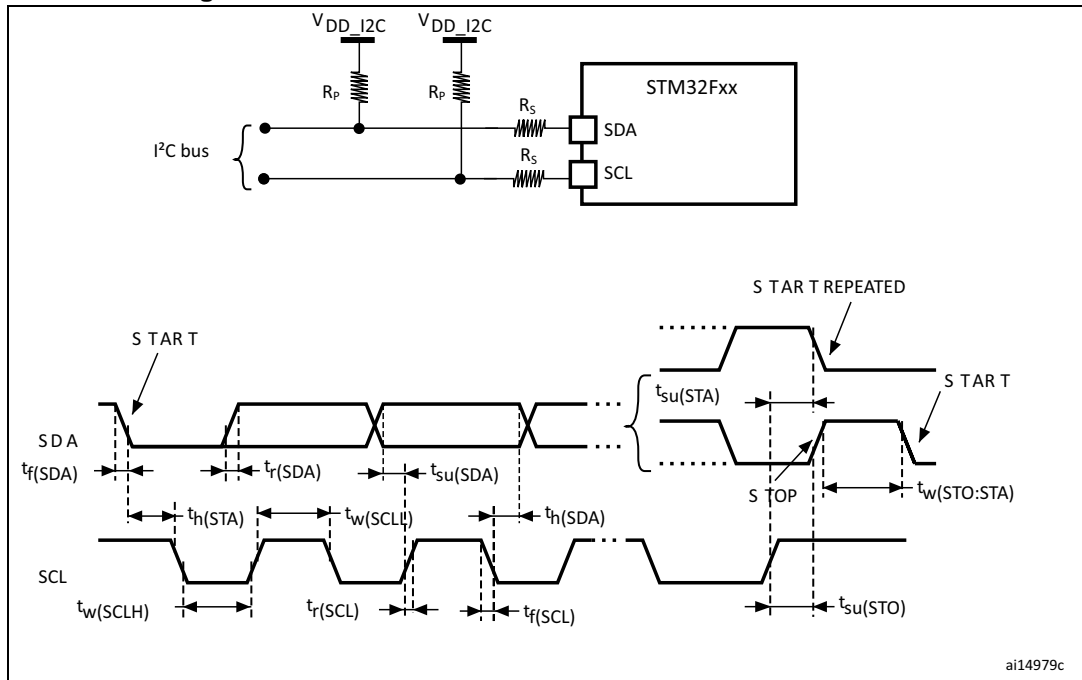
The I<sup>2</sup>C characteristics are described in [Table 61](#). Refer also to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output alternate function characteristics (SDA and SCL).

Table 61. I<sup>2</sup>C characteristics

Symbol	Parameter	Standard mode I <sup>2</sup> C <sup>(1)(2)</sup>		Fast mode I <sup>2</sup> C <sup>(1)(2)</sup>		Unit
		Min	Max	Min	Max	
t <sub>w(SCLL)</sub>	SCL clock low time	4.7	-	1.3	-	μs
t <sub>w(SCLH)</sub>	SCL clock high time	4.0	-	0.6	-	
t <sub>su(SDA)</sub>	SDA setup time	250	-	100	-	ns
t <sub>h(SDA)</sub>	SDA data hold time	-	3450 <sup>(3)</sup>	-	900 <sup>(4)</sup>	
t <sub>r(SDA)</sub> t <sub>r(SCL)</sub>	SDA and SCL rise time	-	1000	-	300	
t <sub>f(SDA)</sub> t <sub>f(SCL)</sub>	SDA and SCL fall time	-	300	-	300	μs
t <sub>h(STA)</sub>	Start condition hold time	4.0	-	0.6	-	
t <sub>su(STA)</sub>	Repeated Start condition setup time	4.7	-	0.6	-	μs
t <sub>su(STO)</sub>	Stop condition setup time	4.0	-	0.6	-	μs
t <sub>w(STO:STA)</sub>	Stop to Start condition time (bus free)	4.7	-	1.3	-	μs
t <sub>SP</sub>	Pulse width of the spikes that are suppressed by the analog filter for standard and fast mode	0	50 <sup>(5)</sup>	0	50 <sup>(5)</sup>	μs
C <sub>b</sub>	Capacitive load for each bus line	-	400	-	400	pF

1. Guaranteed by design, not tested in production.
2. f<sub>CLK1</sub> must be at least 2 MHz to achieve standard mode I<sup>2</sup>C frequencies. It must be at least 4 MHz to achieve fast mode I<sup>2</sup>C frequencies, and a multiple of 10 MHz to reach the 400 kHz maximum I<sup>2</sup>C fast mode clock.
3. The device must internally provide a hold time of at least 300 ns for the SDA signal in order to bridge the undefined region of the falling edge of SCL.
4. The maximum data hold time has only to be met if the interface does not stretch the low period of SCL signal.
5. The minimum width of the spikes filtered by the analog filter is above t<sub>SP(max)</sub>.

Figure 38. I<sup>2</sup>C bus AC waveforms and measurement circuit



1.  $R_S$  = series protection resistor.
2.  $R_P$  = external pull-up resistor.
3.  $V_{DD\_I2C}$  is the I2C bus power supply.

Table 62. SCL frequency ( $f_{PCLK1} = 42\text{ MHz}, V_{DD} = V_{DD\_I2C} = 3.3\text{ V}$ )<sup>(1)(2)</sup>

$f_{SCL}$ (kHz)	I2C_CCR value
	$R_P = 4.7\text{ k}\Omega$
400	0x8019
300	0x8021
200	0x8032
100	0x0096
50	0x012C
20	0x02EE

1.  $R_P$  = External pull-up resistance,  $f_{SCL}$  = I<sup>2</sup>C speed,
2. For speeds around 200 kHz, the tolerance on the achieved speed is of  $\pm 5\%$ . For other speed ranges, the tolerance on the achieved speed  $\pm 2\%$ . These variations depend on the accuracy of the external components used to design the application.

**SPI interface characteristics**

Unless otherwise specified, the parameters given in [Table 63](#) for the SPI interface are derived from tests performed under the ambient temperature,  $f_{PCLKx}$  frequency and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#), with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels:  $0.5V_{DD}$

Refer to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output alternate function characteristics (NSS, SCK, MOSI, MISO for SPI).

**Table 63. SPI dynamic characteristics<sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$f_{SCK}$ $1/t_{c(SCK)}$	SPI clock frequency	Master mode, SPI1/4/5/6, $2.7 V \leq V_{DD} \leq 3.6 V$	-	-	45	MHz
		Slave mode, SPI1/4/5/6, $2.7 V \leq V_{DD} \leq 3.6 V$			Receiver	
			Transmitter/ full-duplex	38 <sup>(2)</sup>		
		Master mode, SPI1/2/3/4/5/6, $1.7 V \leq V_{SS} \leq 3.6 V$	-	-	22.5	
	Slave mode, SPI1/2/3/4/5/6, $1.7 V \leq V_{SS} \leq 3.6 V$	-	-	22.5		
Duty(SCK)	Duty cycle of SPI clock frequency	Slave mode	30	50	70	%
$t_{w(SCKH)}$	SCK high and low time	Master mode, SPI presc = 2, $2.7 V \leq V_{DD} \leq 3.6 V$	$T_{PCLK} - 0.5$	$T_{PCLK}$	$T_{PCLK} + 0.5$	ns
$t_{w(SCKL)}$		Master mode, SPI presc = 2, $1.7 V \leq V_{SS} \leq 3.6 V$	$T_{PCLK} - 2$	$T_{PCLK}$	$T_{PCLK} + 2$	ns
$t_{su(NSS)}$	NSS setup time	Slave mode, SPI presc = 2	$4T_{PCLK}$	-	-	ns
$t_{h(NSS)}$	NSS hold time	Slave mode, SPI presc = 2	$2T_{PCLK}$	-	-	ns
$t_{su(MI)}$	Data input setup time	Master mode	3	-	-	ns
$t_{su(SI)}$		Slave mode	0	-	-	ns
$t_{h(MI)}$	Data input hold time	Master mode	0.5	-	-	ns
$t_{h(SI)}$		Slave mode	2	-	-	ns
$t_{a(SO)}^{(3)}$	Data output access time	Slave mode, SPI presc = 2	0	-	$4T_{PCLK}$	ns
$t_{dis(SO)}^{(4)}$	Data output disable time	Slave mode, SPI1/4/5/6, $2.7 V \leq V_{DD} \leq 3.6 V$	0	-	8.5	ns
		Slave mode, SPI1/2/3/4/5/6 and $1.7 V \leq V_{SS} \leq 3.6 V$	0	-	16.5	ns

Table 63. SPI dynamic characteristics<sup>(1)</sup> (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{v(SO)}$ $t_{h(SO)}$	Data output valid/hold time	Slave mode (after enable edge), SPI1/4/5/6 and $2.7V \leq V_{DD} \leq 3.6V$	-	11	13	ns
		Slave mode (after enable edge), SPI2/3, $2.7V \leq V_{DD} \leq 3.6V$	-	14	15	ns
		Slave mode (after enable edge), SPI1/4/5/6, $1.7V \leq V_{SS} \leq 3.6V$	-	15.5	19	ns
		Slave mode (after enable edge), SPI2/3, $1.7V \leq V_{SS} \leq 3.6V$	-	15.5	17.5	ns
$t_{v(MO)}$	Data output valid time	Master mode (after enable edge), SPI1/4/5/6, $2.7V \leq V_{DD} \leq 3.6V$	-	-	2.5	ns
		Master mode (after enable edge), SPI1/2/3/4/5/6, $1.7V \leq V_{SS} \leq 3.6V$	-	-	4.5	ns
$t_{h(MO)}$	Data output hold time	Master mode (after enable edge)	0	-	-	ns

1. Data based on characterization results, not tested in production.
2. Maximum frequency in Slave transmitter mode is determined by the sum of  $t_{v(SO)}$  and  $t_{su(MI)}$  which has to fit into SCK low or high phase preceding the SCK sampling edge. This value can be achieved when the SPI communicates with a master having  $t_{su(MI)} = 0$  while Duty(SCK) = 50%
3. The minimum value of this timing corresponds to the minimum time to drive the output, and the maximum value to maximum time to validate the data.
4. The minimum value of this timing corresponds to the minimum time to invalidate the output and the maximum value to the maximum time to put the data in Hi-Z.

Figure 39. SPI timing diagram - slave mode and CPHA = 0

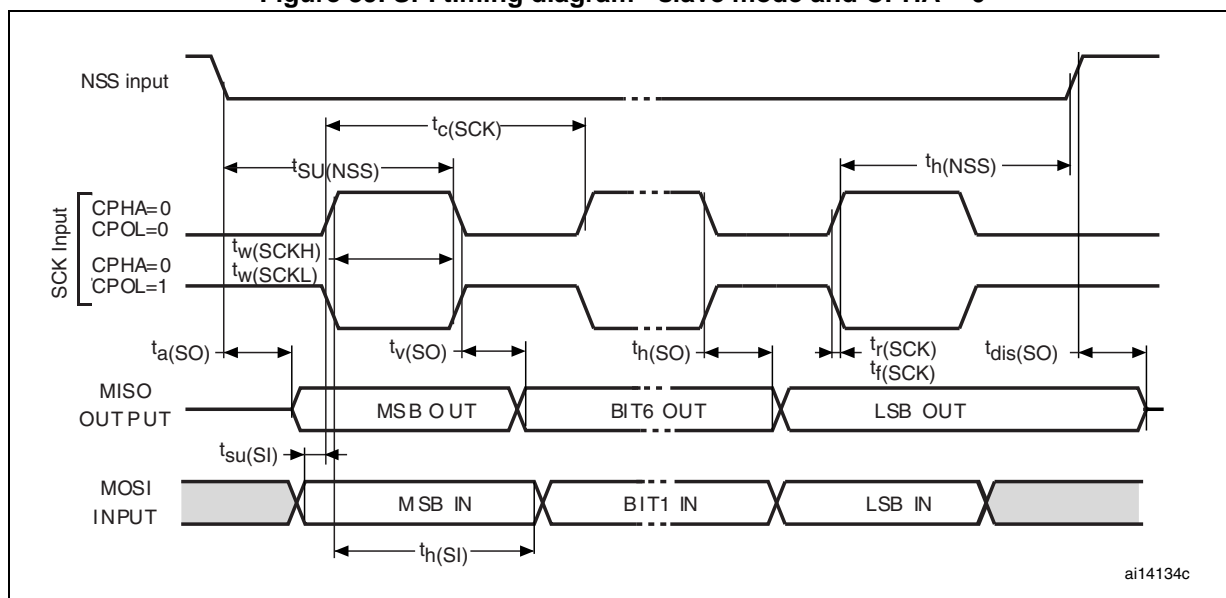


Figure 40. SPI timing diagram - slave mode and CPHA = 1<sup>(1)</sup>

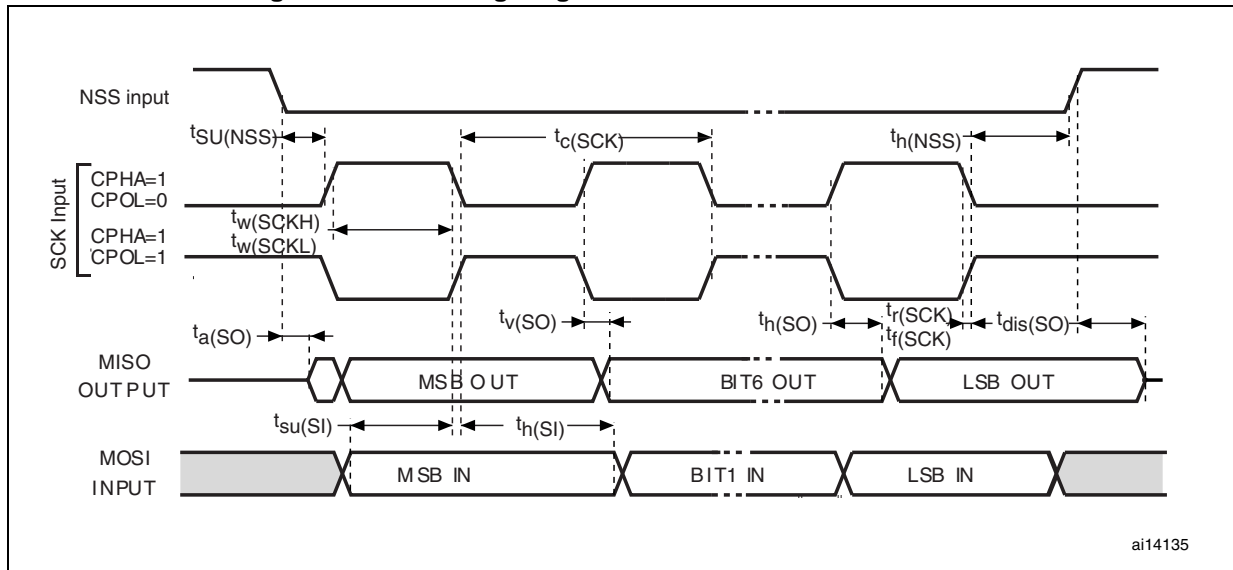
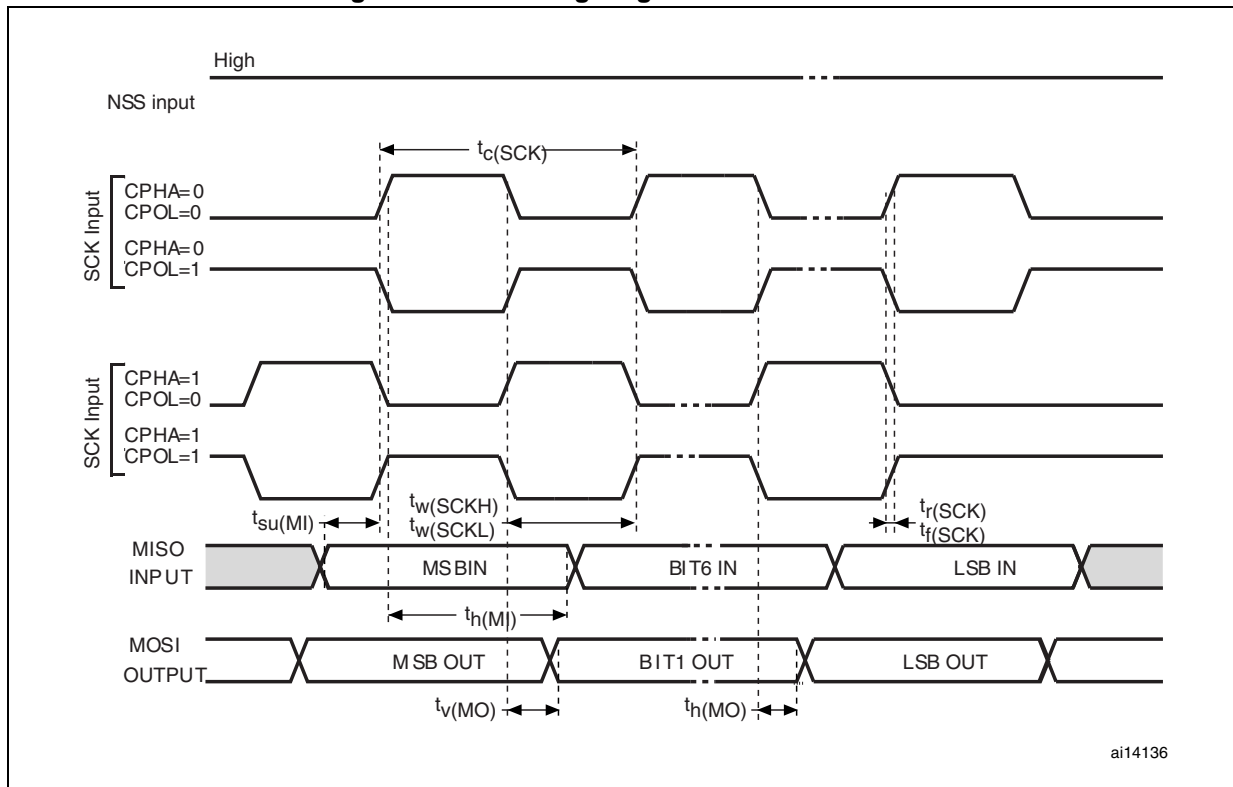


Figure 41. SPI timing diagram - master mode<sup>(1)</sup>



### I<sup>2</sup>S interface characteristics

Unless otherwise specified, the parameters given in [Table 64](#) for the I<sup>2</sup>S interface are derived from tests performed under the ambient temperature,  $f_{PCLKX}$  frequency and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#), with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C = 30 pF
- Measurement points are done at CMOS levels: 0.5V<sub>DD</sub>

Refer to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output alternate function characteristics (CK, SD, WS).

**Table 64. I<sup>2</sup>S dynamic characteristics<sup>(1)</sup>**

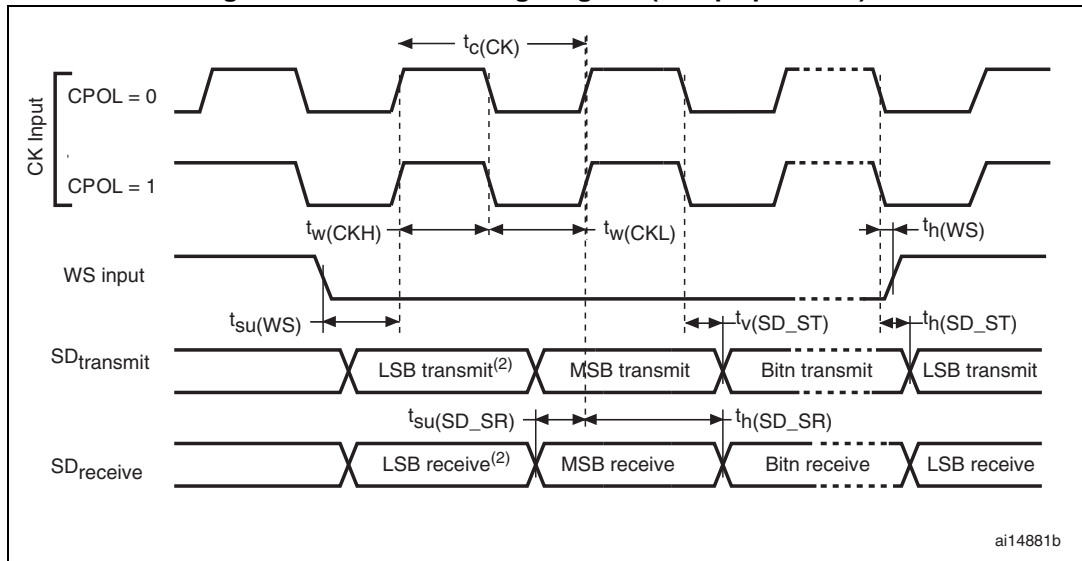
Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>MCK</sub>	I2S Main clock output	-	256x8K	256xFs <sup>(2)</sup>	MHz
f <sub>CK</sub>	I2S clock frequency	Master data: 32 bits	-	64xFs	MHz
		Slave data: 32 bits	-	64xFs	
D <sub>CK</sub>	I2S clock frequency duty cycle	Slave receiver	30	70	%
t <sub>v(WS)</sub>	WS valid time	Master mode	0	6	ns
t <sub>h(WS)</sub>	WS hold time	Master mode	0	-	
t <sub>su(WS)</sub>	WS setup time	Slave mode	1	-	
t <sub>h(WS)</sub>	WS hold time	Slave mode	0	-	
t <sub>su(SD_MR)</sub>	Data input setup time	Master receiver	7.5	-	
t <sub>su(SD_SR)</sub>		Slave receiver	2	-	
t <sub>h(SD_MR)</sub>	Data input hold time	Master receiver	0	-	
t <sub>h(SD_SR)</sub>		Slave receiver	0	-	
t <sub>v(SD_ST)</sub>	Data output valid time	Slave transmitter (after enable edge)	-	27	
t <sub>h(SD_ST)</sub>		Master transmitter (after enable edge)	-	20	
t <sub>v(SD_MT)</sub>			-	20	
t <sub>h(SD_MT)</sub>	Data output hold time	Master transmitter (after enable edge)	2.5	-	

1. Data based on characterization results, not tested in production.
2. The maximum value of 256xFs is 45 MHz (APB1 maximum frequency).

**Note:** Refer to the I2S section of RM0090 reference manual for more details on the sampling frequency (F<sub>S</sub>).

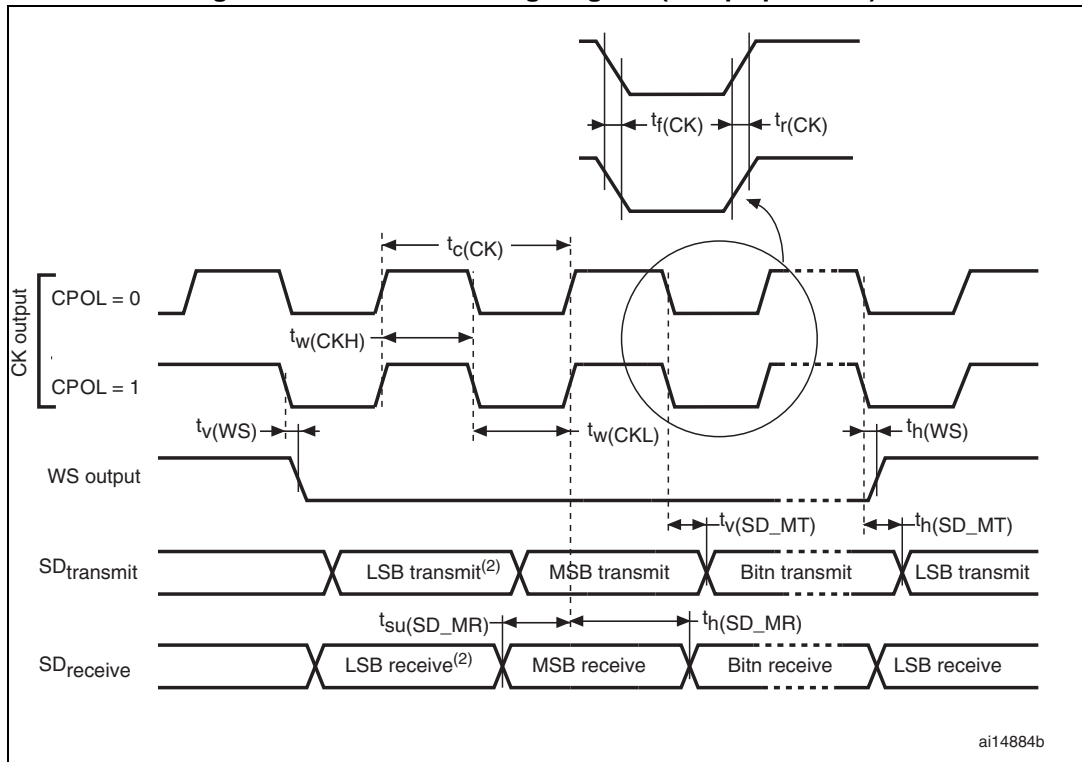
f<sub>MCK</sub>, f<sub>CK</sub>, and D<sub>CK</sub> values reflect only the digital peripheral behavior. The values of these parameters might be slightly impacted by the source clock precision. D<sub>CK</sub> depends mainly on the value of ODD bit. The digital contribution leads to a minimum value of (I2SDIV/(2\*I2SDIV+ODD)) and a maximum value of (I2SDIV+ODD)/(2\*I2SDIV+ODD). F<sub>S</sub> maximum value is supported for each mode/condition.

Figure 42. I<sup>2</sup>S slave timing diagram (Philips protocol)<sup>(1)</sup>



1. .LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

Figure 43. I<sup>2</sup>S master timing diagram (Philips protocol)<sup>(1)</sup>



1. LSB transmit/receive of the previously transmitted byte. No LSB transmit/receive is sent before the first byte.

**SAI characteristics**

Unless otherwise specified, the parameters given in [Table 65](#) for SAI are derived from tests performed under the ambient temperature,  $f_{PCLKx}$  frequency and VDD supply voltage conditions summarized in [Table 17](#), with the following configuration:

- Output speed is set to OSPEEDRy[1:0] = 10
- Capacitive load C=30 pF
- Measurement points are performed at CMOS levels: 0.5V<sub>DD</sub>

Refer to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output alternate function characteristics (SCK,SD,WS).

**Table 65. SAI characteristics<sup>(1)</sup>**

Symbol	Parameter	Conditions	Min	Max	Unit
f <sub>MCKL</sub>	SAI Main clock output	-	256 x 8K	256xFs <sup>(2)</sup>	MHz
F <sub>SCK</sub>	SAI clock frequency	Master data: 32 bits	-	64xFs	MHz
		Slave data: 32 bits	-	64xFs	
D <sub>SCK</sub>	SAI clock frequency duty cycle	Slave receiver	30	70	%
t <sub>v(FS)</sub>	FS valid time	Master mode	8	22	ns
t <sub>su(FS)</sub>	FS setup time	Slave mode	2	-	
t <sub>h(FS)</sub>	FS hold time	Master mode	8	-	
		Slave mode	0	-	
t <sub>su(SD_MR)</sub>	Data input setup time	Master receiver	5	-	
t <sub>su(SD_SR)</sub>		Slave receiver	3	-	
t <sub>h(SD_MR)</sub>	Data input hold time	Master receiver	0	-	
t <sub>h(SD_SR)</sub>		Slave receiver	0	-	
t <sub>v(SD_ST)</sub>	Data output valid time	Slave transmitter (after enable edge)	-	22	
t <sub>h(SD_ST)</sub>		Master transmitter (after enable edge)	-	20	
t <sub>v(SD_MT)</sub>	Data output hold time	Master transmitter (after enable edge)	8	-	
t <sub>h(SD_MT)</sub>					

1. Data based on characterization results, not tested in production.
2. 256xFs maximum corresponds to 45 MHz (APB2 xaximum frequency)

Figure 44. SAI master timing waveforms

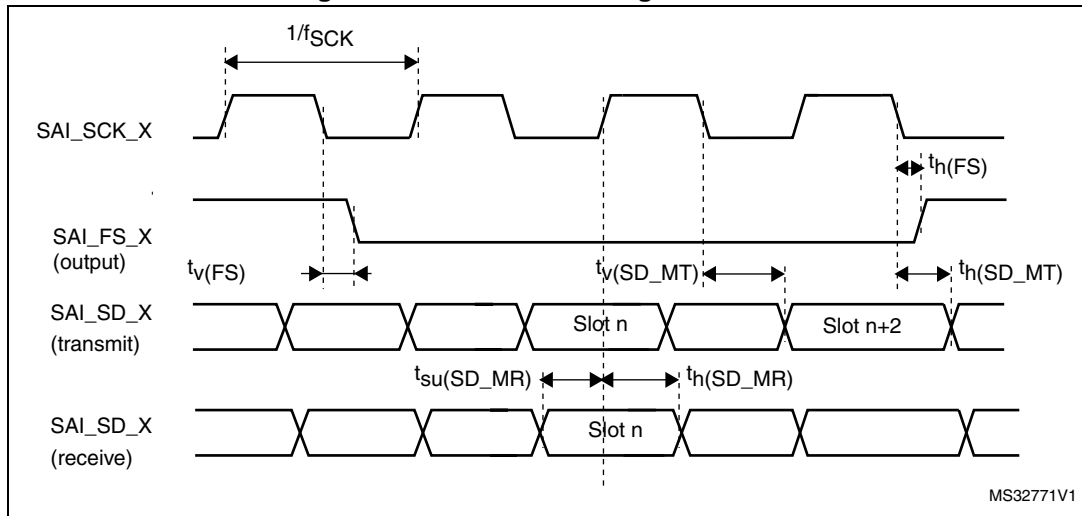
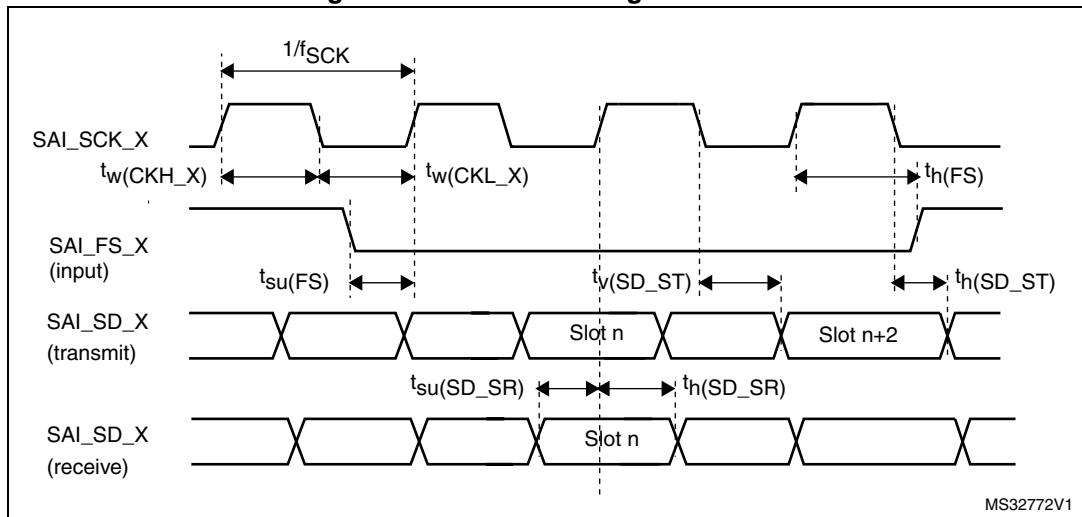


Figure 45. SAI slave timing waveforms



**USB OTG full speed (FS) characteristics**

This interface is present in both the USB OTG HS and USB OTG FS controllers.

**Table 66. USB OTG full speed startup time**

Symbol	Parameter	Max	Unit
$t_{STARTUP}^{(1)}$	USB OTG full speed transceiver startup time	1	$\mu s$

1. Guaranteed by design, not tested in production.

**Table 67. USB OTG full speed DC electrical characteristics**

Symbol	Parameter	Conditions	Min. <sup>(1)</sup>	Typ.	Max. <sup>(1)</sup>	Unit	
Input levels	$V_{DD}$	USB OTG full speed transceiver operating voltage	3.0 <sup>(2)</sup>	-	3.6	V	
	$V_{DI}^{(3)}$	Differential input sensitivity	I(USB_FS_DP/DM, USB_HS_DP/DM)	0.2	-	-	V
	$V_{CM}^{(3)}$	Differential common mode range	Includes $V_{DI}$ range	0.8	-	2.5	
	$V_{SE}^{(3)}$	Single ended receiver threshold		1.3	-	2.0	
Output levels	$V_{OL}$	Static output level low	$R_L$ of 1.5 k $\Omega$ to 3.6 V <sup>(4)</sup>	-	-	0.3	V
	$V_{OH}$	Static output level high	$R_L$ of 15 k $\Omega$ to $V_{SS}^{(4)}$	2.8	-	3.6	
$R_{PD}$	PA11, PA12, PB14, PB15 (USB_FS_DP/DM, USB_HS_DP/DM)	$V_{IN} = V_{DD}$	17	21	24	k $\Omega$	
	PA9, PB13 (OTG_FS_VBUS, OTG_HS_VBUS)		0.65	1.1	2.0		
$R_{PU}$	PA12, PB15 (USB_FS_DP, USB_HS_DP)	$V_{IN} = V_{SS}$	1.5	1.8	2.1		
	PA9, PB13 (OTG_FS_VBUS, OTG_HS_VBUS)	$V_{IN} = V_{SS}$	0.25	0.37	0.55		

1. All the voltages are measured from the local ground potential.
2. The USB OTG full speed transceiver functionality is ensured down to 2.7 V but not the full USB full speed electrical characteristics which are degraded in the 2.7-to-3.0 V  $V_{DD}$  voltage range.
3. Guaranteed by design, not tested in production.
4.  $R_L$  is the load connected on the USB OTG full speed drivers.

*Note: When VBUS sensing feature is enabled, PA9 and PB13 should be left at their default state (floating input), not as alternate function. A typical 200  $\mu A$  current consumption of the sensing block (current to voltage conversion to determine the different sessions) can be observed on PA9 and PB13 when the feature is enabled.*

Figure 46. USB OTG full speed timings: definition of data signal rise and fall time

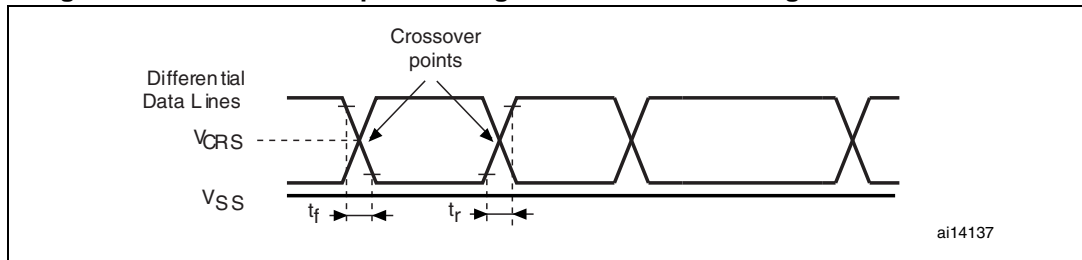


Table 68. USB OTG full speed electrical characteristics<sup>(1)</sup>

Driver characteristics					
Symbol	Parameter	Conditions	Min	Max	Unit
$t_r$	Rise time <sup>(2)</sup>	$C_L = 50 \text{ pF}$	4	20	ns
$t_f$	Fall time <sup>(2)</sup>	$C_L = 50 \text{ pF}$	4	20	ns
$t_{rfm}$	Rise/ fall time matching	$t_r/t_f$	90	110	%
$V_{CRS}$	Output signal crossover voltage		1.3	2.0	V
$Z_{DRV}$	Output driver impedance <sup>(3)</sup>	Driving high or low	28	44	$\Omega$

1. Guaranteed by design, not tested in production.
2. Measured from 10% to 90% of the data signal. For more detailed informations, please refer to USB Specification - Chapter 7 (version 2.0).
3. No external termination series resistors are required on DP (D+) and DM (D-) pins since the matching impedance is included in the embedded driver.

**USB high speed (HS) characteristics**

Unless otherwise specified, the parameters given in [Table 71](#) for ULPI are derived from tests performed under the ambient temperature,  $f_{HCLK}$  frequency summarized in [Table 70](#) and  $V_{DD}$  supply voltage conditions summarized in [Table 69](#), with the following configuration:

- Output speed is set to  $OSPEEDRy[1:0] = 10$ , unless otherwise specified
- Capacitive load  $C = 30 \text{ pF}$ , unless otherwise specified
- Measurement points are done at CMOS levels:  $0.5V_{DD}$ .

Refer to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output characteristics.

Table 69. USB HS DC electrical characteristics

Symbol	Parameter	Min. <sup>(1)</sup>	Max. <sup>(1)</sup>	Unit
Input level $V_{DD}$	USB OTG HS operating voltage	1.7	3.6	V

1. All the voltages are measured from the local ground potential.

Table 70. USB HS clock timing parameters<sup>(1)</sup>

Symbol	Parameter		Min	Typ	Max	Unit
	f <sub>HCLK</sub> value to guarantee proper operation of USB HS interface		30	-	-	MHz
F <sub>START_8BIT</sub>	Frequency (first transition)	8-bit ±10%	54	60	66	MHz
F <sub>STEADY</sub>	Frequency (steady state) ±500 ppm		59.97	60	60.03	MHz
D <sub>START_8BIT</sub>	Duty cycle (first transition)	8-bit ±10%	40	50	60	%
D <sub>STEADY</sub>	Duty cycle (steady state) ±500 ppm		49.975	50	50.025	%
t <sub>STEADY</sub>	Time to reach the steady state frequency and duty cycle after the first transition		-	-	1.4	ms
t <sub>START_DEV</sub>	Clock startup time after the de-assertion of SuspendM	Peripheral	-	-	5.6	ms
t <sub>START_HOST</sub>		Host	-	-	-	
t <sub>PREP</sub>	PHY preparation time after the first transition of the input clock		-	-	-	µs

1. Guaranteed by design, not tested in production.

Figure 47. ULPI timing diagram

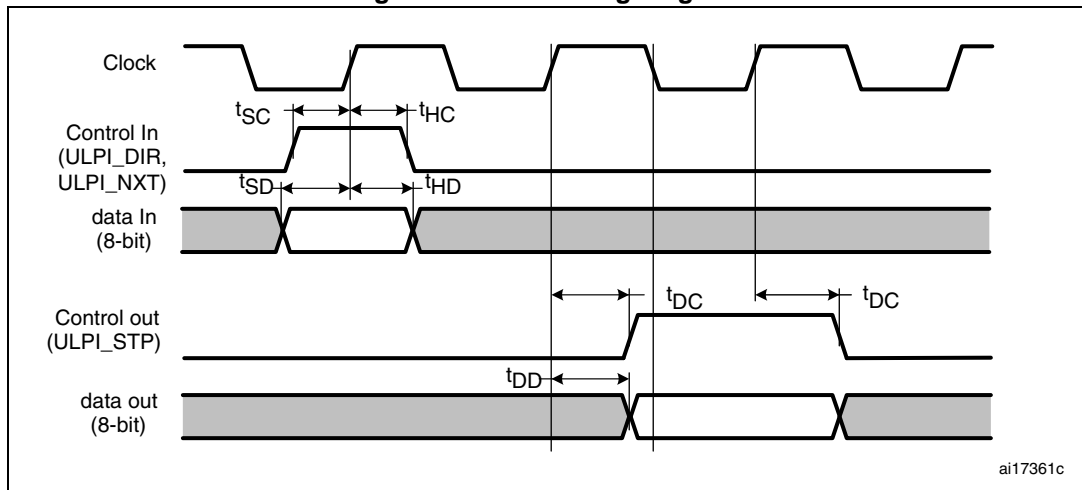


Table 71. Dynamic characteristics: USB ULPI<sup>(1)</sup>

Symbol	Parameter	Conditions	Min.	Typ.	Max.	Unit
t <sub>SC</sub>	Control in (ULPI_DIR, ULPI_NXT) setup time		2	-	-	ns
t <sub>HC</sub>	Control in (ULPI_DIR, ULPI_NXT) hold time		0.5	-	-	
t <sub>SD</sub>	Data in setup time		1.5	-	-	
t <sub>HD</sub>	Data in hold time		2	-	-	
t <sub>DC</sub> /t <sub>DD</sub>	Data/control output delay	2.7 V < V <sub>DD</sub> < 3.6 V, C <sub>L</sub> = 15 pF and OSPEEDRy[1:0] = 11	-	9	9.5	
		2.7 V < V <sub>DD</sub> < 3.6 V, C <sub>L</sub> = 20 pF and OSPEEDRy[1:0] = 10	-	12	15	
		1.7 V < V <sub>DD</sub> < 3.6 V, C <sub>L</sub> = 15 pF and OSPEEDRy[1:0] = 11	-			

1. Data based on characterization results, not tested in production.

**Ethernet characteristics**

Unless otherwise specified, the parameters given in [Table 73](#), [Table 74](#) and [Table 75](#) for SMI, RMII and MII are derived from tests performed under the ambient temperature,  $f_{HCLK}$  frequency summarized in [Table 17](#) and  $V_{DD}$  supply voltage conditions summarized in [Table 72](#), with the following configuration:

- Output speed is set to  $OSPEEDRy[1:0] = 10$
- Capacitive load  $C = 30 \text{ pF}$
- Measurement points are done at CMOS levels:  $0.5V_{DD}$ .

Refer to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output characteristics.

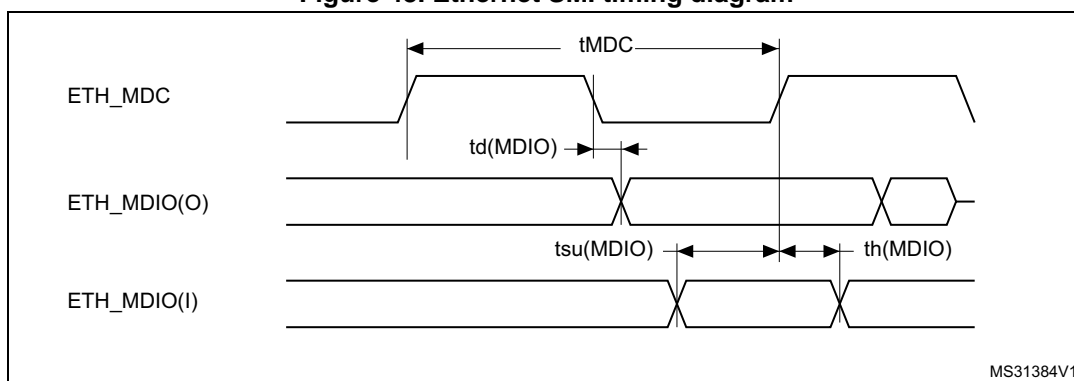
**Table 72. Ethernet DC electrical characteristics**

Symbol		Parameter	Min. <sup>(1)</sup>	Max. <sup>(1)</sup>	Unit
Input level	$V_{DD}$	Ethernet operating voltage	2.7	3.6	V

1. All the voltages are measured from the local ground potential.

[Table 73](#) gives the list of Ethernet MAC signals for the SMI (station management interface) and [Figure 48](#) shows the corresponding timing diagram.

**Figure 48. Ethernet SMI timing diagram**



MS31384V1

**Table 73. Dynamics characteristics: Ethernet MAC signals for SMI<sup>(1)</sup>**

Symbol	Parameter	Min	Typ	Max	Unit
$t_{MDC}$	MDC cycle time(2.38 MHz)	411	420	425	ns
$T_d(MDIO)$	Write data valid time	6	10	13	
$t_{su}(MDIO)$	Read data setup time	12	-	-	
$t_h(MDIO)$	Read data hold time	0	-	-	

1. Data based on characterization results, not tested in production.

[Table 74](#) gives the list of Ethernet MAC signals for the RMII and [Figure 49](#) shows the corresponding timing diagram.

Figure 49. Ethernet RMII timing diagram

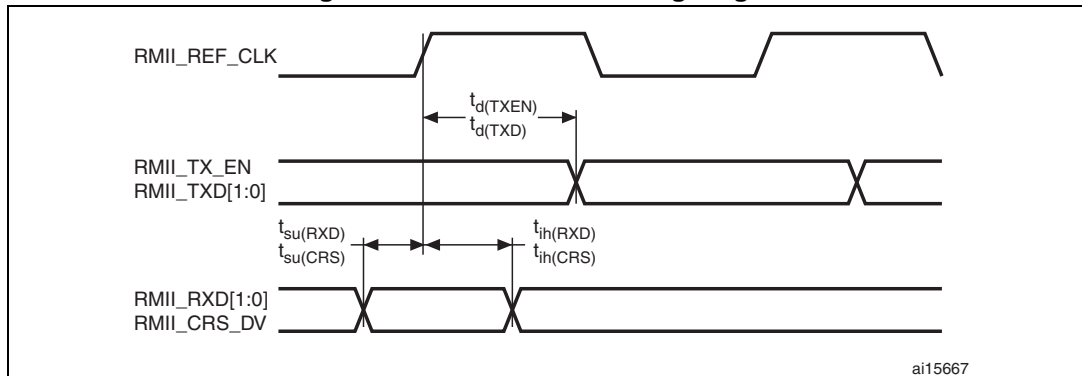


Table 74. Dynamics characteristics: Ethernet MAC signals for RMII<sup>(1)</sup>

Symbol	Parameter	Min	Typ	Max	Unit
$t_{su(RXD)}$	Receive data setup time	1.5	-	-	ns
$t_{th(RXD)}$	Receive data hold time	0	-	-	
$t_{su(CRS)}$	Carrier sense setup time	1	-	-	
$t_{th(CRS)}$	Carrier sense hold time	1	-	-	
$t_{d(TXEN)}$	Transmit enable valid delay time	0	10.5	12	
$t_{d(TXD)}$	Transmit data valid delay time	0	11	12.5	

1. Data based on characterization results, not tested in production.

Table 75 gives the list of Ethernet MAC signals for MII and Figure 49 shows the corresponding timing diagram.

Figure 50. Ethernet MII timing diagram

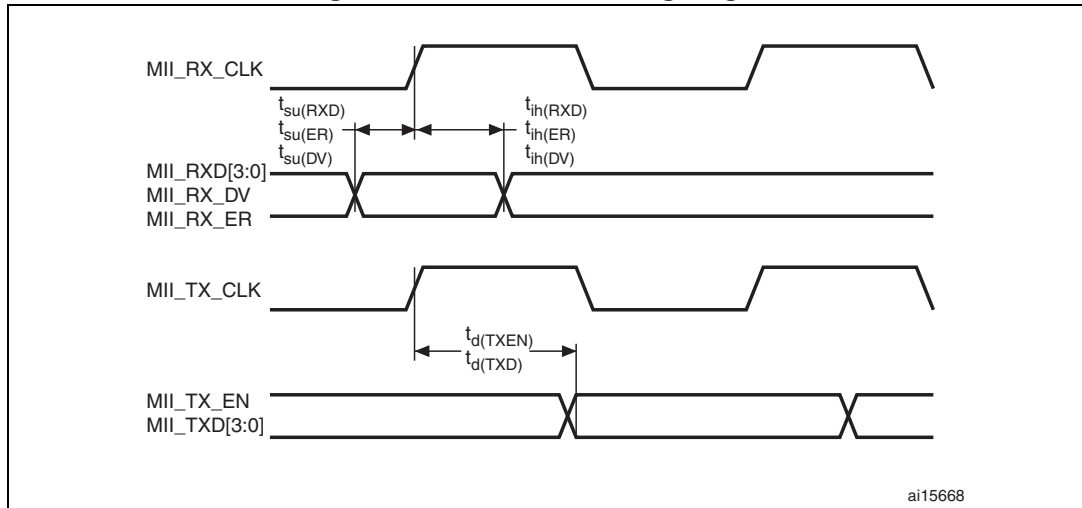


Table 75. Dynamics characteristics: Ethernet MAC signals for MII<sup>(1)</sup>

Symbol	Parameter	Min	Typ	Max	Unit
$t_{su(RXD)}$	Receive data setup time	9		-	ns
$t_{ih(RXD)}$	Receive data hold time	10		-	
$t_{su(DV)}$	Data valid setup time	9		-	
$t_{ih(DV)}$	Data valid hold time	8		-	
$t_{su(ER)}$	Error setup time	6		-	
$t_{ih(ER)}$	Error hold time	8		-	
$t_d(TXEN)$	Transmit enable valid delay time	0	10	14	
$t_d(TXD)$	Transmit data valid delay time	0	10	15	

1. Data based on characterization results, not tested in production.

### CAN (controller area network) interface

Refer to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output alternate function characteristics (CANx\_TX and CANx\_RX).

#### 6.3.21 12-bit ADC characteristics

Unless otherwise specified, the parameters given in [Table 76](#) are derived from tests performed under the ambient temperature,  $f_{PCLK2}$  frequency and  $V_{DDA}$  supply voltage conditions summarized in [Table 17](#).

Table 76. ADC characteristics

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{DDA}$	Power supply	$V_{DDA} - V_{REF+} < 1.2\text{ V}$	1.8 <sup>(1)</sup>	-	3.6	V
$V_{REF+}$	Positive reference voltage		1.8 <sup>(1)</sup>	-	$V_{DDA}$	V
$f_{ADC}$	ADC clock frequency	$V_{DDA} = 1.8^{(1)}\text{ to }2.4\text{ V}$	0.6	15	18	MHz
		$V_{DDA} = 2.4\text{ to }3.6\text{ V}$	0.6	30	36	MHz
$f_{TRIG}^{(2)}$	External trigger frequency	$f_{ADC} = 30\text{ MHz}$ , 12-bit resolution	-	-	1764	kHz
			-	-	17	$1/f_{ADC}$
$V_{AIN}$	Conversion voltage range <sup>(3)</sup>		0 ( $V_{SSA}$ or $V_{REF-}$ tied to ground)	-	$V_{REF+}$	V
$R_{AIN}^{(2)}$	External input impedance	See <a href="#">Equation 1</a> for details	-	-	50	k $\Omega$
$R_{ADC}^{(2)(4)}$	Sampling switch resistance		-	-	6	k $\Omega$
$C_{ADC}^{(2)}$	Internal sample and hold capacitor		-	4	7	pF
$t_{lat}^{(2)}$	Injection trigger conversion latency	$f_{ADC} = 30\text{ MHz}$	-	-	0.100	$\mu\text{s}$
			-	-	3 <sup>(5)</sup>	$1/f_{ADC}$

Table 76. ADC characteristics (continued)

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$t_{latr}^{(2)}$	Regular trigger conversion latency	$f_{ADC} = 30\text{ MHz}$	-	-	0.067	$\mu\text{s}$
			-	-	$2^{(5)}$	$1/f_{ADC}$
$t_S^{(2)}$	Sampling time	$f_{ADC} = 30\text{ MHz}$	0.100	-	16	$\mu\text{s}$
			3	-	480	$1/f_{ADC}$
$t_{STAB}^{(2)}$	Power-up time		-	2	3	$\mu\text{s}$
$t_{CONV}^{(2)}$	Total conversion time (including sampling time)	$f_{ADC} = 30\text{ MHz}$ 12-bit resolution	0.50	-	16.40	$\mu\text{s}$
		$f_{ADC} = 30\text{ MHz}$ 10-bit resolution	0.43	-	16.34	$\mu\text{s}$
		$f_{ADC} = 30\text{ MHz}$ 8-bit resolution	0.37	-	16.27	$\mu\text{s}$
		$f_{ADC} = 30\text{ MHz}$ 6-bit resolution	0.30	-	16.20	$\mu\text{s}$
		9 to 492 ( $t_S$ for sampling + n-bit resolution for successive approximation)				
$f_S^{(2)}$	Sampling rate ( $f_{ADC} = 30\text{ MHz}$ , and $t_S = 3\text{ ADC cycles}$ )	12-bit resolution Single ADC	-	-	2	MspS
		12-bit resolution Interleave Dual ADC mode	-	-	3.75	MspS
		12-bit resolution Interleave Triple ADC mode	-	-	6	MspS
$I_{VREF+}^{(2)}$	ADC $V_{REF}$ DC current consumption in conversion mode		-	300	500	$\mu\text{A}$
$I_{VDDA}^{(2)}$	ADC $V_{DDA}$ DC current consumption in conversion mode		-	1.6	1.8	mA

- $V_{DDA}$  minimum value of 1.7 V is obtained with the use of an external power supply supervisor (refer to [Section 3.17.2: Internal reset OFF](#)).
- Based on characterization, not tested in production.
- $V_{REF+}$  is internally connected to  $V_{DDA}$  and  $V_{REF-}$  is internally connected to  $V_{SSA}$ .
- $R_{ADC}$  maximum value is given for  $V_{DD}=1.8\text{ V}$ , and minimum value for  $V_{DD}=3.3\text{ V}$ .
- For external triggers, a delay of  $1/f_{PCLK2}$  must be added to the latency specified in [Table 76](#).

Equation 1:  $R_{AIN}$  max formula

$$R_{AIN} = \frac{(k - 0.5)}{f_{ADC} \times C_{ADC} \times \ln(2^{N+2})} - R_{ADC}$$

The formula above (*Equation 1*) is used to determine the maximum external impedance allowed for an error below 1/4 of LSB. N = 12 (from 12-bit resolution) and k is the number of sampling periods defined in the ADC\_SMPR1 register.

**Table 77. ADC static accuracy at  $f_{ADC} = 18\text{ MHz}^{(1)}$**

Symbol	Parameter	Test conditions	Typ	Max <sup>(2)</sup>	Unit
ET	Total unadjusted error	$f_{ADC} = 18\text{ MHz}$ $V_{DDA} = 1.8\text{ to }3.6\text{ V}$ $V_{REF} = 1.8\text{ to }3.6\text{ V}$ $V_{DDA} - V_{REF} < 1.2\text{ V}$	±3	±4	LSB
EO	Offset error		±2	±3	
EG	Gain error		±1	±3	
ED	Differential linearity error		±1	±2	
EL	Integral linearity error		±2	±3	

1. Better performance could be achieved in restricted  $V_{DD}$ , frequency and temperature ranges.
2. Based on characterization, not tested in production.

**Table 78. ADC static accuracy at  $f_{ADC} = 30\text{ MHz}^{(1)}$**

Symbol	Parameter	Test conditions	Typ	Max <sup>(2)</sup>	Unit
ET	Total unadjusted error	$f_{ADC} = 30\text{ MHz}$ , $R_{AIN} < 10\text{ k}\Omega$ , $V_{DDA} = 2.4\text{ to }3.6\text{ V}$ , $V_{REF} = 1.8\text{ to }3.6\text{ V}$ , $V_{DDA} - V_{REF} < 1.2\text{ V}$	±2	±5	LSB
EO	Offset error		±1.5	±2.5	
EG	Gain error		±1.5	±3	
ED	Differential linearity error		±1	±2	
EL	Integral linearity error		±1.5	±3	

1. Better performance could be achieved in restricted  $V_{DD}$ , frequency and temperature ranges.
2. Based on characterization, not tested in production.

**Table 79. ADC static accuracy at  $f_{ADC} = 36\text{ MHz}^{(1)}$**

Symbol	Parameter	Test conditions	Typ	Max <sup>(2)</sup>	Unit
ET	Total unadjusted error	$f_{ADC} = 36\text{ MHz}$ , $V_{DDA} = 2.4\text{ to }3.6\text{ V}$ , $V_{REF} = 1.8\text{ to }3.6\text{ V}$ $V_{DDA} - V_{REF} < 1.2\text{ V}$	±4	±7	LSB
EO	Offset error		±2	±3	
EG	Gain error		±3	±6	
ED	Differential linearity error		±2	±3	
EL	Integral linearity error		±3	±6	

1. Better performance could be achieved in restricted  $V_{DD}$ , frequency and temperature ranges.
2. Based on characterization, not tested in production.

**Table 80. ADC dynamic accuracy at  $f_{ADC} = 18$  MHz - limited test conditions<sup>(1)</sup>**

Symbol	Parameter	Test conditions	Min	Typ	Max	Unit
ENOB	Effective number of bits	$f_{ADC} = 18$ MHz $V_{DDA} = V_{REF+} = 1.8$ V Input Frequency = 20 KHz Temperature = 25 °C	10.3	10.4	-	bits
SINAD	Signal-to-noise and distortion ratio		64	64.2	-	dB
SNR	Signal-to-noise ratio		64	65	-	
THD	Total harmonic distortion		-67	-72	-	

1. Data based on characterization results, not tested in production.

**Table 81. ADC dynamic accuracy at  $f_{ADC} = 36$  MHz - limited test conditions<sup>(1)</sup>**

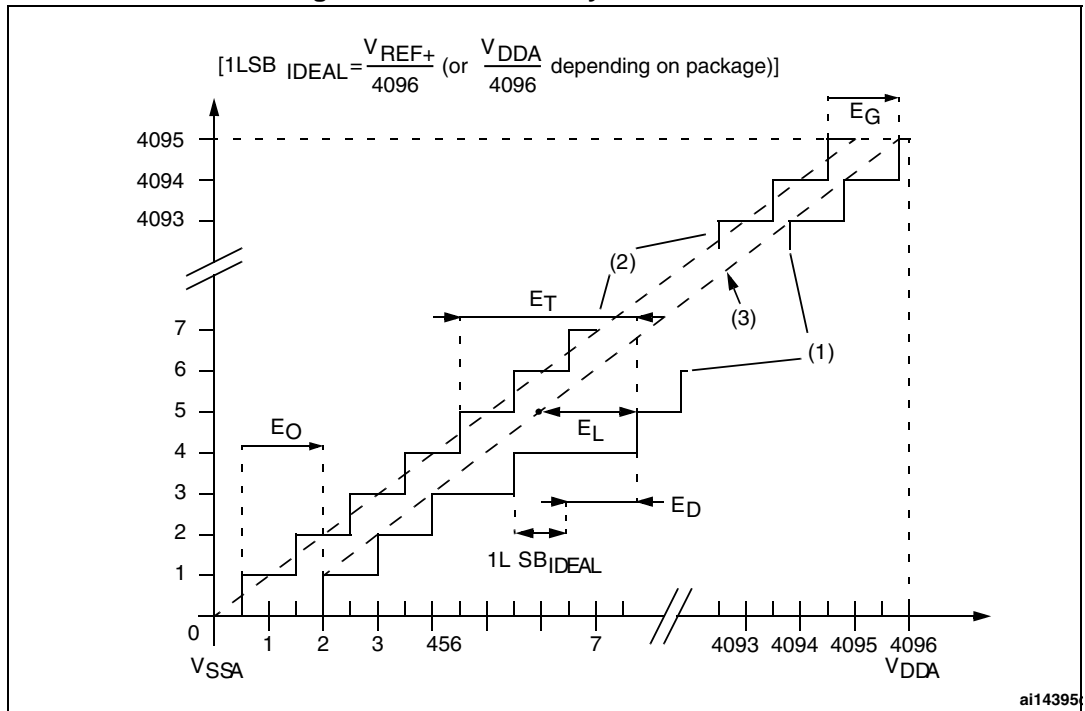
Symbol	Parameter	Test conditions	Min	Typ	Max	Unit
ENOB	Effective number of bits	$f_{ADC} = 36$ MHz $V_{DDA} = V_{REF+} = 3.3$ V Input Frequency = 20 KHz Temperature = 25 °C	10.6	10.8	-	bits
SINAD	Signal-to noise and distortion ratio		66	67	-	dB
SNR	Signal-to noise ratio		64	68	-	
THD	Total harmonic distortion		-70	-72	-	

1. Data based on characterization results, not tested in production.

Note: ADC accuracy vs. negative injection current: injecting a negative current on any analog input pins should be avoided as this significantly reduces the accuracy of the conversion being performed on another analog input. It is recommended to add a Schottky diode (pin to ground) to analog pins which may potentially inject negative currents.

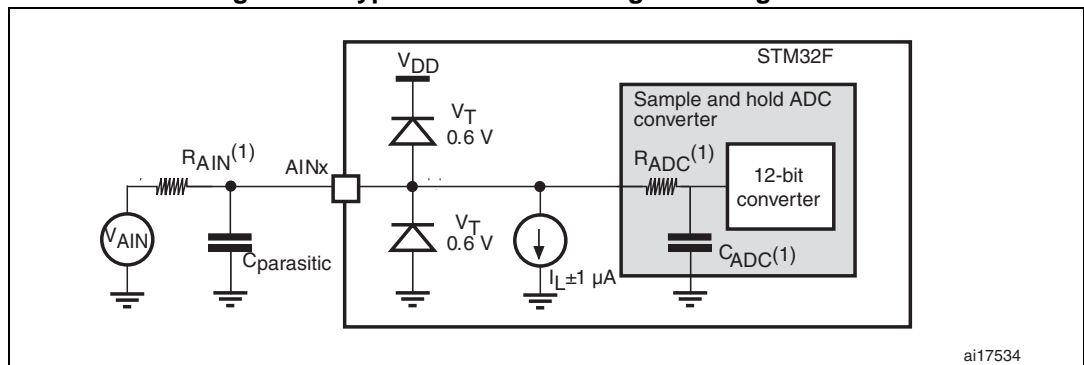
Any positive injection current within the limits specified for  $I_{INJ(PIN)}$  and  $\Sigma I_{INJ(PIN)}$  in [Section 6.3.17](#) does not affect the ADC accuracy.

Figure 51. ADC accuracy characteristics



1. See also [Table 78](#).
2. Example of an actual transfer curve.
3. Ideal transfer curve.
4. End point correlation line.
5.  $E_T$  = Total Unadjusted Error: maximum deviation between the actual and the ideal transfer curves.  
 $E_O$  = Offset Error: deviation between the first actual transition and the first ideal one.  
 $E_G$  = Gain Error: deviation between the last ideal transition and the last actual one.  
 $E_D$  = Differential Linearity Error: maximum deviation between actual steps and the ideal one.  
 $E_L$  = Integral Linearity Error: maximum deviation between any actual transition and the end point correlation line.

Figure 52. Typical connection diagram using the ADC

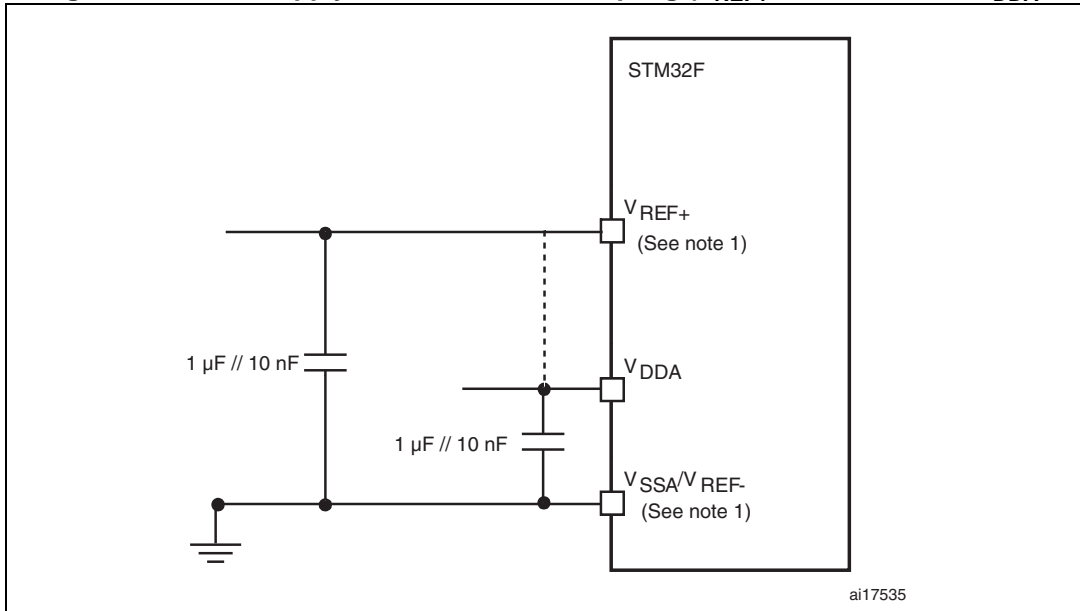


1. Refer to [Table 76](#) for the values of  $R_{AIN}$ ,  $R_{ADC}$  and  $C_{ADC}$ .
2.  $C_{parasitic}$  represents the capacitance of the PCB (dependent on soldering and PCB layout quality) plus the pad capacitance (roughly 5 pF). A high  $C_{parasitic}$  value downgrades conversion accuracy. To remedy this,  $f_{ADC}$  should be reduced.

**General PCB design guidelines**

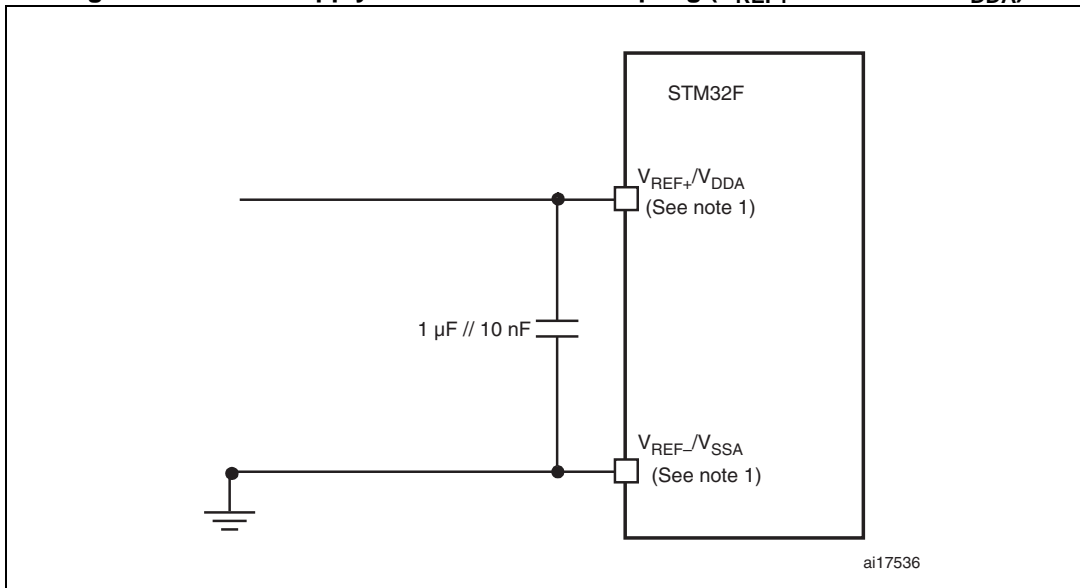
Power supply decoupling should be performed as shown in [Figure 53](#) or [Figure 54](#), depending on whether  $V_{REF+}$  is connected to  $V_{DDA}$  or not. The 10 nF capacitors should be ceramic (good quality). They should be placed them as close as possible to the chip.

**Figure 53. Power supply and reference decoupling ( $V_{REF+}$  not connected to  $V_{DDA}$ )**



1.  $V_{REF+}$  and  $V_{REF-}$  inputs are both available on UFBGA176.  $V_{REF+}$  is also available on LQFP100, LQFP144, and LQFP176. When  $V_{REF+}$  and  $V_{REF-}$  are not available, they are internally connected to  $V_{DDA}$  and  $V_{SSA}$ .

**Figure 54. Power supply and reference decoupling ( $V_{REF+}$  connected to  $V_{DDA}$ )**



1.  $V_{REF+}$  and  $V_{REF-}$  inputs are both available on UFBGA176.  $V_{REF+}$  is also available on LQFP100, LQFP144, and LQFP176. When  $V_{REF+}$  and  $V_{REF-}$  are not available, they are internally connected to  $V_{DDA}$  and  $V_{SSA}$ .

### 6.3.22 Temperature sensor characteristics

**Table 82. Temperature sensor characteristics**

Symbol	Parameter	Min	Typ	Max	Unit
$T_L^{(1)}$	$V_{SENSE}$ linearity with temperature	-	$\pm 1$	$\pm 2$	$^{\circ}\text{C}$
Avg_Slope <sup>(1)</sup>	Average slope	-	2.5		mV/ $^{\circ}\text{C}$
$V_{25}^{(1)}$	Voltage at 25 $^{\circ}\text{C}$	-	0.76		V
$t_{START}^{(2)}$	Startup time	-	6	10	$\mu\text{s}$
$T_{S\_temp}^{(3)(2)}$	ADC sampling time when reading the temperature (1 $^{\circ}\text{C}$ accuracy)	10	-	-	$\mu\text{s}$

1. Based on characterization, not tested in production.
2. Guaranteed by design, not tested in production.
3. Shortest sampling time can be determined in the application by multiple iterations.

**Table 83. Temperature sensor calibration values**

Symbol	Parameter	Memory address
TS_CAL1	TS ADC raw data acquired at temperature of 30 $^{\circ}\text{C}$ , $V_{DDA} = 3.3\text{ V}$	0x1FFF 7A2C - 0x1FFF 7A2D
TS_CAL2	TS ADC raw data acquired at temperature of 110 $^{\circ}\text{C}$ , $V_{DDA} = 3.3\text{ V}$	0x1FFF 7A2E - 0x1FFF 7A2F

### 6.3.23 $V_{BAT}$ monitoring characteristics

**Table 84.  $V_{BAT}$  monitoring characteristics**

Symbol	Parameter	Min	Typ	Max	Unit
R	Resistor bridge for $V_{BAT}$	-	50	-	K $\Omega$
Q	Ratio on $V_{BAT}$ measurement	-	4	-	
$Er^{(1)}$	Error on Q	-1	-	+1	%
$T_{S\_vbat}^{(2)(2)}$	ADC sampling time when reading the $V_{BAT}$ 1 mV accuracy	5	-	-	$\mu\text{s}$

1. Guaranteed by design, not tested in production.
2. Shortest sampling time can be determined in the application by multiple iterations.

### 6.3.24 reference voltage

The parameters given in [Table 85](#) are derived from tests performed under ambient temperature and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#).

**Table 85. internal reference voltage**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{REFINT}$	Internal reference voltage	$-40\text{ }^{\circ}\text{C} < T_A < +105\text{ }^{\circ}\text{C}$	1.18	1.21	1.24	V
$T_{S\_vrefint}^{(1)}$	ADC sampling time when reading the internal reference voltage		10	-	-	$\mu\text{s}$

**Table 85. internal reference voltage (continued)**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
$V_{RERINT\_s}^{(2)}$	Internal reference voltage spread over the temperature range	$V_{DD} = 3V \pm 10mV$	-	3	5	mV
$T_{Ccoeff}^{(2)}$	Temperature coefficient		-	30	50	ppm/°C
$t_{START}^{(2)}$	Startup time		-	6	10	µs

1. Shortest sampling time can be determined in the application by multiple iterations.
2. Guaranteed by design, not tested in production

**Table 86. Internal reference voltage calibration values**

Symbol	Parameter	Memory address
$V_{REFIN\_CAL}$	Raw data acquired at temperature of 30 °C $V_{DDA} = 3.3 V$	0x1FFF 7A2A - 0x1FFF 7A2B

**6.3.25 DAC electrical characteristics**

**Table 87. DAC characteristics**

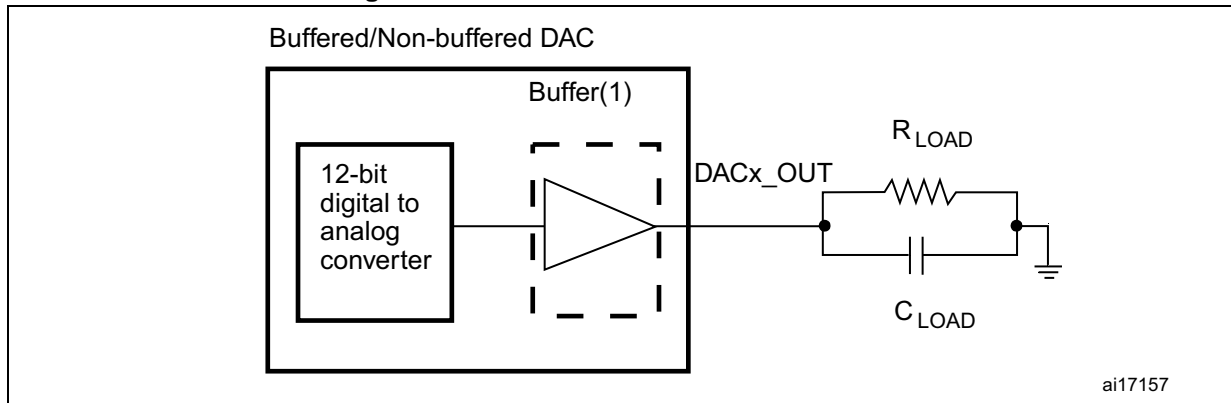
Symbol	Parameter	Min	Typ	Max	Unit	Comments
$V_{DDA}$	Analog supply voltage	1.8 <sup>(1)</sup>	-	3.6	V	
$V_{REF+}$	Reference supply voltage	1.8 <sup>(1)</sup>	-	3.6	V	$V_{REF+} \leq V_{DDA}$
$V_{SSA}$	Ground	0	-	0	V	
$R_{LOAD}^{(2)}$	Resistive load with buffer ON	5	-	-	kΩ	
$R_O^{(2)}$	Impedance output with buffer OFF	-	-	15	kΩ	When the buffer is OFF, the Minimum resistive load between DAC_OUT and $V_{SS}$ to have a 1% accuracy is 1.5 MΩ
$C_{LOAD}^{(2)}$	Capacitive load	-	-	50	pF	Maximum capacitive load at DAC_OUT pin (when the buffer is ON).
$DAC\_OUT_{min}^{(2)}$	Lower DAC_OUT voltage with buffer ON	0.2	-	-	V	It gives the maximum output excursion of the DAC. It corresponds to 12-bit input code (0x0E0) to (0xF1C) at $V_{REF+} = 3.6 V$ and (0x1C7) to (0xE38) at $V_{REF+} = 1.8 V$
$DAC\_OUT_{max}^{(2)}$	Higher DAC_OUT voltage with buffer ON	-	-	$V_{DDA} - 0.2$	V	
$DAC\_OUT_{min}^{(2)}$	Lower DAC_OUT voltage with buffer OFF	-	0.5	-	mV	It gives the maximum output excursion of the DAC.
$DAC\_OUT_{max}^{(2)}$	Higher DAC_OUT voltage with buffer OFF	-	-	$V_{REF+} - 1LSB$	V	
$I_{VREF+}^{(4)}$	DAC DC $V_{REF}$ current consumption in quiescent mode (Standby mode)	-	170	240	µA	With no load, worst code (0x800) at $V_{REF+} = 3.6 V$ in terms of DC consumption on the inputs
		-	50	75		With no load, worst code (0xF1C) at $V_{REF+} = 3.6 V$ in terms of DC consumption on the inputs

Table 87. DAC characteristics (continued)

Symbol	Parameter	Min	Typ	Max	Unit	Comments
$I_{DDA}^{(4)}$	DAC DC VDDA current consumption in quiescent mode <sup>(3)</sup>	-	280	380	$\mu\text{A}$	With no load, middle code (0x800) on the inputs
		-	475	625	$\mu\text{A}$	With no load, worst code (0xF1C) at $V_{REF+} = 3.6\text{ V}$ in terms of DC consumption on the inputs
DNL <sup>(4)</sup>	Differential non linearity Difference between two consecutive code-1LSB)	-	-	$\pm 0.5$	LSB	Given for the DAC in 10-bit configuration.
		-	-	$\pm 2$	LSB	Given for the DAC in 12-bit configuration.
INL <sup>(4)</sup>	Integral non linearity (difference between measured value at Code i and the value at Code i on a line drawn between Code 0 and last Code 1023)	-	-	$\pm 1$	LSB	Given for the DAC in 10-bit configuration.
		-	-	$\pm 4$	LSB	Given for the DAC in 12-bit configuration.
Offset <sup>(4)</sup>	Offset error (difference between measured value at Code (0x800) and the ideal value = $V_{REF+}/2$ )	-	-	$\pm 10$	mV	Given for the DAC in 12-bit configuration
		-	-	$\pm 3$	LSB	Given for the DAC in 10-bit at $V_{REF+} = 3.6\text{ V}$
		-	-	$\pm 12$	LSB	Given for the DAC in 12-bit at $V_{REF+} = 3.6\text{ V}$
Gain error <sup>(4)</sup>	Gain error	-	-	$\pm 0.5$	%	Given for the DAC in 12-bit configuration
$t_{SETTLING}^{(4)}$	Settling time (full scale: for a 10-bit input code transition between the lowest and the highest input codes when DAC_OUT reaches final value $\pm 4\text{LSB}$ )	-	3	6	$\mu\text{s}$	$C_{LOAD} \leq 50\text{ pF}$ , $R_{LOAD} \geq 5\text{ k}\Omega$
THD <sup>(4)</sup>	Total Harmonic Distortion Buffer ON	-	-	-	dB	$C_{LOAD} \leq 50\text{ pF}$ , $R_{LOAD} \geq 5\text{ k}\Omega$
Update rate <sup>(2)</sup>	Max frequency for a correct DAC_OUT change when small variation in the input code (from code i to i+1LSB)	-	-	1	MS/s	$C_{LOAD} \leq 50\text{ pF}$ , $R_{LOAD} \geq 5\text{ k}\Omega$
$t_{WAKEUP}^{(4)}$	Wakeup time from off state (Setting the ENx bit in the DAC Control register)	-	6.5	10	$\mu\text{s}$	$C_{LOAD} \leq 50\text{ pF}$ , $R_{LOAD} \geq 5\text{ k}\Omega$ input code between lowest and highest possible ones.
PSRR <sup>+</sup> (2)	Power supply rejection ratio (to $V_{DDA}$ ) (static DC measurement)	-	-67	-40	dB	No $R_{LOAD}$ , $C_{LOAD} = 50\text{ pF}$

- $V_{DDA}$  minimum value of 1.7 V is obtained with the use of an external power supply supervisor (refer to [Section 3.17.2: Internal reset OFF](#)).
- Guaranteed by design, not tested in production.
- The quiescent mode corresponds to a state where the DAC maintains a stable output level to ensure that no dynamic consumption occurs.
- Guaranteed by characterization, not tested in production.

Figure 55. 12-bit buffered /non-buffered DAC



1. The DAC integrates an output buffer that can be used to reduce the output impedance and to drive external loads directly without the use of an external operational amplifier. The buffer can be bypassed by configuring the BOFFx bit in the DAC\_CR register.

### 6.3.26 FMC characteristics

Unless otherwise specified, the parameters given in [Table 88](#) to [Table 103](#) for the FMC interface are derived from tests performed under the ambient temperature,  $f_{HCLK}$  frequency and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#), with the following configuration:

- Output speed is set to  $OSPEEDRy[1:0] = 10$  except at  $V_{DD}$  range 1.7 to 2.1V where  $OSPEEDRy[1:0] = 11$
- Measurement points are done at CMOS levels:  $0.5V_{DD}$

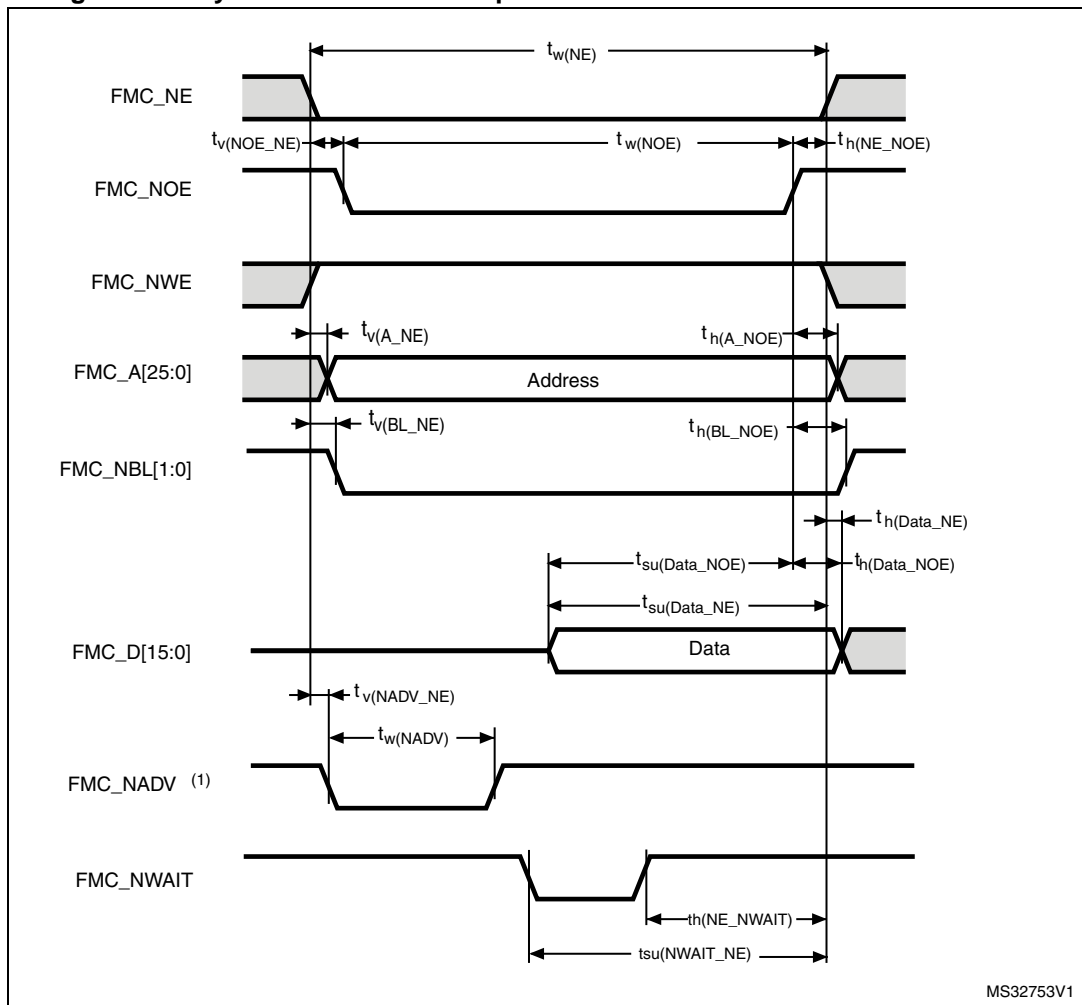
Refer to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output characteristics.

### Asynchronous waveforms and timings

[Figure 56](#) through [Figure 59](#) represent asynchronous waveforms and [Table 88](#) through [Table 95](#) provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- AddressSetupTime = 0x1
- AddressHoldTime = 0x1
- DataSetupTime = 0x1 (except for asynchronous NWAIT mode , DataSetupTime = 0x5)
- BusTurnAroundDuration = 0x0
- For SDRAM memories,  $V_{DD}$  ranges from 2.7 to 3.6 V and maximum frequency FMC\_SDCLK = 90 MHz
- For Mobile LPDDR SDRAM memories,  $V_{DD}$  ranges from 1.7 to 1.95 V and maximum frequency FMC\_SDCLK = 84 MHz

Figure 56. Asynchronous non-multiplexed SRAM/PSRAM/NOR read waveforms



MS32753V1

1. Mode 2/B, C and D only. In Mode 1, FMC\_NADV is not used.

Table 88. Asynchronous non-multiplexed SRAM/PSRAM/NOR - read timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$2T_{HCLK} - 0.5$	$2T_{HCLK} + 0.5$	ns
$t_{v(NOE\_NE)}$	FMC_NEx low to FMC_NOE low	0	1	ns
$t_{w(NOE)}$	FMC_NOE low time	$2T_{HCLK}$	$2T_{HCLK} + 0.5$	ns
$t_{h(NE\_NOE)}$	FMC_NOE high to FMC_NE high hold time	0	-	ns
$t_{v(A\_NE)}$	FMC_NEx low to FMC_A valid	-	2	ns
$t_{h(A\_NOE)}$	Address hold time after FMC_NOE high	0	-	ns
$t_{v(BL\_NE)}$	FMC_NEx low to FMC_BL valid	-	2	ns
$t_{h(BL\_NOE)}$	FMC_BL hold time after FMC_NOE high	0	-	ns
$t_{su(Data\_NE)}$	Data to FMC_NEx high setup time	$T_{HCLK} + 2.5$	-	ns
$t_{su(Data\_NOE)}$	Data to FMC_NOEx high setup time	$T_{HCLK} + 2$	-	ns

**Table 88. Asynchronous non-multiplexed SRAM/PSRAM/NOR - read timings<sup>(1)(2)</sup> (continued)**

Symbol	Parameter	Min	Max	Unit
$t_{h(Data\_NOE)}$	Data hold time after FMC_NOE high	0	-	ns
$t_{h(Data\_NE)}$	Data hold time after FMC_NEx high	0	-	ns
$t_{v(NADV\_NE)}$	FMC_NEx low to FMC_NADV low	-	0	ns
$t_{w(NADV)}$	FMC_NADV low time	-	$T_{HCLK} + 1$	ns

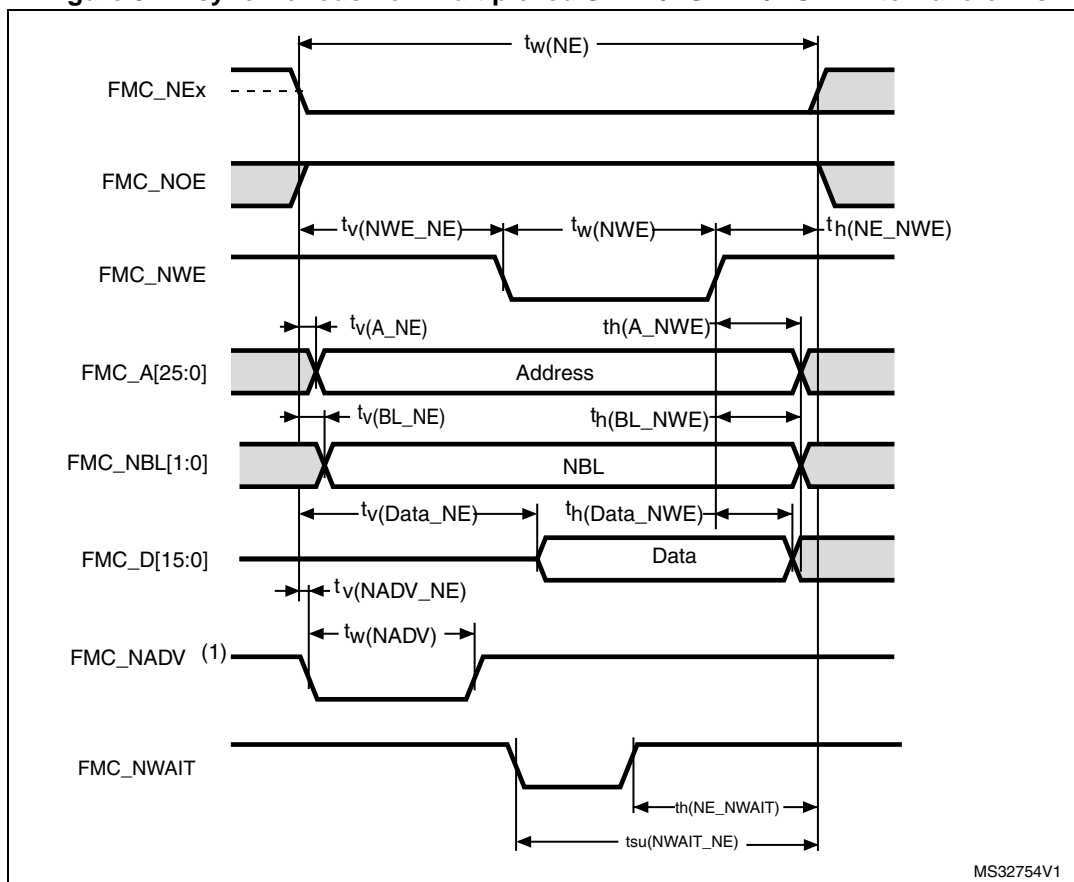
1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

**Table 89. Asynchronous non-multiplexed SRAM/PSRAM/NOR read - NWAIT timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$7T_{HCLK} + 0.5$	$7T_{HCLK} + 1$	ns
$t_{w(NOE)}$	FMC_NWE low time	$5T_{HCLK} - 1.5$	$5T_{HCLK} + 2$	
$t_{su(NWAIT\_NE)}$	FMC_NWAIT valid before FMC_NEx high	$5T_{HCLK} + 1.5$	-	
$t_{h(NE\_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK} + 1$	-	

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

Figure 57. Asynchronous non-multiplexed SRAM/PSRAM/NOR write waveforms



MS32754V1

1. Mode 2/B, C and D only. In Mode 1, FMC\_NADV is not used.

Table 90. Asynchronous non-multiplexed SRAM/PSRAM/NOR write timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$3T_{HCLK}$	$3T_{HCLK}+1$	ns
$t_{v(NWE\_NE)}$	FMC_NEx low to FMC_NWE low	$T_{HCLK}-0.5$	$T_{HCLK}+0.5$	ns
$t_{w(NWE)}$	FMC_NWE low time	$T_{HCLK}$	$T_{HCLK}+0.5$	ns
$t_{h(NE\_NWE)}$	FMC_NWE high to FMC_NE high hold time	$T_{HCLK}+1.5$	-	ns
$t_{v(A\_NE)}$	FMC_NEx low to FMC_A valid	-	0	ns
$t_{h(A\_NWE)}$	Address hold time after FMC_NWE high	$T_{HCLK}+0.5$	-	ns
$t_{v(BL\_NE)}$	FMC_NEx low to FMC_NBL valid	-	1.5	ns
$t_{h(BL\_NWE)}$	FMC_NBL hold time after FMC_NWE high	$T_{HCLK}+0.5$	-	ns
$t_{v(Data\_NE)}$	Data to FMC_NEx low to Data valid	-	$T_{HCLK}+2$	ns
$t_{h(Data\_NWE)}$	Data hold time after FMC_NWE high	$T_{HCLK}+0.5$	-	ns
$t_{v(NADV\_NE)}$	FMC_NEx low to FMC_NADV low	-	0.5	ns
$t_{w(NADV)}$	FMC_NADV low time	-	$T_{HCLK}+0.5$	ns

1.  $C_L = 30$  pF.

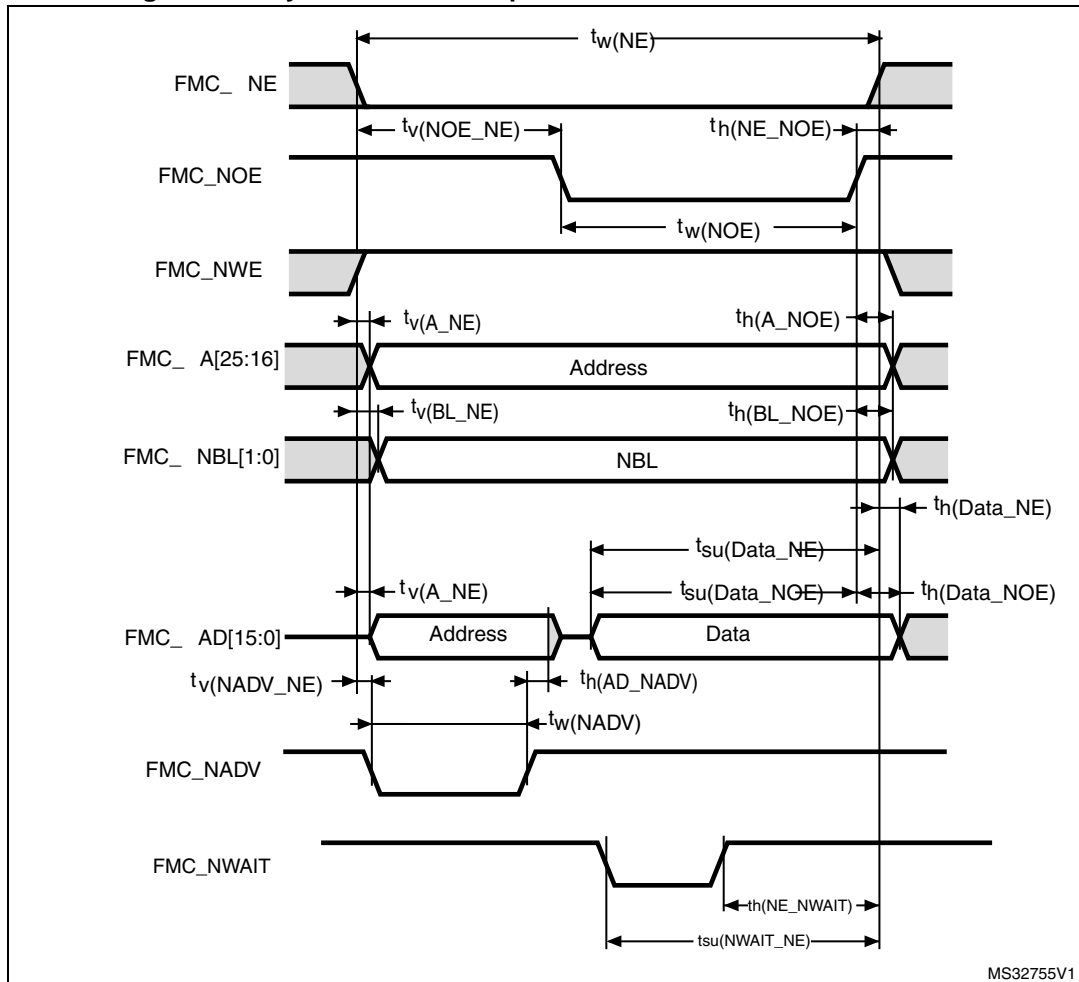
2. Based on characterization, not tested in production.

**Table 91. Asynchronous non-multiplexed SRAM/PSRAM/NOR write - NWAIT timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{HCLK}+1$	$8T_{HCLK}+2$	ns
$t_{w(NWE)}$	FMC_NWE low time	$6T_{HCLK}-1$	$6T_{HCLK}+2$	ns
$t_{su(NWAIT\_NE)}$	FMC_NWAIT valid before FMC_NEx high	$6T_{HCLK}+1.5$	-	ns
$t_{h(NE\_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}+1$		ns

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

**Figure 58. Asynchronous multiplexed PSRAM/NOR read waveforms**



MS32755V1

**Table 92. Asynchronous multiplexed PSRAM/NOR read timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$3T_{HCLK}-1$	$3T_{HCLK}+0.5$	ns
$t_{v(NOE\_NE)}$	FMC_NEx low to FMC_NOE low	$2T_{HCLK}-0.5$	$2T_{HCLK}$	ns
$t_{tw(NOE)}$	FMC_NOE low time	$T_{HCLK}-1$	$T_{HCLK}+1$	ns
$t_{h(NE\_NOE)}$	FMC_NOE high to FMC_NE high hold time	1	-	ns
$t_{v(A\_NE)}$	FMC_NEx low to FMC_A valid	-	2	ns
$t_{v(NADV\_NE)}$	FMC_NEx low to FMC_NADV low	0	2	ns
$t_{w(NADV)}$	FMC_NADV low time	$T_{HCLK}-0.5$	$T_{HCLK}+0.5$	ns
$t_{h(AD\_NADV)}$	FMC_AD(address) valid hold time after FMC_NADV high	0	-	ns
$t_{h(A\_NOE)}$	Address hold time after FMC_NOE high	$T_{HCLK}-0.5$	-	ns
$t_{h(BL\_NOE)}$	FMC_BL time after FMC_NOE high	0	-	ns
$t_{v(BL\_NE)}$	FMC_NEx low to FMC_BL valid	-	2	ns
$t_{su(Data\_NE)}$	Data to FMC_NEx high setup time	$T_{HCLK}+1.5$	-	ns
$t_{su(Data\_NOE)}$	Data to FMC_NOE high setup time	$T_{HCLK}+1$	-	ns
$t_{h(Data\_NE)}$	Data hold time after FMC_NEx high	0	-	ns
$t_{h(Data\_NOE)}$	Data hold time after FMC_NOE high	0	-	ns

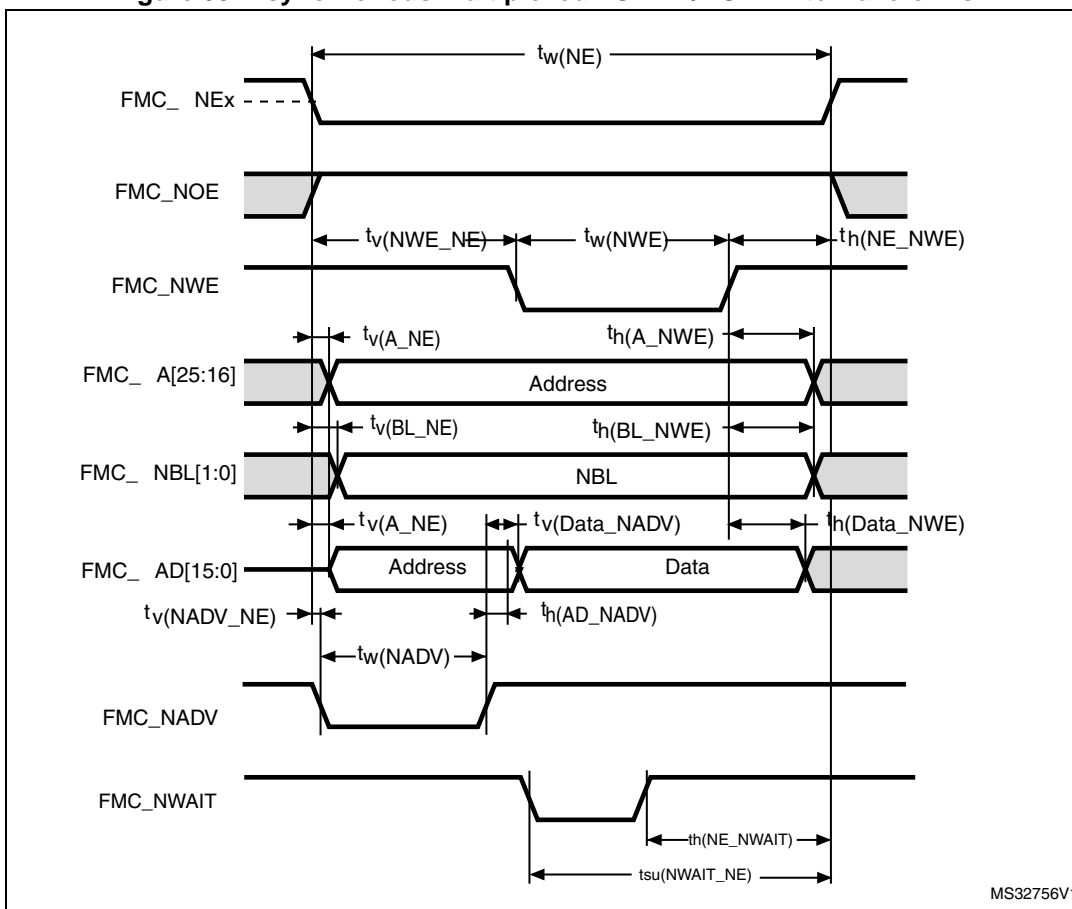
1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

**Table 93. Asynchronous multiplexed PSRAM/NOR read-NWAIT timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$8T_{HCLK}+0.5$	$8T_{HCLK}+2$	ns
$t_{w(NOE)}$	FMC_NWE low time	$5T_{HCLK}-1$	$5T_{HCLK}+1.5$	ns
$t_{su(NWAIT\_NE)}$	FMC_NWAIT valid before FMC_NEx high	$5T_{HCLK}+1.5$	-	ns
$t_{h(NE\_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK}+1$	-	ns

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

Figure 59. Asynchronous multiplexed PSRAM/NOR write waveforms



MS32756V1

Table 94. Asynchronous multiplexed PSRAM/NOR write timings<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
$t_w(NE)$	FMC_NE low time	$4T_{HCLK}$	$4T_{HCLK}+0.5$	ns
$t_v(NWE\_NE)$	FMC_NEx low to FMC_NWE low	$T_{HCLK}-1$	$T_{HCLK}+0.5$	ns
$t_w(NWE)$	FMC_NWE low time	$2T_{HCLK}$	$2T_{HCLK}+0.5$	ns
$t_h(NE\_NWE)$	FMC_NWE high to FMC_NE high hold time	$T_{HCLK}$	-	ns
$t_v(A\_NE)$	FMC_NEx low to FMC_A valid	-	0	ns
$t_v(NADV\_NE)$	FMC_NEx low to FMC_NADV low	0.5	1	ns
$t_w(NADV)$	FMC_NADV low time	$T_{HCLK}-0.5$	$T_{HCLK}+0.5$	ns
$t_h(AD\_NADV)$	FMC_AD(address) valid hold time after FMC_NADV high	$T_{HCLK}-2$	-	ns
$t_h(A\_NWE)$	Address hold time after FMC_NWE high	$T_{HCLK}$	-	ns
$t_h(BL\_NWE)$	FMC_BL hold time after FMC_NWE high	$T_{HCLK}-2$	-	ns
$t_v(BL\_NE)$	FMC_NEx low to FMC_BL valid	-	2	ns

**Table 94. Asynchronous multiplexed PSRAM/NOR write timings<sup>(1)(2)</sup> (continued)**

Symbol	Parameter	Min	Max	Unit
$t_{v(Data\_NADV)}$	FMC_NADV high to Data valid	-	$T_{HCLK} + 1.5$	ns
$t_{h(Data\_NWE)}$	Data hold time after FMC_NWE high	$T_{HCLK} + 0.5$	-	ns

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

**Table 95. Asynchronous multiplexed PSRAM/NOR write-NWAIT timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(NE)}$	FMC_NE low time	$9T_{HCLK}$	$9T_{HCLK} + 0.5$	ns
$t_{w(NWE)}$	FMC_NWE low time	$7T_{HCLK}$	$7T_{HCLK} + 2$	ns
$t_{su(NWAIT\_NE)}$	FMC_NWAIT valid before FMC_NEx high	$6T_{HCLK} + 1.5$	-	ns
$t_{h(NE\_NWAIT)}$	FMC_NEx hold time after FMC_NWAIT invalid	$4T_{HCLK} - 1$	-	ns

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

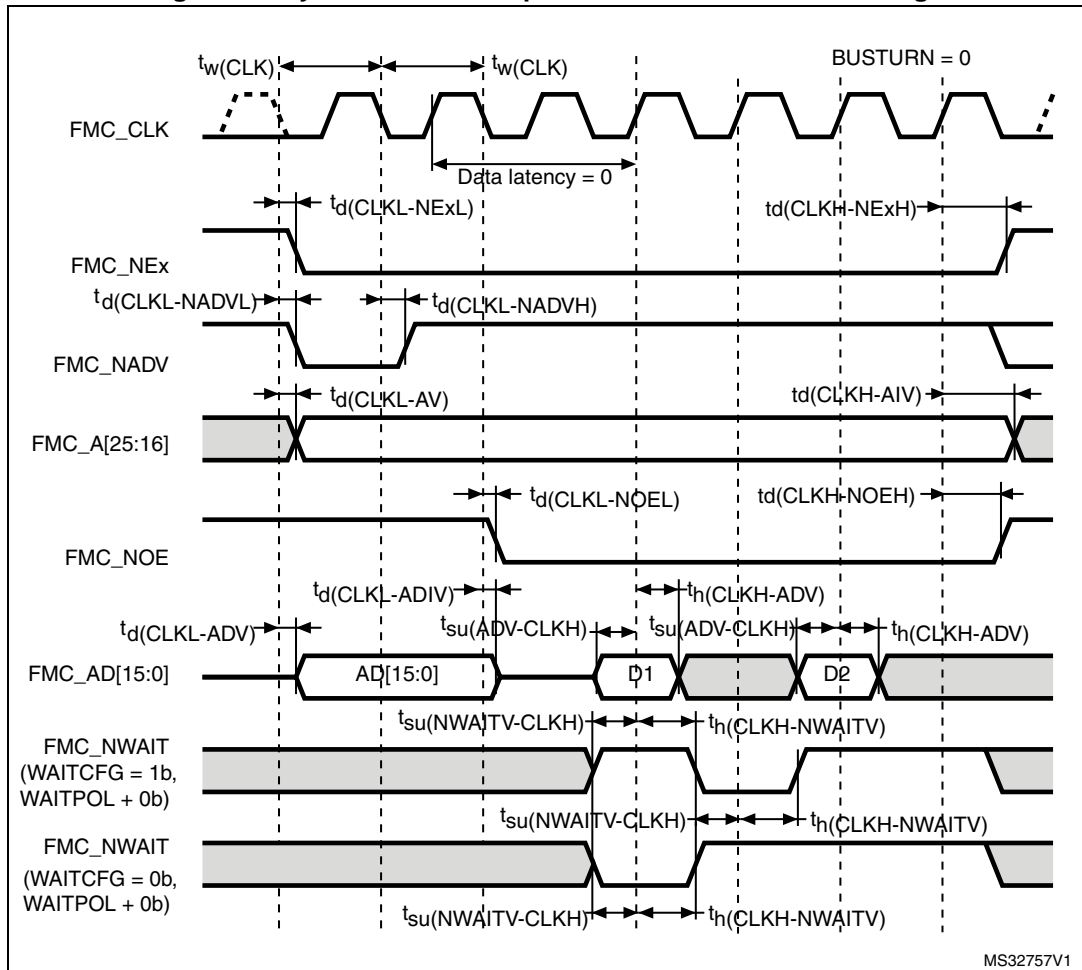
### Synchronous waveforms and timings

Figure 60 through Figure 63 represent synchronous waveforms and Table 96 through Table 99 provide the corresponding timings. The results shown in these tables are obtained with the following FMC configuration:

- BurstAccessMode = FMC\_BurstAccessMode\_Enable;
- MemoryType = FMC\_MemoryType\_CRAM;
- WriteBurst = FMC\_WriteBurst\_Enable;
- CLKDivision = 1; (0 is not supported, see the STM32F4xx reference manual : RM0090)
- DataLatency = 1 for NOR Flash; DataLatency = 0 for PSRAM

In all timing tables, the  $T_{HCLK}$  is the HCLK clock period (with maximum FMC\_CLK = 90 MHz).

Figure 60. Synchronous multiplexed NOR/PSRAM read timings



MS32757V1

Table 96. Synchronous multiplexed NOR/PSRAM read timings<sup>(1)(2)</sup>

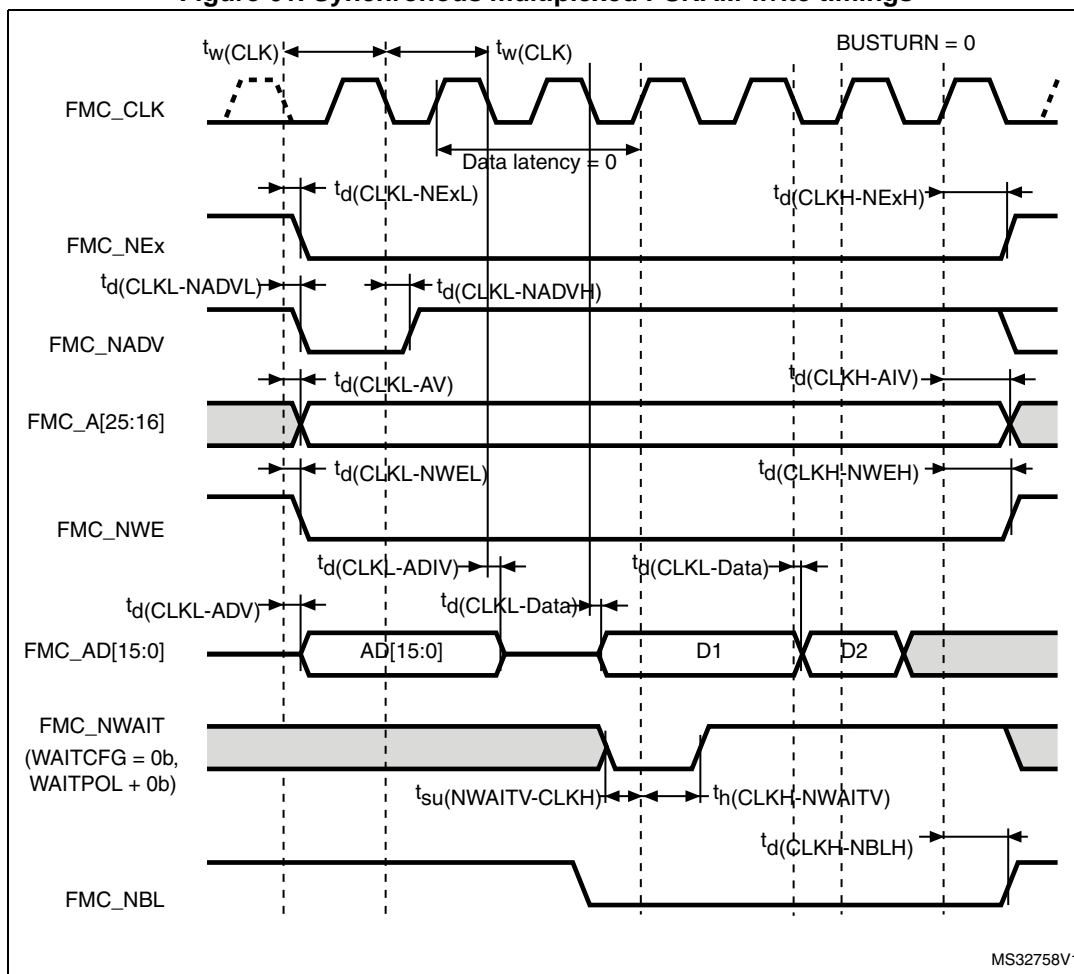
Symbol	Parameter	Min	Max	Unit
$t_{w(CLK)}$	FMC_CLK period	$2T_{HCLK}-1$	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low (x=0..2)	-	0	ns
$t_{d(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high (x= 0...2)	$T_{HCLK}$	-	ns
$t_{d(CLKL-NADV)}$	FMC_CLK low to FMC_NADV low	-	0	ns
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV high	0	-	ns
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16...25)	-	0	ns
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x=16...25)	0	-	ns
$t_{d(CLKL-NOEL)}$	FMC_CLK low to FMC_NOE low	-	$T_{HCLK}+0.5$	ns
$t_{d(CLKH-NOEH)}$	FMC_CLK high to FMC_NOE high	$T_{HCLK}-0.5$	-	ns
$t_{d(CLKL-ADV)}$	FMC_CLK low to FMC_AD[15:0] valid	-	0.5	ns
$t_{d(CLKL-ADIV)}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	ns

**Table 96. Synchronous multiplexed NOR/PSRAM read timings<sup>(1)(2)</sup> (continued)**

Symbol	Parameter	Min	Max	Unit
$t_{su(ADV-CLKH)}$	FMC_A/D[15:0] valid data before FMC_CLK high	5	-	ns
$t_h(CLKH-ADV)$	FMC_A/D[15:0] valid data after FMC_CLK high	0	-	ns
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	4	-	ns
$t_h(CLKH-NWAIT)$	FMC_NWAIT valid after FMC_CLK high	0	-	ns

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

**Figure 61. Synchronous multiplexed PSRAM write timings**



**Table 97. Synchronous multiplexed PSRAM write timings<sup>(1)(2)</sup>**

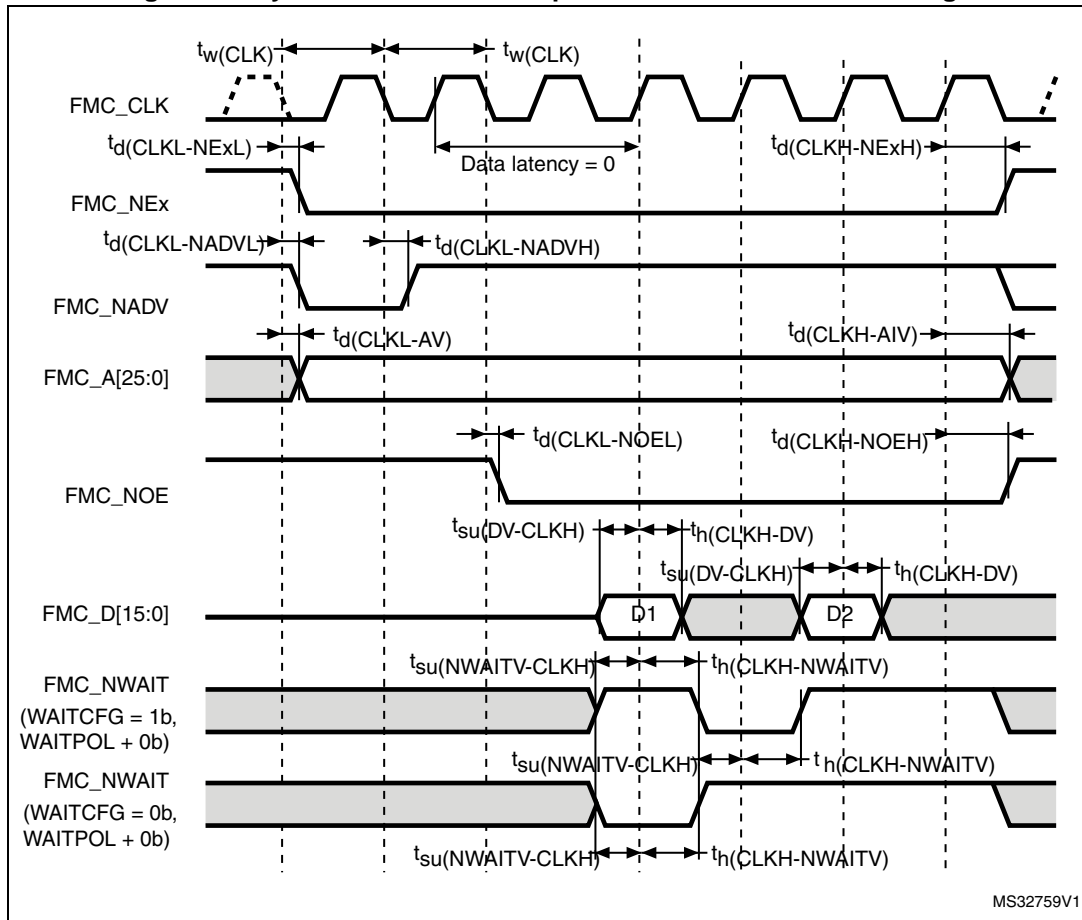
Symbol	Parameter	Min	Max	Unit
$t_w(CLK)$	FMC_CLK period, VDD range= 2.7 to 3.6 V	$2T_{HCLK}-1$	-	ns
$t_d(CLKL-NExL)$	FMC_CLK low to FMC_NEx low (x=0..2)	-	1.5	ns
$t_d(CLKH-NExH)$	FMC_CLK high to FMC_NEx high (x= 0..2)	$T_{HCLK}$	-	ns

Table 97. Synchronous multiplexed PSRAM write timings<sup>(1)(2)</sup> (continued)

Symbol	Parameter	Min	Max	Unit
$t_{d(CLKL-NADVL)}$	FMC_CLK low to FMC_NADV low	-	0	ns
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV high	0	-	ns
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16...25)	-	0	ns
$t_{d(CLKH-AIV)}$	FMC_CLK high to FMC_Ax invalid (x=16...25)	$T_{HCLK}$	-	ns
$t_{d(CLKL-NWEL)}$	FMC_CLK low to FMC_NWE low	-	0	ns
$t_{d(CLKH-NWEH)}$	FMC_CLK high to FMC_NWE high	$T_{HCLK}-0.5$	-	ns
$t_{d(CLKL-ADV)}$	FMC_CLK low to FMC_AD[15:0] valid	-	3	ns
$t_{d(CLKL-ADIV)}$	FMC_CLK low to FMC_AD[15:0] invalid	0	-	ns
$t_{d(CLKL-DATA)}$	FMC_A/D[15:0] valid data after FMC_CLK low	-	3	ns
$t_{d(CLKL-NBLL)}$	FMC_CLK low to FMC_NBL low	0	-	ns
$t_{d(CLKH-NBLH)}$	FMC_CLK high to FMC_NBL high	$T_{HCLK}-0.5$	-	ns
$t_{su(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	4	-	ns
$t_h(CLKH-NWAIT)$	FMC_NWAIT valid after FMC_CLK high	0	-	ns

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

Figure 62. Synchronous non-multiplexed NOR/PSRAM read timings



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Table 98. Synchronous non-multiplexed NOR/PSRAM read timings<sup>(1)(2)</sup>

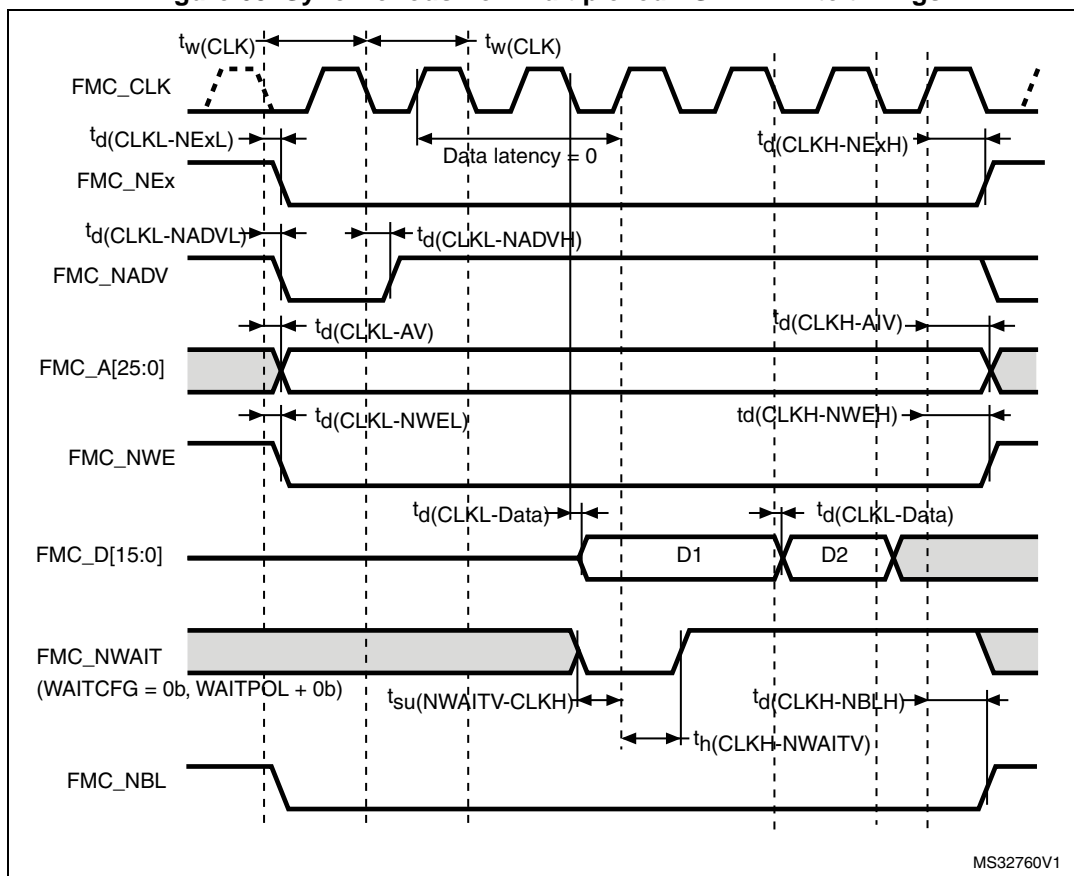
Symbol	Parameter	Min	Max	Unit
$t_w(\text{CLK})$	FMC_CLK period	$2T_{\text{HCLK}}-1$	-	ns
$t_{\text{d}}(\text{CLKL-NExL})$	FMC_CLK low to FMC_NEx low (x=0..2)	-	0.5	ns
$t_{\text{d}}(\text{CLKH-NExH})$	FMC_CLK high to FMC_NEx high (x= 0...2)	$T_{\text{HCLK}}$	-	ns
$t_{\text{d}}(\text{CLKL-NADV})$	FMC_CLK low to FMC_NADV low	-	0	ns
$t_{\text{d}}(\text{CLKL-NADVH})$	FMC_CLK low to FMC_NADV high	0	-	ns
$t_{\text{d}}(\text{CLKL-AV})$	FMC_CLK low to FMC_Ax valid (x=16...25)	-	0	ns
$t_{\text{d}}(\text{CLKH-AIV})$	FMC_CLK high to FMC_Ax invalid (x=16...25)	$T_{\text{HCLK}}-0.5$	-	ns
$t_{\text{d}}(\text{CLKL-NOEL})$	FMC_CLK low to FMC_NOE low	-	$T_{\text{HCLK}}+2$	ns
$t_{\text{d}}(\text{CLKH-NOEH})$	FMC_CLK high to FMC_NOE high	$T_{\text{HCLK}}-0.5$	-	ns
$t_{\text{su}}(\text{DV-CLKH})$	FMC_D[15:0] valid data before FMC_CLK high	5	-	ns

**Table 98. Synchronous non-multiplexed NOR/PSRAM read timings<sup>(1)(2)</sup> (continued)**

Symbol	Parameter	Min	Max	Unit
$t_{h(CLKH-DV)}$	FMC_D[15:0] valid data after FMC_CLK high	0	-	ns
$t_{(NWAIT-CLKH)}$	FMC_NWAIT valid before FMC_CLK high	4		
$t_{h(CLKH-NWAIT)}$	FMC_NWAIT valid after FMC_CLK high	0		

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

**Figure 63. Synchronous non-multiplexed PSRAM write timings**



**Table 99. Synchronous non-multiplexed PSRAM write timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{(CLK)}$	FMC_CLK period	$2T_{HCLK}-1$	-	ns
$t_{d(CLKL-NExL)}$	FMC_CLK low to FMC_NEx low (x=0..2)	-	0.5	ns
$t_{(CLKH-NExH)}$	FMC_CLK high to FMC_NEx high (x= 0..2)	$T_{HCLK}$	-	ns
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV low	-	0	ns
$t_{d(CLKL-NADVH)}$	FMC_CLK low to FMC_NADV high	0	-	ns
$t_{d(CLKL-AV)}$	FMC_CLK low to FMC_Ax valid (x=16..25)	-	0	ns

**Table 99. Synchronous non-multiplexed PSRAM write timings<sup>(1)(2)</sup> (continued)**

Symbol	Parameter	Min	Max	Unit
$t_{d(\text{CLKH-AIV})}$	FMC_CLK high to FMC_Ax invalid (x=16...25)	0	-	ns
$t_{d(\text{CLKL-NWEL})}$	FMC_CLK low to FMC_NWE low	-	0	ns
$t_{d(\text{CLKH-NWEH})}$	FMC_CLK high to FMC_NWE high	$T_{\text{HCLK}}-0.5$	-	ns
$t_{d(\text{CLKL-Data})}$	FMC_D[15:0] valid data after FMC_CLK low	-	2.5	ns
$t_{d(\text{CLKL-NBLL})}$	FMC_CLK low to FMC_NBL low	0	-	ns
$t_{d(\text{CLKH-NBLH})}$	FMC_CLK high to FMC_NBL high	$T_{\text{HCLK}}-0.5$	-	ns
$t_{\text{su}(\text{NWAIT-CLKH})}$	FMC_NWAIT valid before FMC_CLK high	4		
$t_{\text{h}(\text{CLKH-NWAIT})}$	FMC_NWAIT valid after FMC_CLK high	0		

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

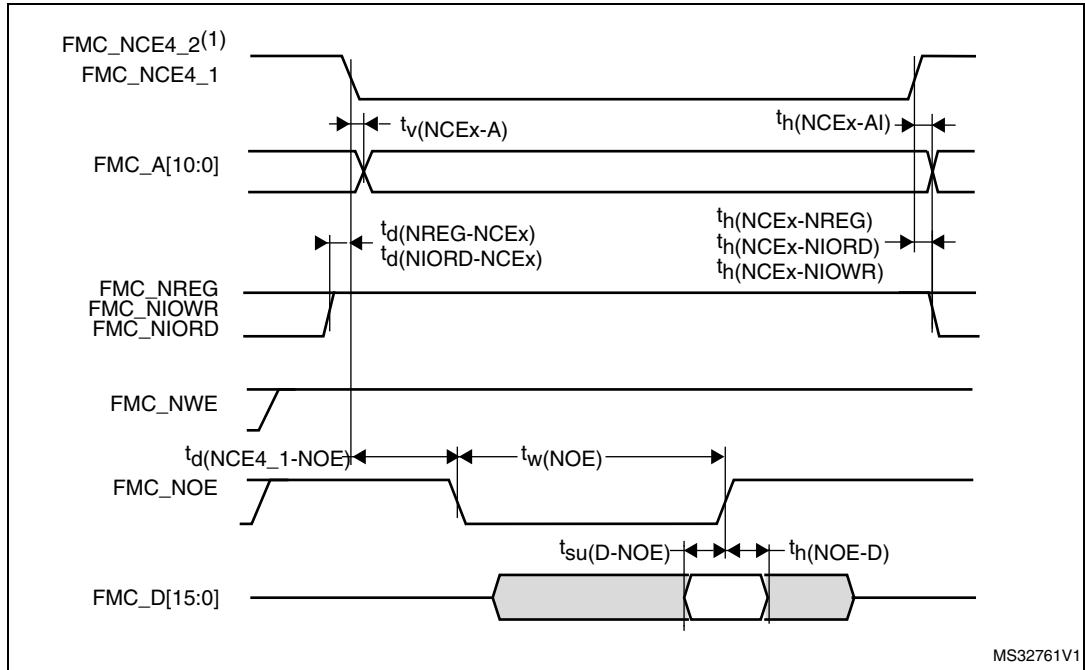
### PC Card/CompactFlash controller waveforms and timings

[Figure 64](#) through [Figure 69](#) represent synchronous waveforms, and [Table 100](#) and [Table 101](#) provide the corresponding timings. The results shown in this table are obtained with the following FMC configuration:

- COM.FMC\_SetupTime = 0x04;
- COM.FMC\_WaitSetupTime = 0x07;
- COM.FMC\_HoldSetupTime = 0x04;
- COM.FMC\_HiZSetupTime = 0x00;
- ATT.FMC\_SetupTime = 0x04;
- ATT.FMC\_WaitSetupTime = 0x07;
- ATT.FMC\_HoldSetupTime = 0x04;
- ATT.FMC\_HiZSetupTime = 0x00;
- IO.FMC\_SetupTime = 0x04;
- IO.FMC\_WaitSetupTime = 0x07;
- IO.FMC\_HoldSetupTime = 0x04;
- IO.FMC\_HiZSetupTime = 0x00;
- TCLRSetupTime = 0;
- TARSetupTime = 0.

In all timing tables, the  $T_{\text{HCLK}}$  is the HCLK clock period.

Figure 64. PC Card/CompactFlash controller waveforms for common memory read access



1. FMC\_NCE4\_2 remains high (inactive during 8-bit access).

Figure 65. PC Card/CompactFlash controller waveforms for common memory write access

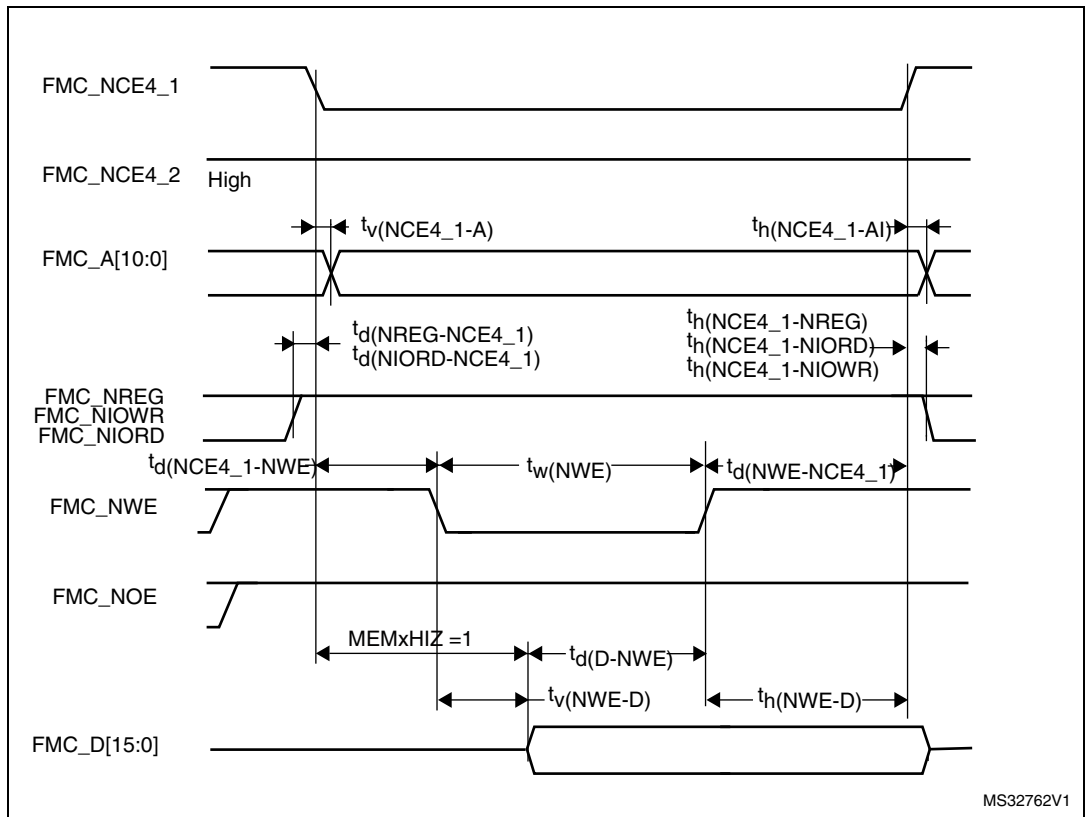
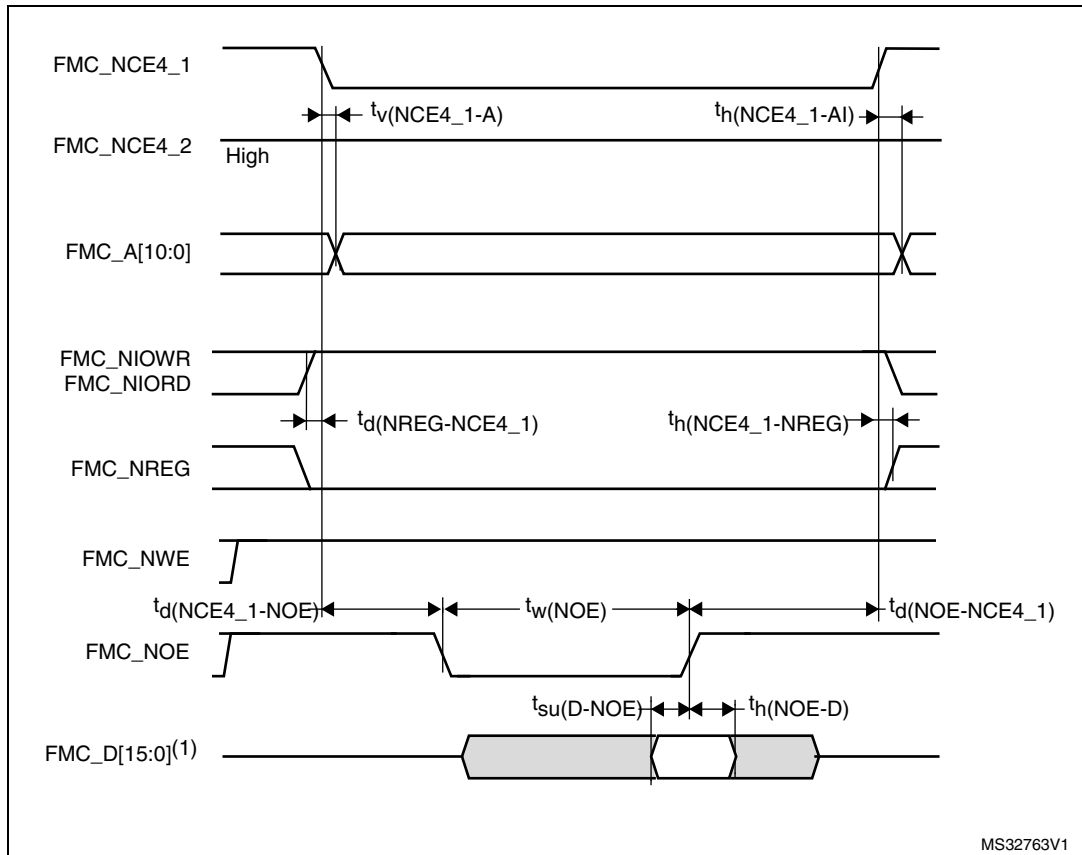
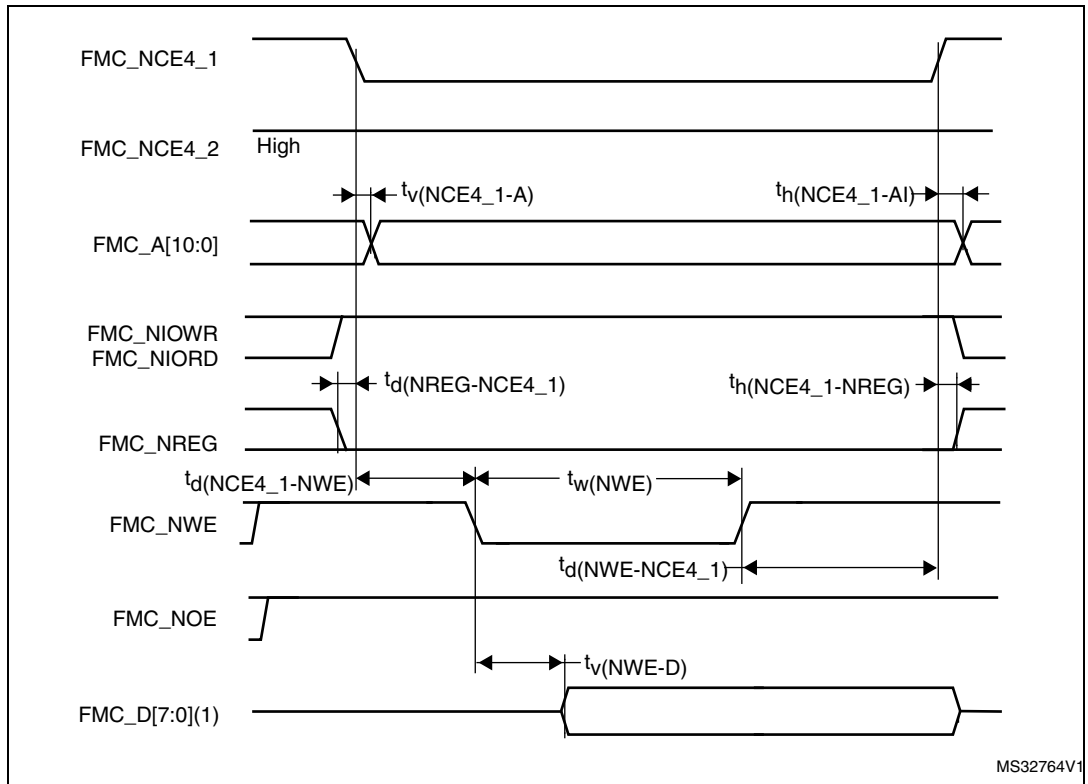


Figure 66. PC Card/CompactFlash controller waveforms for attribute memory read access



1. Only data bits 0...7 are read (bits 8...15 are disregarded).

**Figure 67. PC Card/CompactFlash controller waveforms for attribute memory write access**



1. Only data bits 0...7 are driven (bits 8...15 remains Hi-Z).

**Figure 68. PC Card/CompactFlash controller waveforms for I/O space read access**

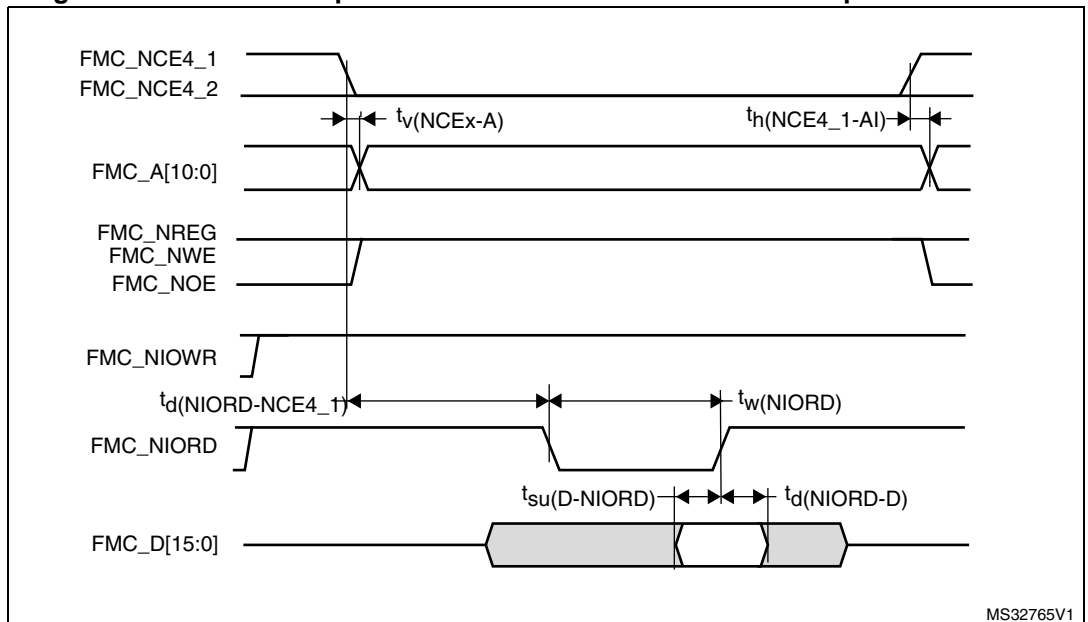
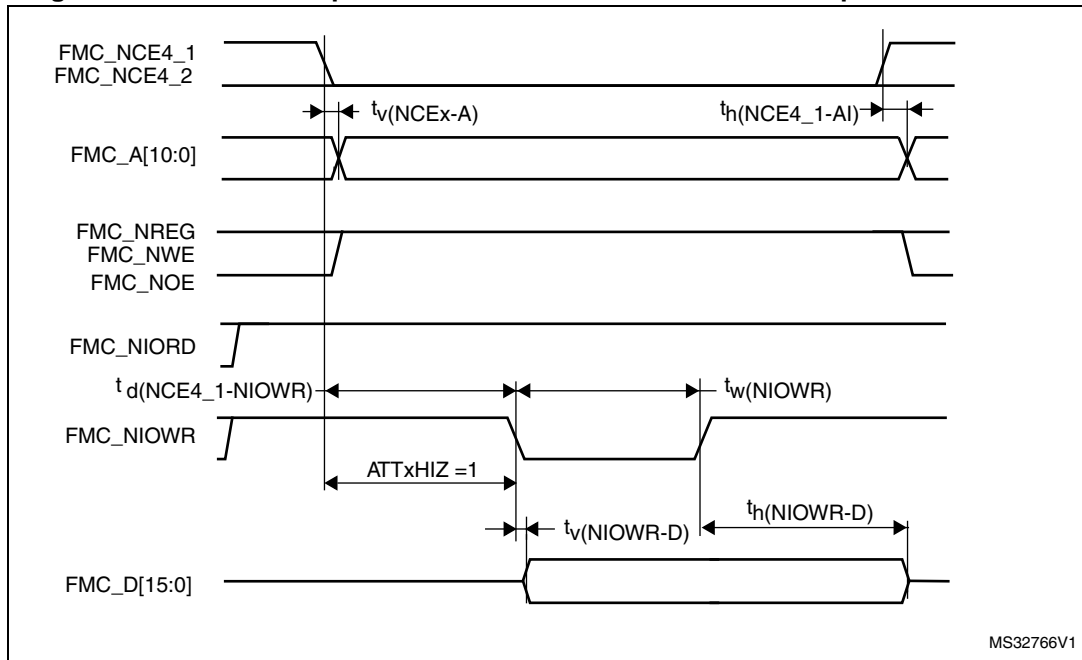


Figure 69. PC Card/CompactFlash controller waveforms for I/O space write access



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Table 100. Switching characteristics for PC Card/CF read and write cycles in attribute/common space<sup>(1)(2)</sup>

Symbol	Parameter	Min	Max	Unit
$t_{v(NCEx-A)}$	FMC_Ncex low to FMC_Ay valid	-	0	ns
$t_{h(NCEx\_AI)}$	FMC_NCEx high to FMC_Ax invalid	0	-	ns
$t_{d(NREG-NCEx)}$	FMC_NCEx low to FMC_NREG valid	-	1	ns
$t_{h(NCEx-NREG)}$	FMC_NCEx high to FMC_NREG invalid	$T_{HCLK}-2$	-	ns
$t_{d(NCEx-NWE)}$	FMC_NCEx low to FMC_NWE low	-	$5T_{HCLK}$	ns
$t_{w(NWE)}$	FMC_NWE low width	$8T_{HCLK}-0.5$	$8T_{HCLK}+0.5$	ns
$t_{d(NWE\_NCEx)}$	FMC_NWE high to FMC_NCEx high	$5T_{HCLK}+1$	-	ns
$t_{v(NWE-D)}$	FMC_NWE low to FMC_D[15:0] valid	-	0	ns
$t_{h(NWE-D)}$	FMC_NWE high to FMC_D[15:0] invalid	$9T_{HCLK}-0.5$	-	ns
$t_{d(D-NWE)}$	FMC_D[15:0] valid before FMC_NWE high	$13T_{HCLK}-3$	-	ns
$t_{d(NCEx-NOE)}$	FMC_NCEx low to FMC_NOE low	-	$5T_{HCLK}$	ns
$t_{w(NOE)}$	FMC_NOE low width	$8T_{HCLK}-0.5$	$8T_{HCLK}+0.5$	ns
$t_{d(NOE\_NCEx)}$	FMC_NOE high to FMC_NCEx high	$5T_{HCLK}-1$	-	ns
$t_{su(D-NOE)}$	FMC_D[15:0] valid data before FMC_NOE high	$1T_{HCLK}$	-	ns
$t_{h(NOE-D)}$	FMC_NOE high to FMC_D[15:0] invalid	0	-	ns

1.  $C_L = 30$  pF.
2. Based on characterization, not tested in production.

**Table 101. Switching characteristics for PC Card/CF read and write cycles in I/O space<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
tw(NIOWR)	FMC_NIOWR low width	$8T_{HCLK}-0.5$	-	ns
tv(NIOWR-D)	FMC_NIOWR low to FMC_D[15:0] valid	-	0	ns
th(NIOWR-D)	FMC_NIOWR high to FMC_D[15:0] invalid	$9T_{HCLK}-2$	-	ns
td(NCE4_1-NIOWR)	FMC_NCE4_1 low to FMC_NIOWR valid	-	$5T_{HCLK}$	ns
th(NCEx-NIOWR)	FMC_NCEx high to FMC_NIOWR invalid	$5T_{HCLK}$	-	ns
td(NIORD-NCEx)	FMC_NCEx low to FMC_NIORD valid	-	$5T_{HCLK}$	ns
th(NCEx-NIORD)	FMC_NCEx high to FMC_NIORD) valid	$6T_{HCLK}+2$	-	ns
tw(NIORD)	FMC_NIORD low width	$8T_{HCLK}-0.5$	$8T_{HCLK}+0.5$	ns
tsu(D-NIORD)	FMC_D[15:0] valid before FMC_NIORD high	$1T_{HCLK}$	-	ns
td(NIORD-D)	FMC_D[15:0] valid after FMC_NIORD high	0	-	ns

1.  $C_L = 30$  pF.

2. Based on characterization, not tested in production.

### NAND controller waveforms and timings

[Figure 70](#) through [Figure 73](#) represent synchronous waveforms, and [Table 102](#) and [Table 103](#) provide the corresponding timings. The results shown in this table are obtained with the following FMC configuration:

- COM.FMC\_SetupTime = 0x01;
- COM.FMC\_WaitSetupTime = 0x03;
- COM.FMC\_HoldSetupTime = 0x02;
- COM.FMC\_HiZSetupTime = 0x01;
- ATT.FMC\_SetupTime = 0x01;
- ATT.FMC\_WaitSetupTime = 0x03;
- ATT.FMC\_HoldSetupTime = 0x02;
- ATT.FMC\_HiZSetupTime = 0x01;
- Bank = FMC\_Bank\_NAND;
- MemoryDataWidth = FMC\_MemoryDataWidth\_16b;
- ECC = FMC\_ECC\_Enable;
- ECCPageSize = FMC\_ECCPageSize\_512Bytes;
- TCLRSetupTime = 0;
- TARSetupTime = 0.

In all timing tables, the  $T_{HCLK}$  is the HCLK clock period.

Figure 70. NAND controller waveforms for read access

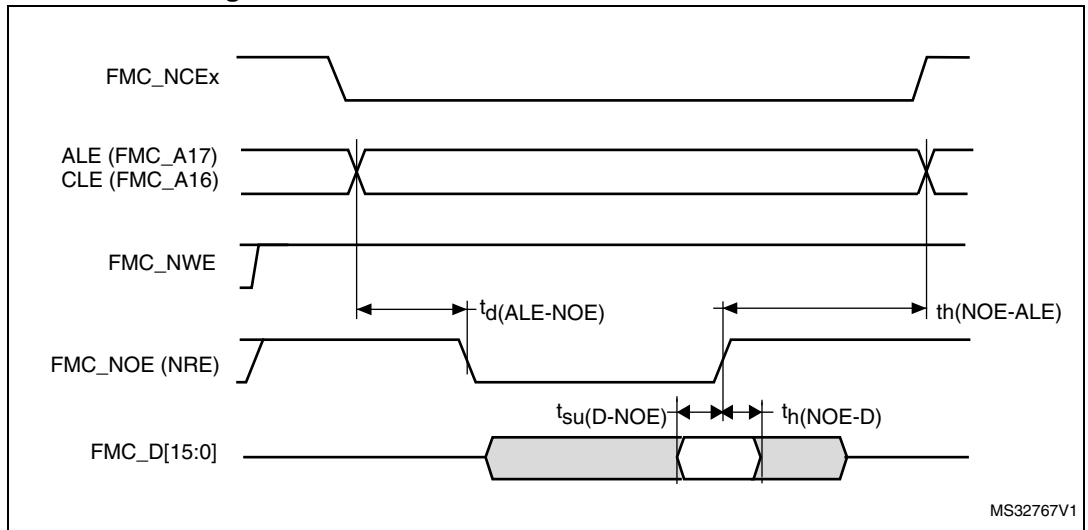


Figure 71. NAND controller waveforms for write access

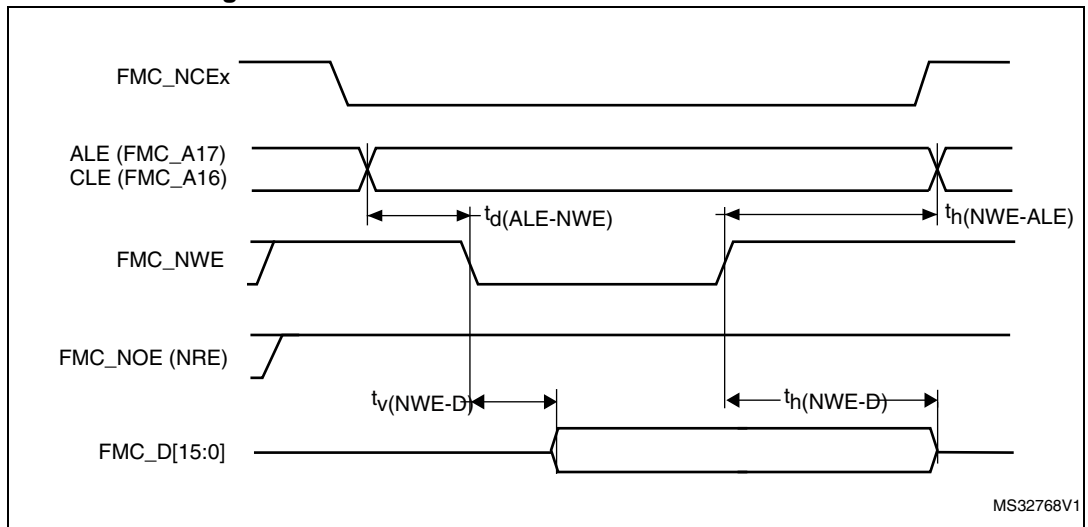
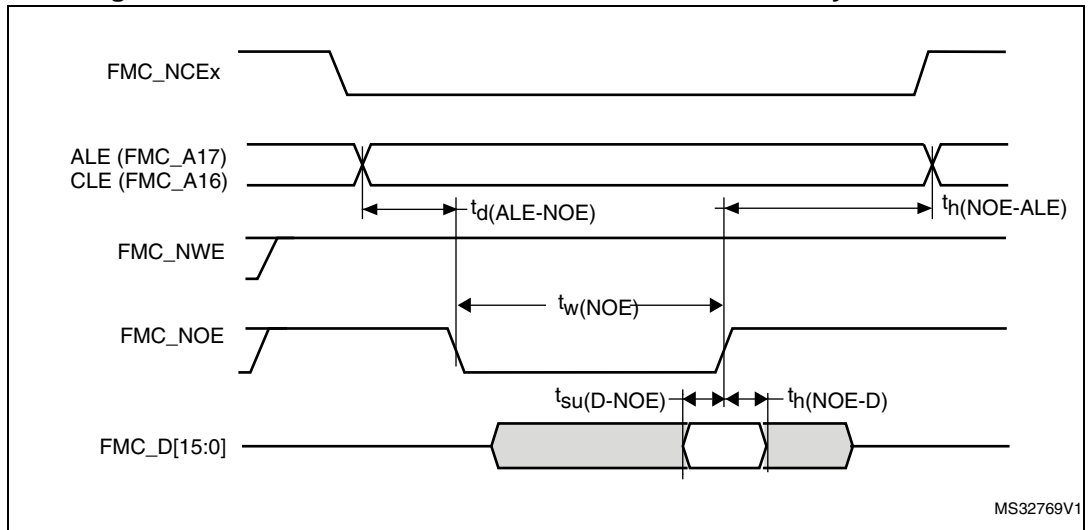
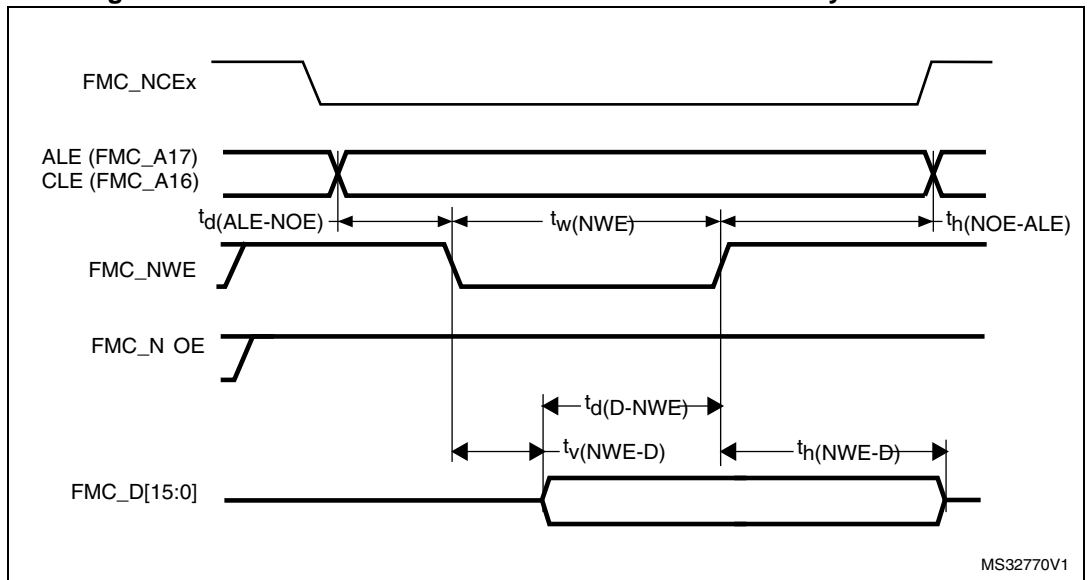


Figure 72. NAND controller waveforms for common memory read access



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Figure 73. NAND controller waveforms for common memory write access



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Table 102. Switching characteristics for NAND Flash read cycles<sup>(1)</sup>

Symbol	Parameter	Min	Max	Unit
$t_w(\text{NOE})$	FMC_NOE low width	$4T_{\text{HCLK}}-0.5$	$4T_{\text{HCLK}}+0.5$	ns
$t_{su}(\text{D-NOE})$	FMC_D[15-0] valid data before FMC_NOE high	9	-	ns
$t_h(\text{NOE-D})$	FMC_D[15-0] valid data after FMC_NOE high	0	-	ns
$t_d(\text{ALE-NOE})$	FMC_ALE valid before FMC_NOE low	-	$3T_{\text{HCLK}}-0.5$	ns
$t_h(\text{NOE-ALE})$	FMC_NWE high to FMC_ALE invalid	$3T_{\text{HCLK}}-2$	-	ns

1.  $C_L = 30 \text{ pF}$ .

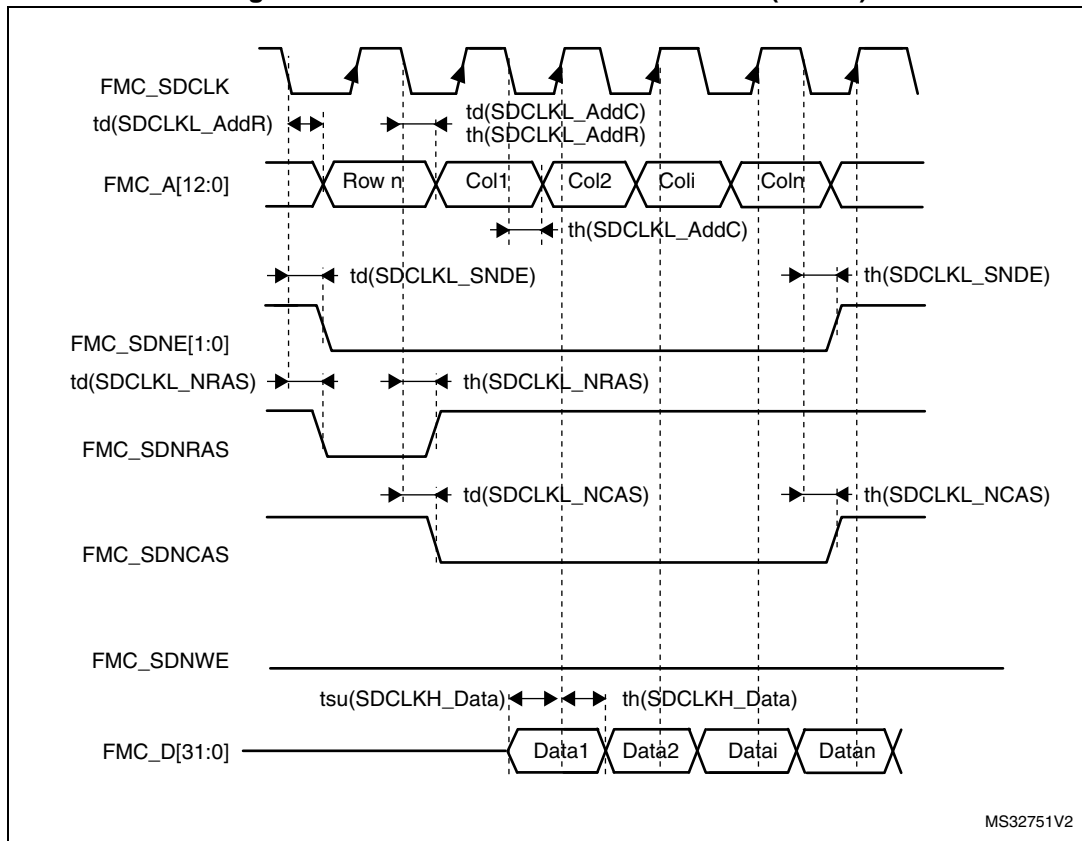
Table 103. Switching characteristics for NAND Flash write cycles<sup>(1)</sup>

Symbol	Parameter	Min	Max	Unit
$t_{w(NWE)}$	FMC_NWE low width	$4T_{HCLK}$	$4T_{HCLK}+1$	ns
$t_{v(NWE-D)}$	FMC_NWE low to FMC_D[15-0] valid	0	-	ns
$t_{h(NWE-D)}$	FMC_NWE high to FMC_D[15-0] invalid	$3T_{HCLK}-1$	-	ns
$t_{d(D-NWE)}$	FMC_D[15-0] valid before FMC_NWE high	$5T_{HCLK}-3$	-	ns
$t_{d(ALE-NWE)}$	FMC_ALE valid before FMC_NWE low	-	$3T_{HCLK}-0.5$	ns
$t_{h(NWE-ALE)}$	FMC_NWE high to FMC_ALE invalid	$3T_{HCLK}-1$	-	ns

1.  $C_L = 30$  pF.

SDRAM waveforms and timings

Figure 74. SDRAM read access waveforms (CL = 1)



**Table 104. SDRAM read timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(SDCLK)}$	FMC_SDCLK period	$2T_{HCLK}-0.5$	$2T_{HCLK}+0.5$	ns
$t_{su(SDCLKH\_Data)}$	Data input setup time	2	-	
$t_{h(SDCLKH\_Data)}$	Data input hold time	0	-	
$t_{d(SDCLKL\_Add)}$	Address valid time	-	1.5	
$t_{d(SDCLKL\_SDNE)}$	Chip select valid time	-	0.5	
$t_{h(SDCLKL\_SDNE)}$	Chip select hold time	0	-	
$t_{d(SDCLKL\_SDNRAS)}$	SDNRAS valid time	-	0.5	
$t_{h(SDCLKL\_SDNRAS)}$	SDNRAS hold time	0	-	
$t_{d(SDCLKL\_SDNCAS)}$	SDNCAS valid time	-	0.5	
$t_{h(SDCLKL\_SDNCAS)}$	SDNCAS hold time	0	-	

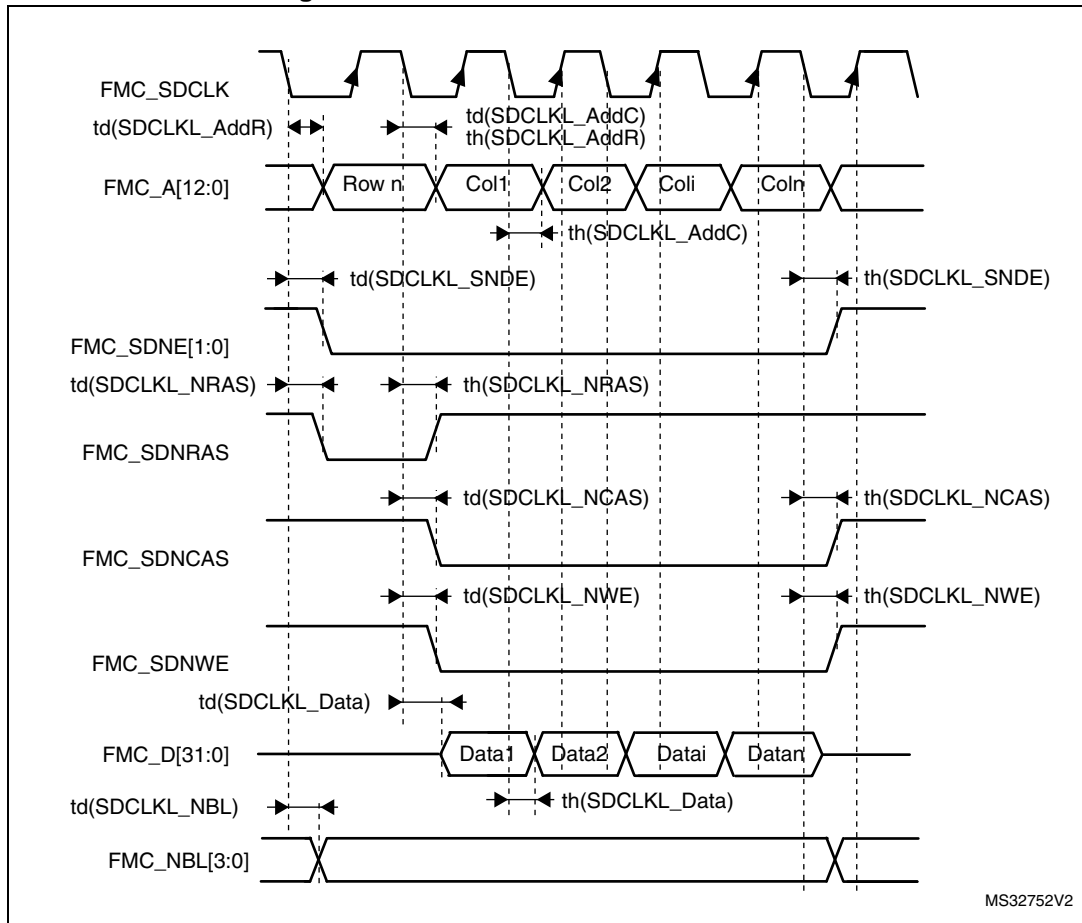
1. CL = 30 pF on data and address lines. CL=15pF on FMC\_SDCLK.
2. Data based on characterization results, not tested in production.

**Table 105. LPDDR SDRAM read timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(SDCLK)}$	FMC_SDCLK period	$2T_{HCLK}-0.5$	$2T_{HCLK}+0.5$	ns
$t_{su(SDCLKH\_Data)}$	Data input setup time	2.5	-	
$t_{h(SDCLKH\_Data)}$	Data input hold time	0	-	
$t_{d(SDCLKL\_Add)}$	Address valid time	-	1	
$t_{d(SDCLKL\_SDNE)}$	Chip select valid time	-	1	
$t_{h(SDCLKL\_SDNE)}$	Chip select hold time	1	-	
$t_{d(SDCLKL\_SDNRAS)}$	SDNRAS valid time	-	1	
$t_{h(SDCLKL\_SDNRAS)}$	SDNRAS hold time	1	-	
$t_{d(SDCLKL\_SDNCAS)}$	SDNCAS valid time	-	1	
$t_{h(SDCLKL\_SDNCAS)}$	SDNCAS hold time	1	-	

1. CL = 10 pF.
2. Data based on characterization results, not tested in production.

Figure 75. SDRAM write access waveforms



**Table 106. SDRAM write timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(SDCLK)}$	FMC_SDCLK period	$2T_{HCLK}-0.5$	$2T_{HCLK}+0.5$	ns
$t_{d(SDCLKL\_Data)}$	Data output valid time	-	2.5	
$t_{h(SDCLKL\_Data)}$	Data output hold time	3.5	-	
$t_{d(SDCLKL\_Add)}$	Address valid time	-	1.5	
$t_{d(SDCLKL\_SDNWE)}$	SDNWE valid time	-	1	
$t_{h(SDCLKL\_SDNWE)}$	SDNWE hold time	0	-	
$t_{d(SDCLKL\_SDNE)}$	Chip select valid time	-	0.5	
$t_{h(SDCLKL\_SDNE)}$	Chip select hold time	0	-	
$t_{d(SDCLKL\_SDNRAS)}$	SDNRAS valid time	-	2	
$t_{h(SDCLKL\_SDNRAS)}$	SDNRAS hold time	0	-	
$t_{d(SDCLKL\_SDNCAS)}$	SDNCAS valid time	-	0.5	
$t_{d(SDCLKL\_SDNCAS)}$	SDNCAS hold time	0	-	
$t_{d(SDCLKL\_NBL)}$	NBL valid time	-	0.5	
$t_{h(SDCLKL\_NBL)}$	NBLoutput time	0	-	

1. CL = 30 pF on data and address lines. CL=15pF on FMC\_SDCLK.
2. Data based on characterization results, not tested in production.

**Table 107. LPSDR SDRAM write timings<sup>(1)(2)</sup>**

Symbol	Parameter	Min	Max	Unit
$t_{w(SDCLK)}$	FMC_SDCLK period	$2T_{HCLK}-0.5$	$2T_{HCLK}+0.5$	ns
$t_{d(SDCLKL\_Data)}$	Data output valid time	-	5	
$t_{h(SDCLKL\_Data)}$	Data output hold time	2	-	
$t_{d(SDCLKL\_Add)}$	Address valid time	-	2.8	
$t_{d(SDCLKL-SDNWE)}$	SDNWE valid time	-	2	
$t_{h(SDCLKL-SDNWE)}$	SDNWE hold time	1	-	
$t_{d(SDCLKL-SDNE)}$	Chip select valid time	-	1.5	
$t_{h(SDCLKL-SDNE)}$	Chip select hold time	1	-	
$t_{d(SDCLKL-SDNRAS)}$	SDNRAS valid time	-	1.5	
$t_{h(SDCLKL-SDNRAS)}$	SDNRAS hold time	1.5	-	
$t_{d(SDCLKL-SDNCAS)}$	SDNCAS valid time	-	1.5	
$t_{d(SDCLKL-SDNCAS)}$	SDNCAS hold time	1.5	-	
$t_{d(SDCLKL\_NBL)}$	NBL valid time	-	1.5	
$t_{h(SDCLKL\_NBL)}$	NBLoutput time	1.5	-	

1. CL = 10 pF.

2. Data based on characterization results, not tested in production.

### 6.3.27 Camera interface (DCMI) timing specifications

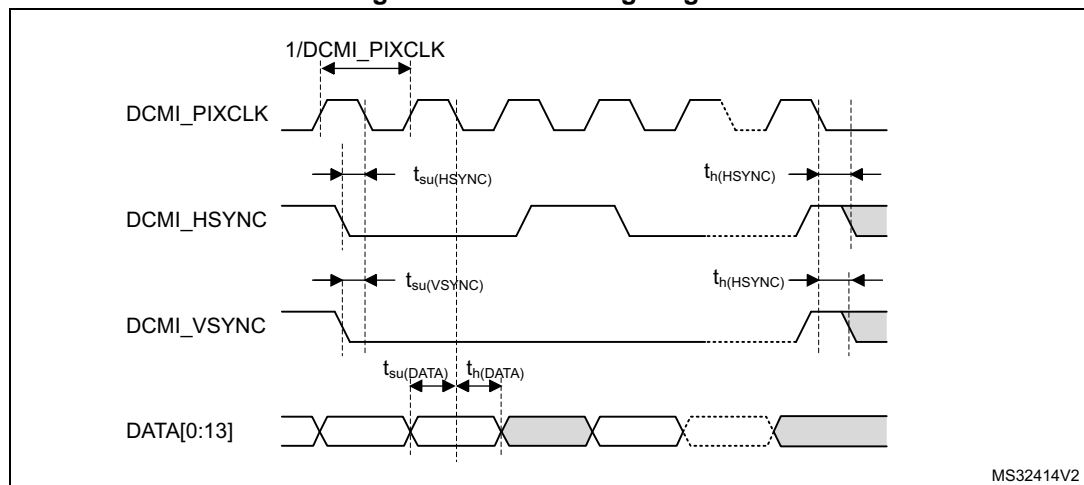
Unless otherwise specified, the parameters given in [Table 108](#) for DCMI are derived from tests performed under the ambient temperature,  $f_{HCLK}$  frequency and  $V_{DD}$  supply voltage summarized in [Table 17](#), with the following configuration:

- DCMI\_PIXCLK polarity: falling
- DCMI\_VSYNC and DCMI\_HSYNC polarity: high
- Data formats: 14 bits

**Table 108. DCMI characteristics**

Symbol	Parameter	Min	Max	Unit
	Frequency ratio DCMI_PIXCLK/ $f_{HCLK}$	-	0.4	
DCMI_PIXCLK	Pixel clock input	-	54	MHz
$D_{Pixel}$	Pixel clock input duty cycle	30	70	%
$t_{su}(DATA)$	Data input setup time	2	-	ns
$t_h(DATA)$	Data input hold time	2.5	-	
$t_{su}(HSYNC)$ $t_{su}(VSYNC)$	DCMI_HSYNC/DCMI_VSYNC input setup time	0.5	-	
$t_h(HSYNC)$ $t_h(VSYNC)$	DCMI_HSYNC/DCMI_VSYNC input hold time	1	-	

**Figure 76. DCMI timing diagram**



### 6.3.28 LCD-TFT controller (LTDC) characteristics

Unless otherwise specified, the parameters given in [Table 109](#) for LCD-TFT are derived from tests performed under the ambient temperature,  $f_{HCLK}$  frequency and VDD supply voltage summarized in [Table 17](#), with the following configuration:

- LCD\_CLK polarity: high
- LCD\_DE polarity : low
- LCD\_VSYNC and LCD\_HSYNC polarity: high
- Pixel formats: 24 bits

**Table 109. LTDC characteristics**

Symbol	Parameter	Min	Max	Unit
$f_{CLK}$	LTDC clock output frequency	-	42	MHz
$D_{CLK}$	LTDC clock output duty cycle	45	55	%
$t_{w(CLKH)}$ $t_{w(CLKL)}$	Clock High time, low time	$tw(CLK)/2-0.5$	$tw(CLK)/2+0.5$	ns
$t_{v(DATA)}$	Data output valid time	-	3.5	
$t_{h(DATA)}$	Data output hold time	1.5	-	
$t_{v(HSYNC)}$	HSYNC/VSYNC/DE output valid time	-	2.5	
$t_{v(VSYNC)}$		-		
$t_{v(DE)}$		-		
$t_{h(HSYNC)}$	HSYNC/VSYNC/DE output hold time	2	-	
$t_{h(VSYNC)}$				
$t_{h(DE)}$				

Figure 77. LCD-TFT horizontal timing diagram

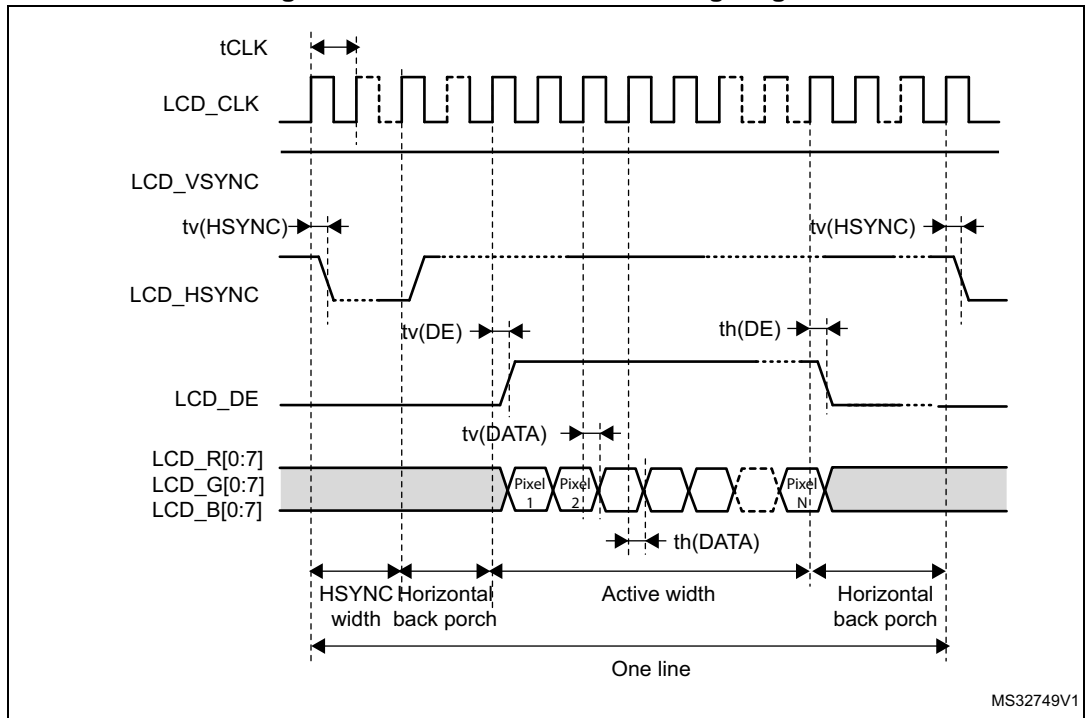
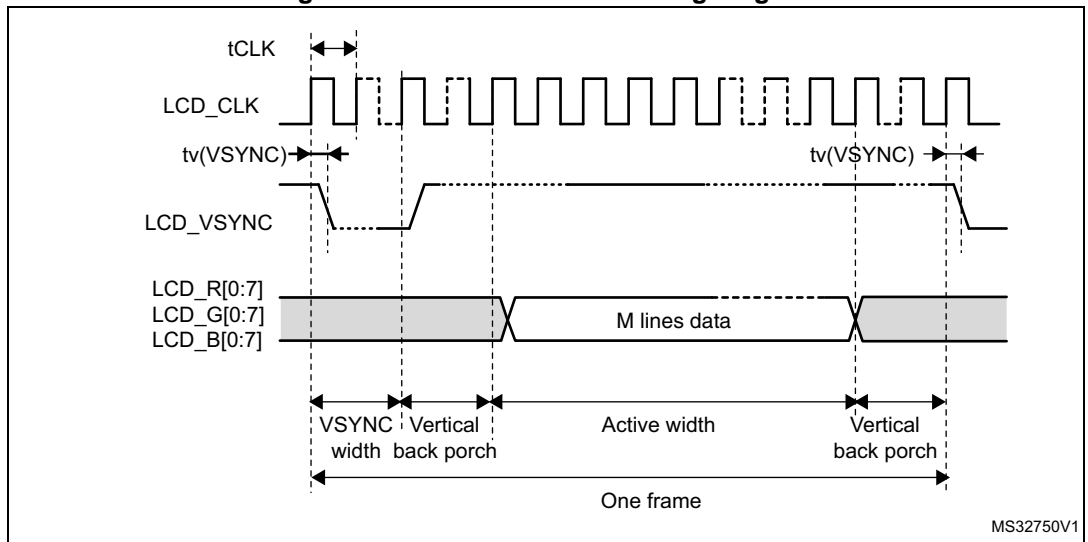


Figure 78. LCD-TFT vertical timing diagram



### 6.3.29 SD/SDIO MMC card host interface (SDIO) characteristics

Unless otherwise specified, the parameters given in [Table 110](#) for the SDIO/MMC interface are derived from tests performed under the ambient temperature,  $f_{PCLK2}$  frequency and  $V_{DD}$  supply voltage conditions summarized in [Table 17](#), with the following configuration:

- Output speed is set to  $OSPEEDRy[1:0] = 10$
- Capacitive load  $C = 30\text{ pF}$
- Measurement points are done at CMOS levels:  $0.5V_{DD}$

Refer to [Section 6.3.17: I/O port characteristics](#) for more details on the input/output characteristics.

Figure 79. SDIO high-speed mode

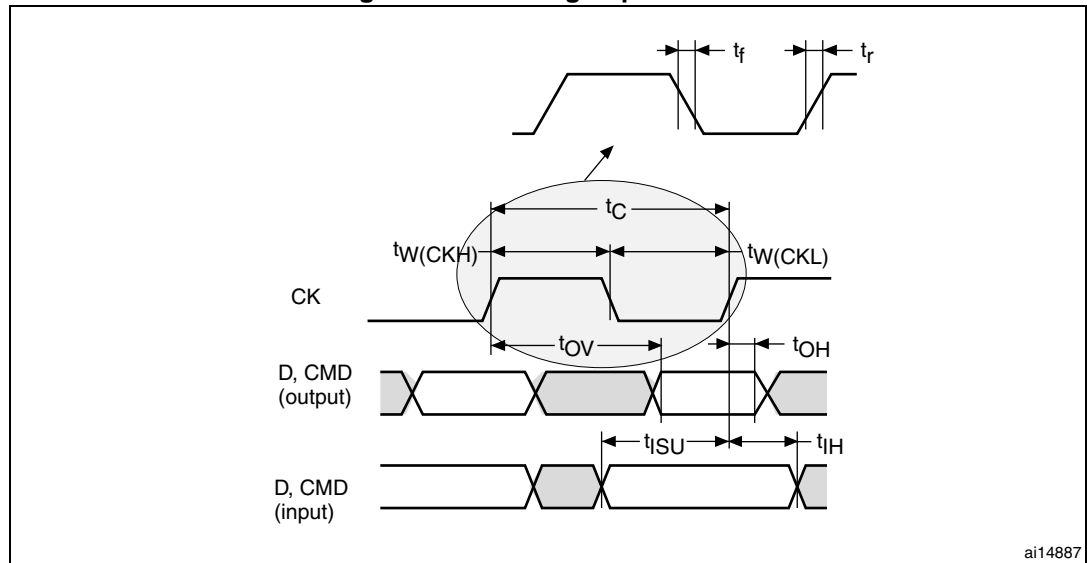
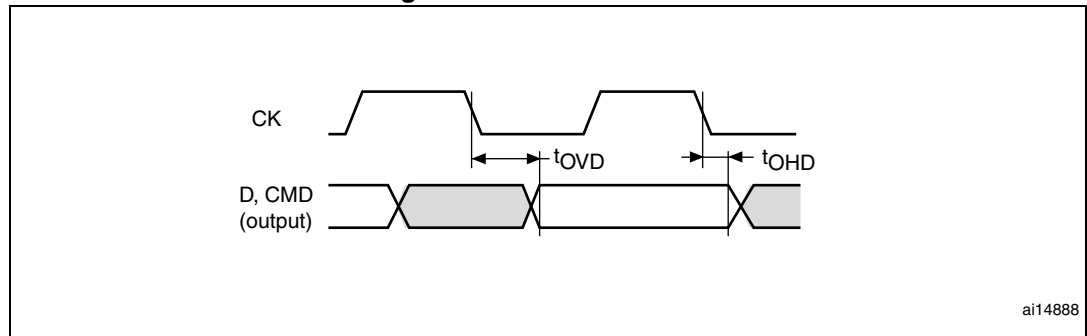


Figure 80. SD default mode



**Table 110. Dynamic characteristics: SD / MMC characteristics<sup>(1)(2)</sup>**

Symbol	Parameter	Conditions	Min	Typ	Max	Unit
f <sub>PP</sub>	Clock frequency in data transfer mode		0		48	MHz
-	SDIO_CK/fPCLK2 frequency ratio		-	-	8/3	-
t <sub>W(CKL)</sub>	Clock low time	f <sub>pp</sub> =48MHz	8.5	9	-	ns
t <sub>W(CKH)</sub>	Clock high time	f <sub>pp</sub> =48MHz	8.3	10	-	
<b>CMD, D inputs (referenced to CK) in MMC and SD HS mode</b>						
t <sub>ISU</sub>	Input setup time HS	f <sub>pp</sub> =48MHz	3.5	-	-	ns
t <sub>IH</sub>	Input hold time HS	f <sub>pp</sub> =48MHz	0	-	-	
<b>CMD, D outputs (referenced to CK) in MMC and SD HS mode</b>						
t <sub>OV</sub>	Output valid time HS	f <sub>pp</sub> =48MHz	-	4.5	7	ns
t <sub>OH</sub>	Output hold time HS	f <sub>pp</sub> =48MHz	3	-	-	
<b>CMD, D inputs (referenced to CK) in SD default mode</b>						
t <sub>ISUD</sub>	Input setup time SD	f <sub>pp</sub> =24MHz	1.5	-	-	ns
t <sub>IHD</sub>	Input hold time SD	f <sub>pp</sub> =24MHz	0.5	-	-	
<b>CMD, D outputs (referenced to CK) in SD default mode</b>						
t <sub>OVD</sub>	Output valid default time SD	f <sub>pp</sub> =24MHz	-	4.5	6.5	ns
t <sub>OHD</sub>	Output hold default time SD	f <sub>pp</sub> =24MHz	3.5	-	-	

1. Data based on characterization results, not tested in production.

2. V<sub>DD</sub> = 2.7 to 3.6 V.

### 6.3.30 RTC characteristics

**Table 111. RTC characteristics**

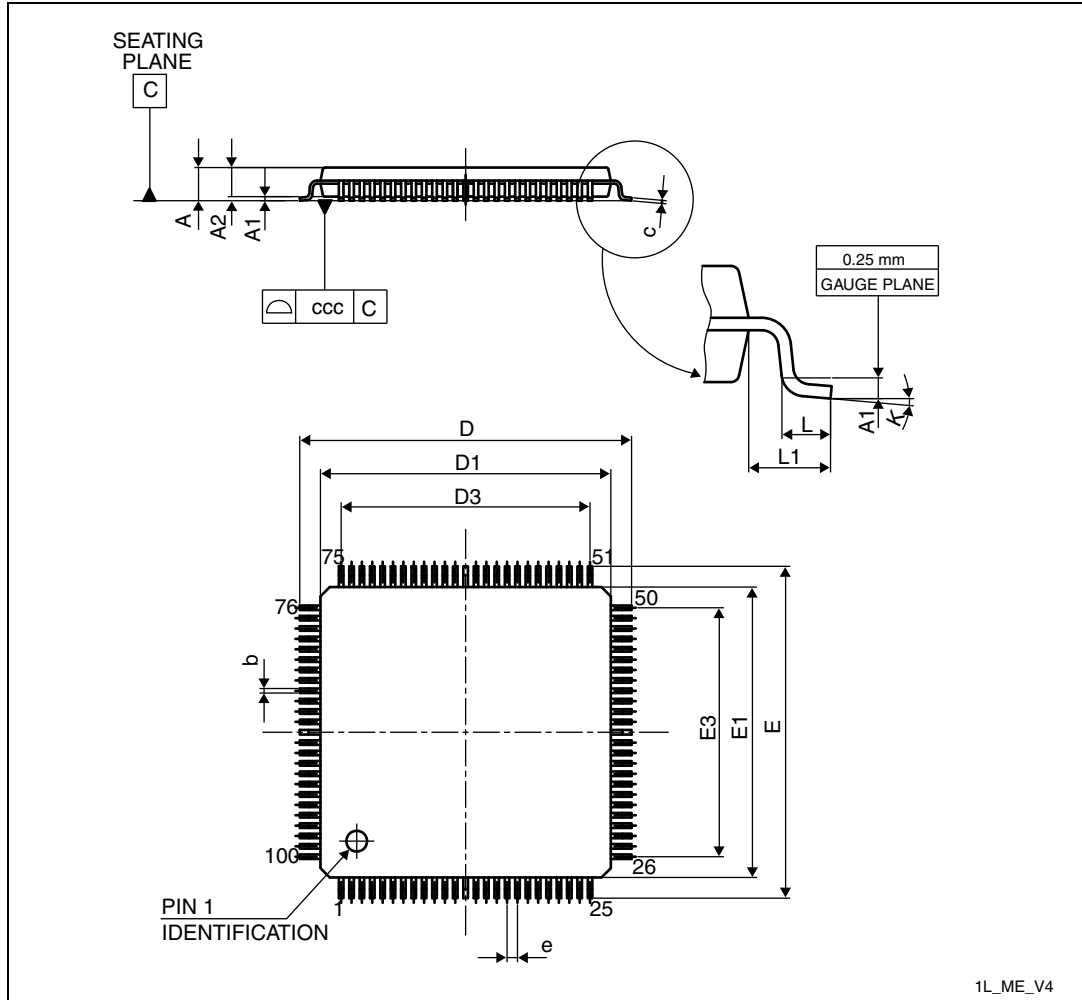
Symbol	Parameter	Conditions	Min	Max
-	f <sub>PCLK1</sub> /RTCCLK frequency ratio	Any read/write operation from/to an RTC register	4	-

## 7 Package characteristics

### 7.1 Package mechanical data

In order to meet environmental requirements, ST offers these devices in different grades of ECOPACK<sup>®</sup> packages, depending on their level of environmental compliance. ECOPACK<sup>®</sup> specifications, grade definitions and product status are available at: [www.st.com](http://www.st.com). ECOPACK<sup>®</sup> is an ST trademark.

Figure 81. LQFP100, 14 x 14 mm 100-pin low-profile quad flat package outline



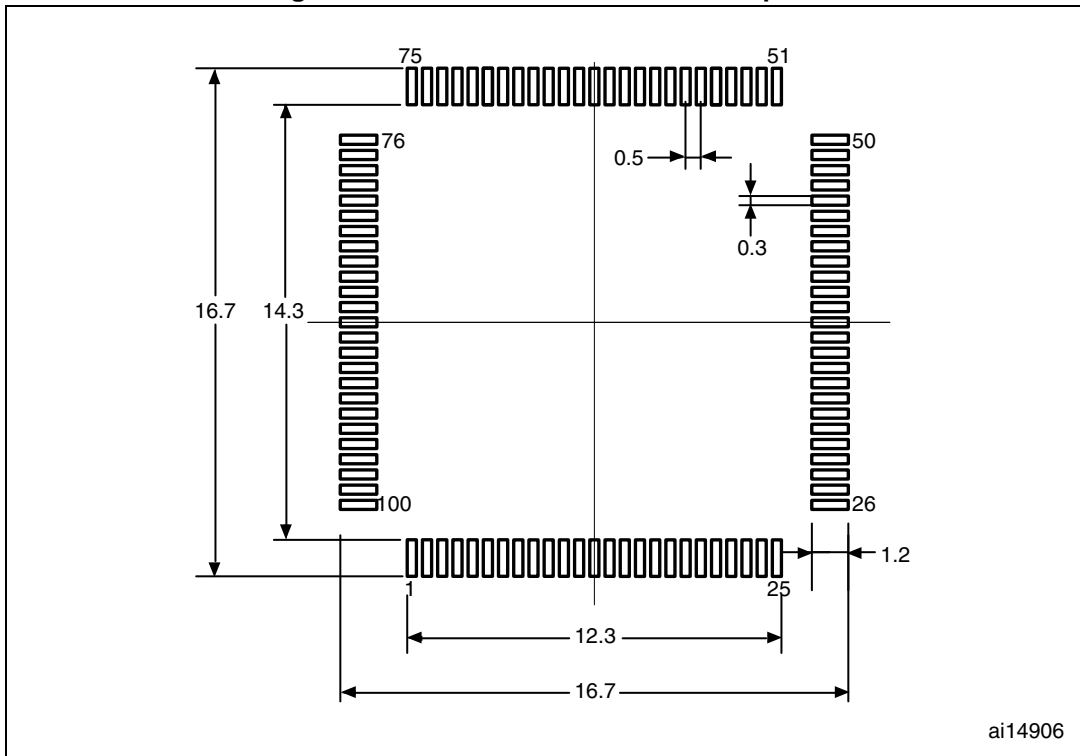
1. Drawing is not to scale.

Table 112. LQPF100, 14 x 14 mm 100-pin low-profile quad flat package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	15.800	16.000	16.200	0.6220	0.6299	0.6378
D1	13.800	14.000	14.200	0.5433	0.5512	0.5591
D3	-	12.000	-	-	0.4724	-
E	15.800	16.000	16.200	0.6220	0.6299	0.6378
E1	13.800	14.000	14.200	0.5433	0.5512	0.5591
E3	-	12.000	-	-	0.4724	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

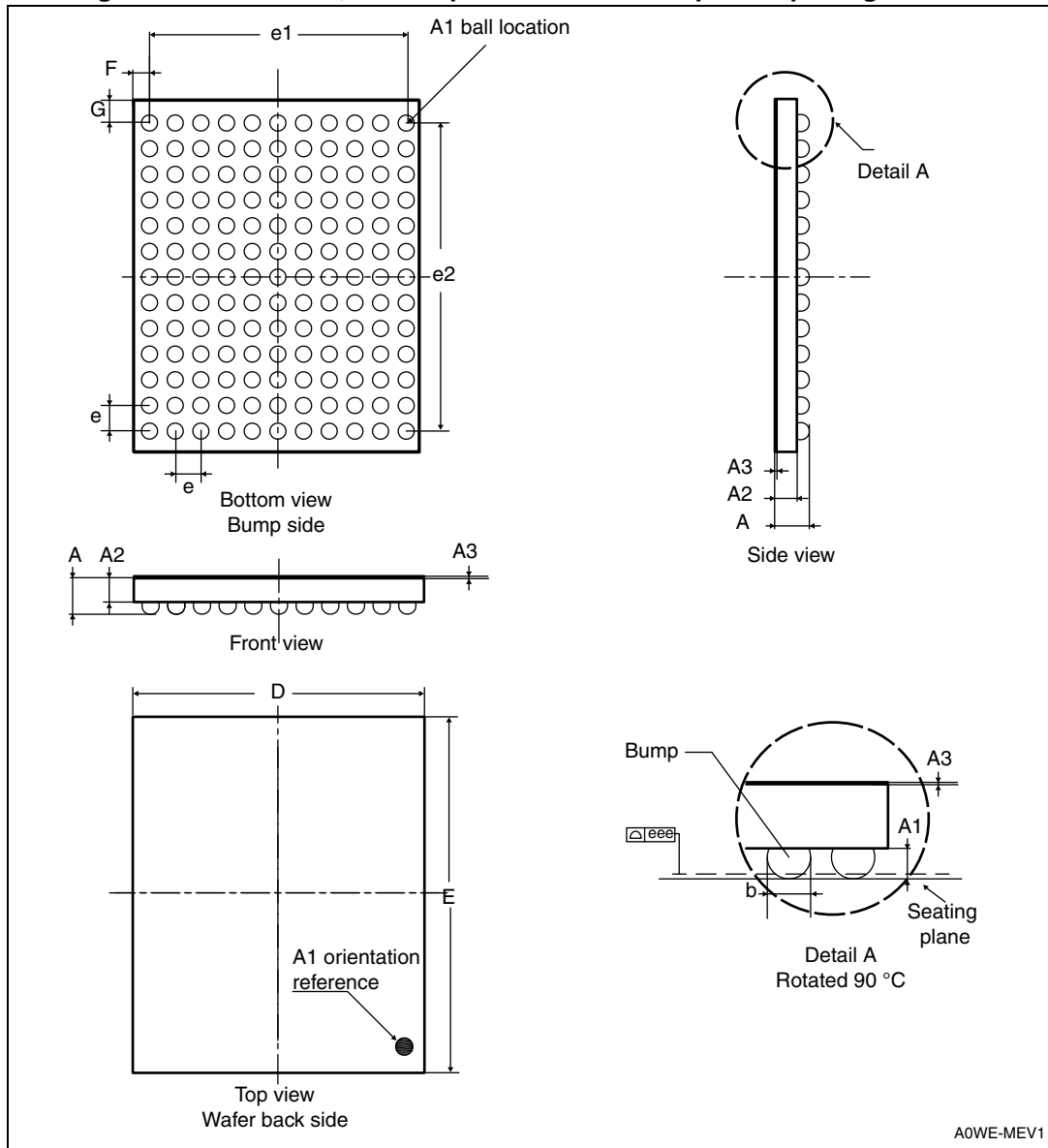
1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 82. LQPF100 recommended footprint



1. Dimensions are expressed in millimeters.

Figure 83. WLCSP143, 0.4 mm pitch wafer level chip scale package outline



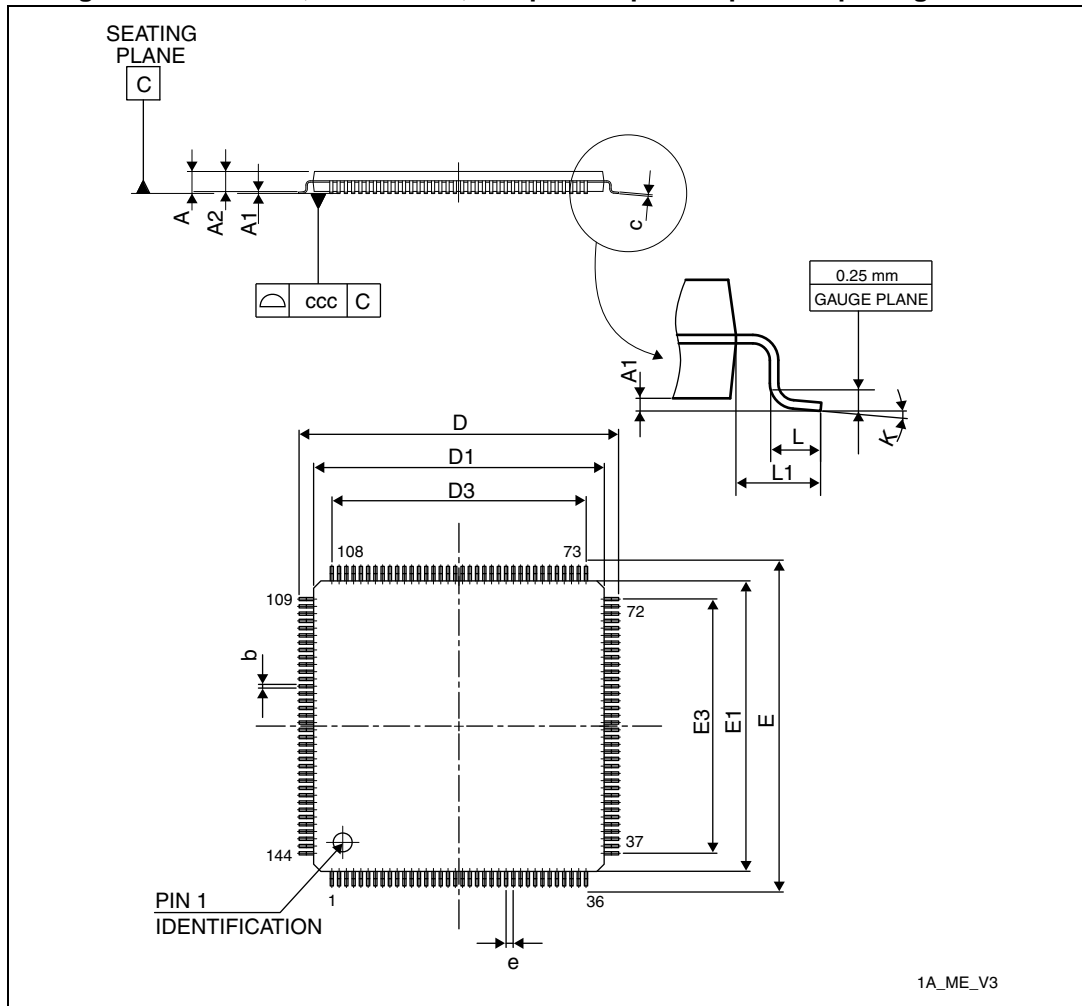
1. Drawing is not to scale.

Table 113. WLCSP143, 0.4 mm pitch wafer level chip scale package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
A	0.525	0.555	0.585	0.0207	0.0219	0.0230
A1	-	0.175	-	-	0.0069	-
A2	-	0.380	-	-	0.0150	-
A3	0.220	0.025	0.280	0.0087	0.0010	0.0110
b	-	0.250°	-	-	0.250°	-
D	4.486	4.521	4.556	0.1766	0.1780	0.1794
E	5.512	5.547	5.582	0.2170	0.2184	0.2198
e	-	0.400	-	-	0.0157	-
e1	-	4.000	-	-	0.1575	-
e2	-	4.800	-	-	0.1890	-
F	-	0.261	-	-	0.0103	-
G	-	0.374	-	-	0.0147	-
eee	-	0.050	-	-	0.0020	-

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 84. LQFP144, 20 x 20 mm, 144-pin low-profile quad flat package outline



1A\_ME\_V3

1. Drawing is not to scale.

Table 114. LQFP144, 20 x 20 mm, 144-pin low-profile quad flat package mechanical data

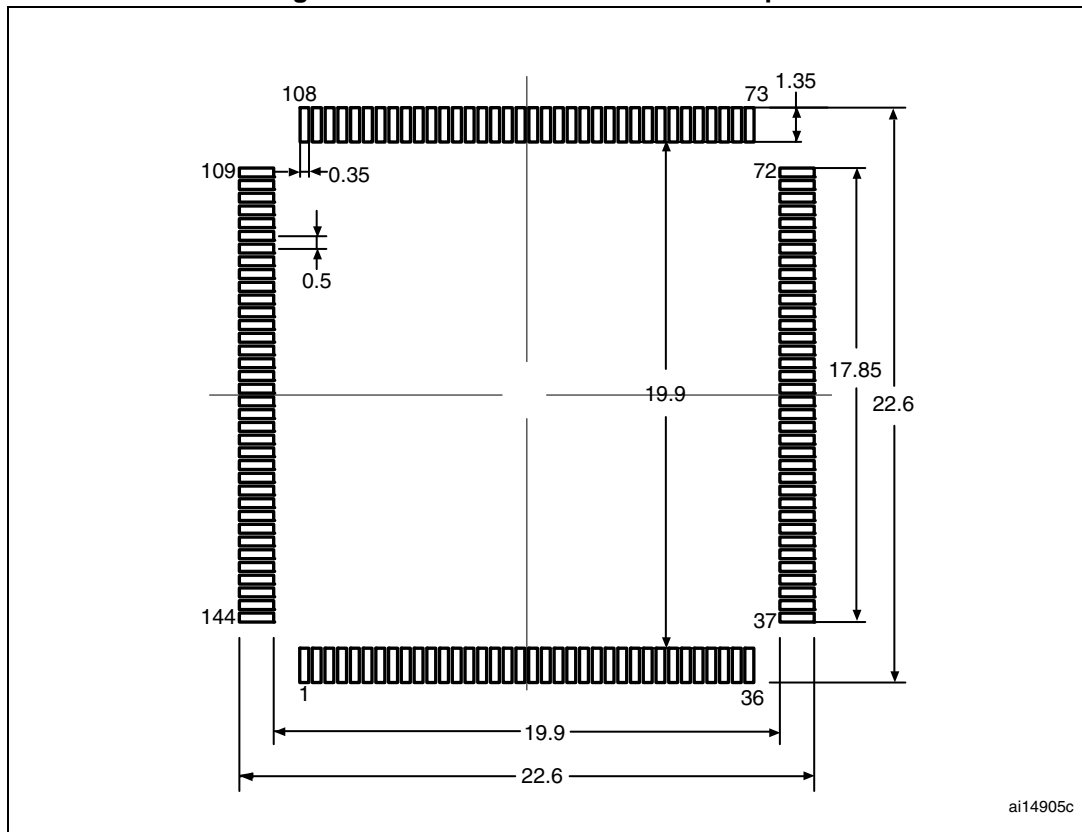
Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	21.800	22.000	22.200	0.8583	0.8661	0.874
D1	19.800	20.000	20.200	0.7795	0.7874	0.7953
D3	-	17.500	-	-	0.689	-

**Table 114. LQFP144, 20 x 20 mm, 144-pin low-profile quad flat package mechanical data (continued)**

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
E	21.800	22.000	22.200	0.8583	0.8661	0.8740
E1	19.800	20.000	20.200	0.7795	0.7874	0.7953
E3	-	17.500	-	-	0.6890	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7°	0°	3.5°	7°
ccc	-	-	0.080	-	-	0.0031

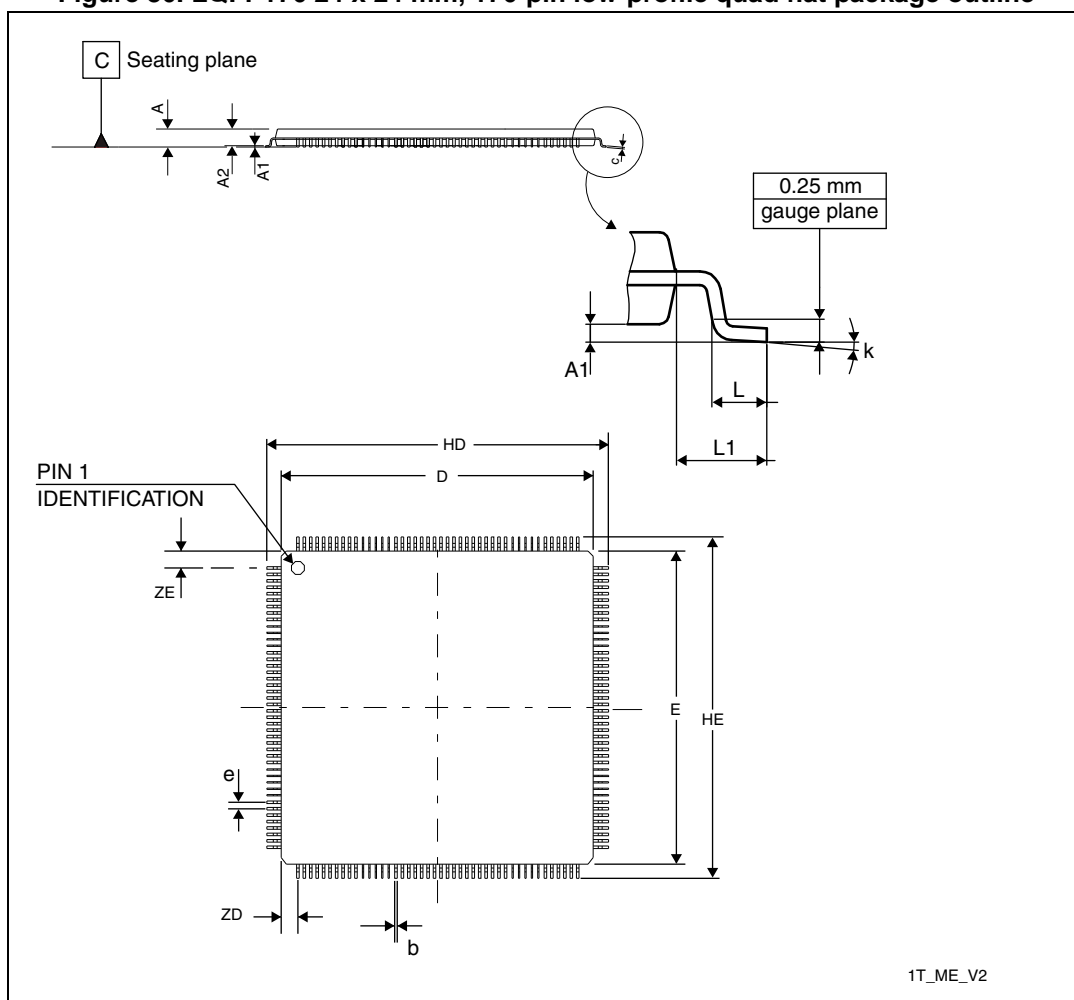
1. Values in inches are converted from mm and rounded to 4 decimal digits.

**Figure 85. LQFP144 recommended footprint**



1. Dimensions are expressed in millimeters.

Figure 86. LQFP176 24 x 24 mm, 176-pin low-profile quad flat package outline



1. Drawing is not to scale.

Table 115. LQFP176, 24 x 24 mm, 176-pin low-profile quad flat package mechanical data

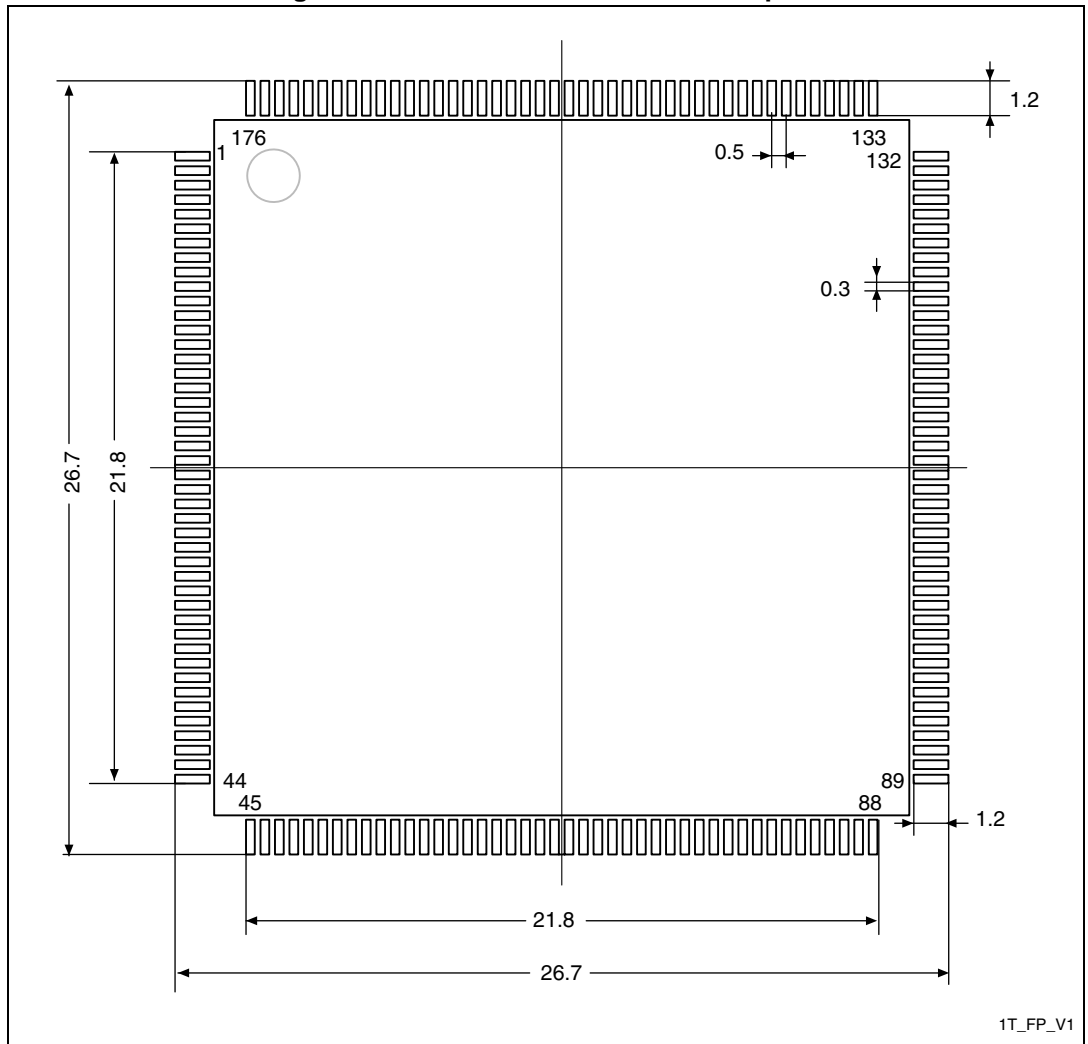
Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	-	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	-	1.450	0.0531	-	0.0060
b	0.170	-	0.270	0.0067	-	0.0106
C	0.090	-	0.200	0.0035	-	0.0079
D	23.900	-	24.100	0.9409	-	0.9488
E	23.900	-	24.100	0.9409	-	0.9488
e	-	0.500	-	-	0.0197	-
HD	25.900	-	26.100	1.0200	-	1.0276

**Table 115. LQFP176, 24 x 24 mm, 176-pin low-profile quad flat package  
mechanical data (continued)**

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
HE	25.900	-	26.100	1.0200	-	1.0276
L	0.450	-	0.750	0.0177	-	0.0295
L1	-	1.000	-	-	0.0394	-
ZD	-	1.250	-	-	0.0492	-
ZE	-	1.250	-	-	0.0492	-
ccc	-	-	0.080	-	-	0.0031
k	0 °	-	7 °	0 °	-	7 °

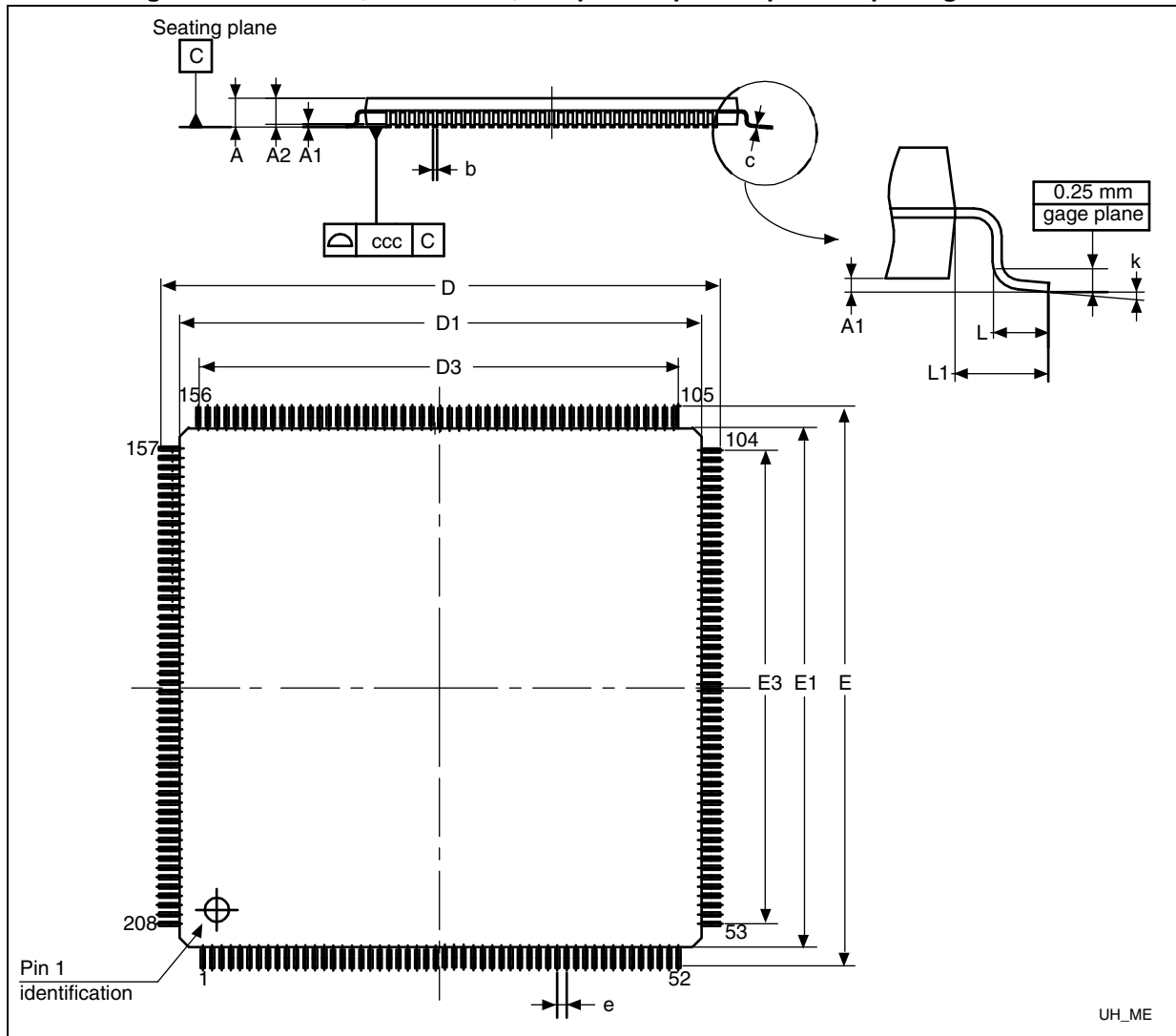
1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 87. LQFP176 recommended footprint



1. Dimensions are expressed in millimeters.

Figure 88. LQFP208, 28 x 28 mm, 208-pin low-profile quad flat package outline



1. Drawing is not to scale.

Table 116. LQFP208, 28 x 28 mm, 208-pin low-profile quad flat package mechanical data

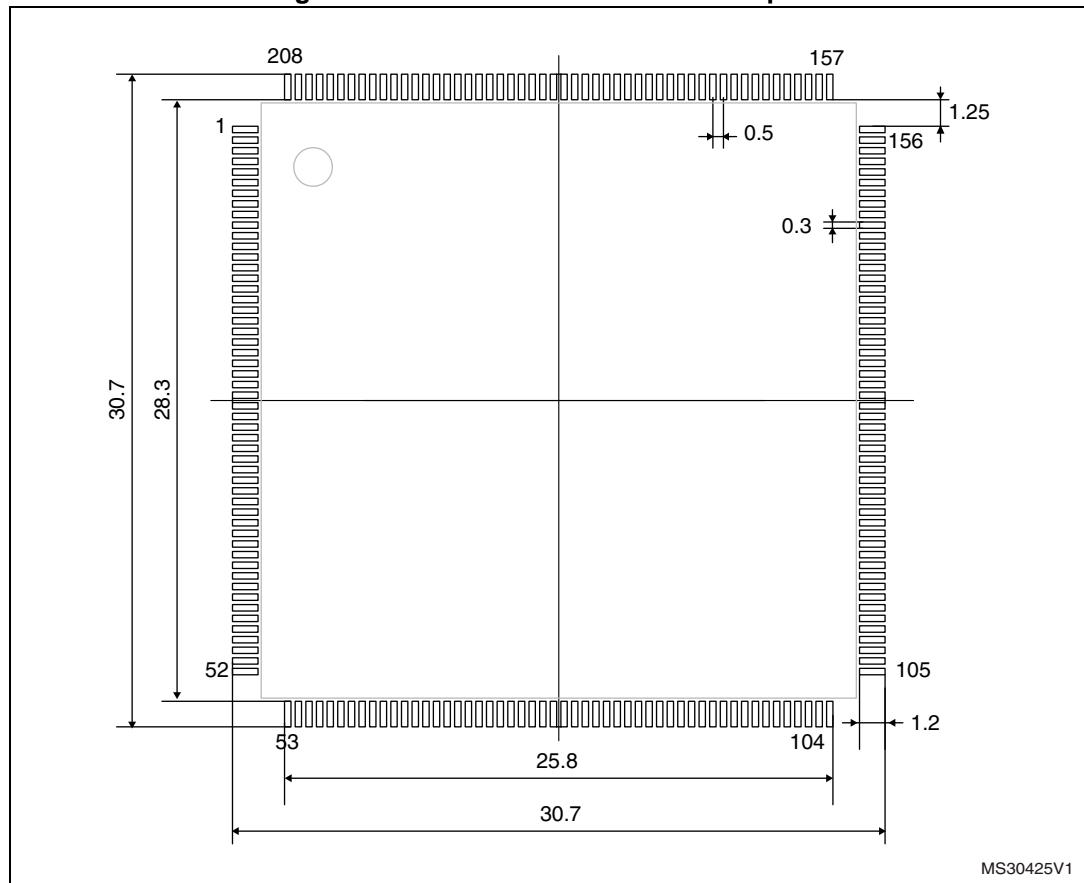
Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.600	--	-	0.0630
A1	0.050	-	0.150	0.0020	-	0.0059
A2	1.350	1.400	1.450	0.0531	0.0551	0.0571
b	0.170	0.220	0.270	0.0067	0.0087	0.0106
c	0.090	-	0.200	0.0035	-	0.0079
D	29.800	30.000	30.200	1.1732	1.1811	1.1890
D1	27.800	28.000	28.200	1.0945	1.1024	1.1102

**Table 116. LQFP208, 28 x 28 mm, 208-pin low-profile quad flat package mechanical data (continued)**

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
D3	-	25.500	-	-	1.0039	-
E	29.800	30.000	30.200	1.1732	1.1811	1.1890
E1	27.800	28.000	28.200	1.0945	1.1024	1.1102
E3	-	25.500	-	-	1.0039	-
e	-	0.500	-	-	0.0197	-
L	0.450	0.600	0.750	0.0177	0.0236	0.0295
L1	-	1.000	-	-	0.0394	-
k	0°	3.5°	7.0°	0°	3.5°	7.0°
ccc	-	-	0.080	-	-	0.0031

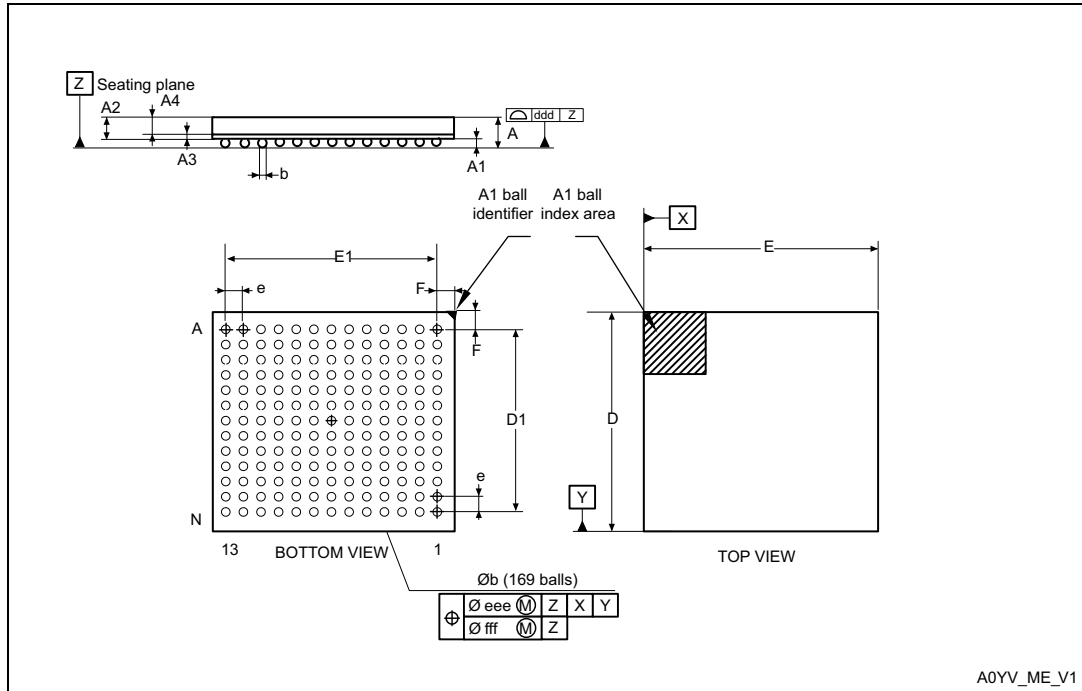
1. Values in inches are converted from mm and rounded to 4 decimal digits.

**Figure 89. LQFP208 recommended footprint**



1. Dimensions are expressed in millimeters.

Figure 90. UFBGA169 - ultra thin fine pitch ball grid array 7 x 7 mm, 0.6 mm, package outline

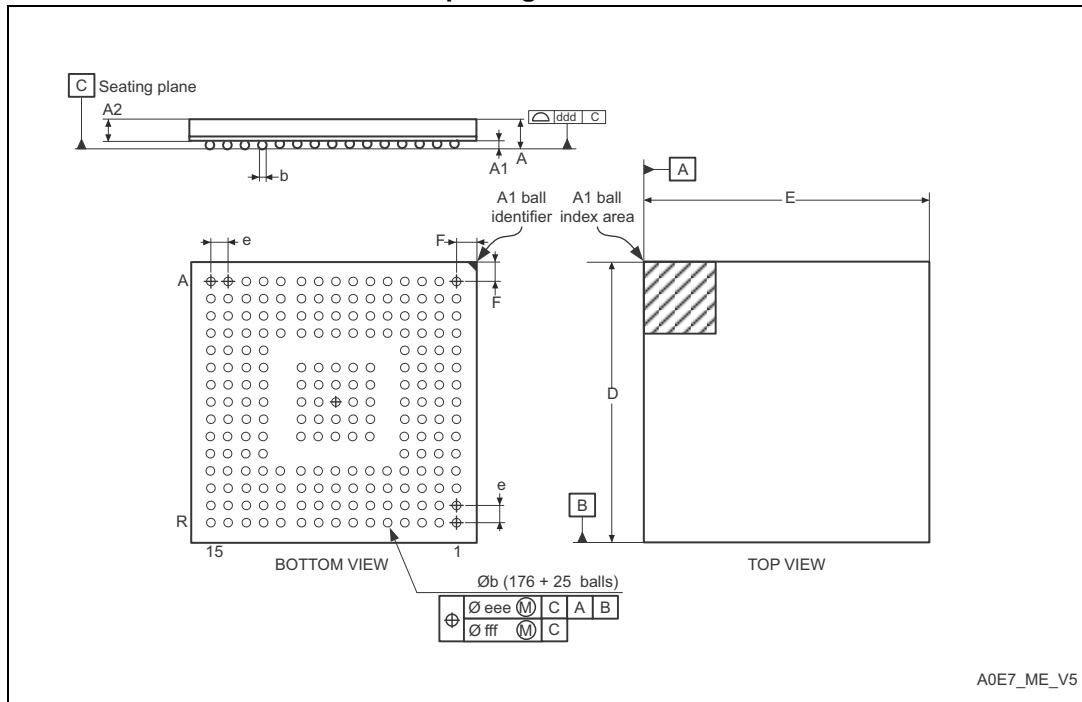


1. Drawing is not to scale.

Table 117. UFBGA169 - ultra thin fine pitch ball grid array 7 x 7 x 0.6 mm mechanical data

Symbol	millimeters			inches		
	Min	Typ	Max	Min	Typ	Max
A	0.460	0.530	0.600	0.0181	0.0209	0.0236
A1	0.050	0.080	0.110	0.0020	0.0031	0.0043
A2	0.400	0.450	0.500	0.0157	0.0177	0.0197
A3	0.080	0.130	0.180	0.0031	0.0051	0.0071
A4	0.270	0.320	0.370	0.0106	0.0126	0.0146
b	0.170	0.280	0.330	0.0067	0.0110	0.0130
D	6.900	7.000	7.100	0.2717	0.2756	0.2795
D1	5.950	6.000	6.050	0.2343	0.2362	0.2382
E	6.900	7.000	7.100	0.2717	0.2756	0.2795
E1	5.950	6.000	6.050	0.2343	0.2362	0.2382
e	-	0.500	-	-	0.0197	-
F	0.450	0.500	0.550	0.0177	0.0197	0.0217
ddd	-	-	0.080	-	-	0.0031
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.050	-	-	0.0020

Figure 91. UFBGA176+25 - ultra thin fine pitch ball grid array 10 × 10 × 0.6 mm, package outline



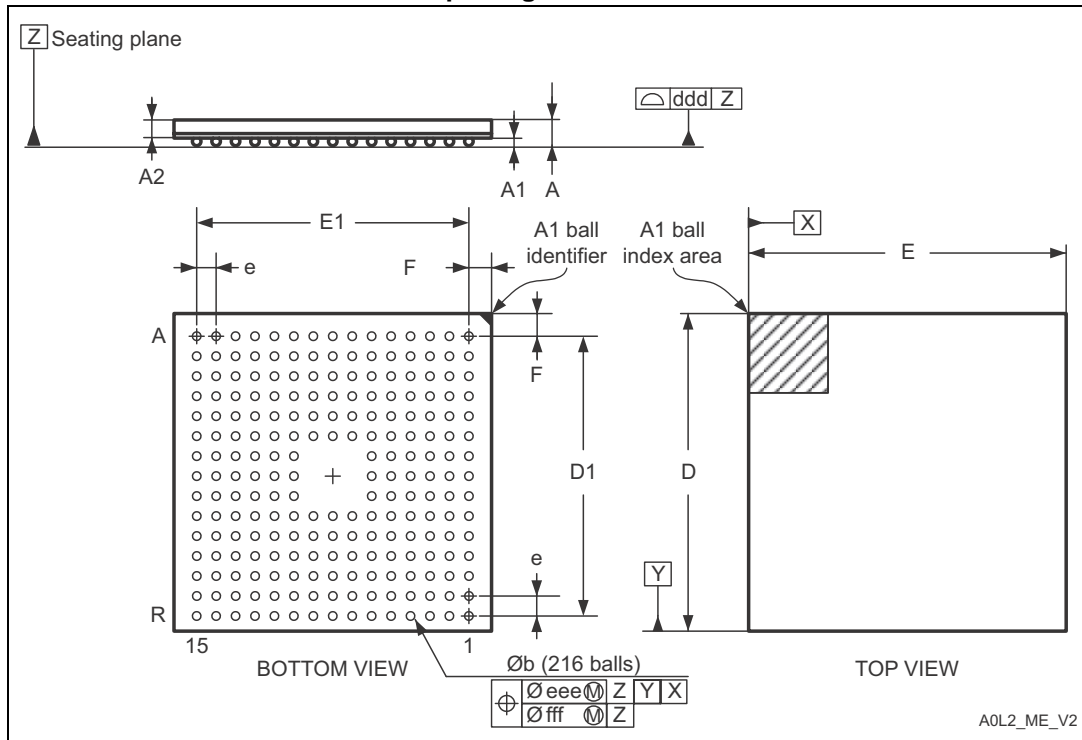
1. Drawing is not to scale.

Table 118. UFBGA176+25 - ultra thin fine pitch ball grid array 10 × 10 × 0.6 mm mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
A	0.460	0.530	0.600	0.0181	0.0209	0.0236
A1	0.050	0.080	0.110	0.002	0.0031	0.0043
A2	0.400	0.450	0.500	0.0157	0.0177	0.0197
b	0.230	0.280	0.330	0.0091	0.0110	0.0130
D	9.950	10.000	10.050	0.3917	0.3937	0.3957
E	9.950	10.000	10.050	0.3917	0.3937	0.3957
e	-	0.650	-	-	0.0256	-
F	0.400	0.450	0.500	0.0157	0.0177	0.0197
ddd	-	-	0.080	-	-	0.0031
eee	-	-	0.150	-	-	0.0059
fff	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

Figure 92. TFBGA216 - thin fine pitch ball grid array 13 × 13 × 0.8mm, package outline



1. Drawing is not to scale.

Table 119. TFBGA216 - thin fine pitch ball grid array 13 × 13 × 0.8mm package mechanical data

Symbol	millimeters			inches <sup>(1)</sup>		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.100	-	-	0.0433
A1	0.150	-	-	0.0059	-	-
A2	-	0.760	-	-	0.0299	-
A4	-	0.210	-	-	0.0083	-
b	0.350	0.400	0.450	0.0138	0.0157	0.0177
D	12.850	13.000	13.150	0.5118	0.5118	0.5177
D1	-	11.200	-	-	0.4409	-
E	12.850	13.000	13.150	0.5118	0.5118	0.5177
E1	-	11.200	-	-	0.4409	-
e	-	0.800	-	-	0.0315	-
F	-	0.900	-	-	0.0354	-
ddd	-	-	0.080	-	-	0.0031

1. Values in inches are converted from mm and rounded to 4 decimal digits.

## 7.2 Thermal characteristics

The maximum chip-junction temperature,  $T_J \text{ max}$ , in degrees Celsius, may be calculated using the following equation:

$$T_J \text{ max} = T_A \text{ max} + (P_D \text{ max} \times \Theta_{JA})$$

Where:

- $T_A \text{ max}$  is the maximum ambient temperature in °C,
- $\Theta_{JA}$  is the package junction-to-ambient thermal resistance, in °C/W,
- $P_D \text{ max}$  is the sum of  $P_{INT} \text{ max}$  and  $P_{I/O} \text{ max}$  ( $P_D \text{ max} = P_{INT} \text{ max} + P_{I/O} \text{ max}$ ),
- $P_{INT} \text{ max}$  is the product of  $I_{DD}$  and  $V_{DD}$ , expressed in Watts. This is the maximum chip internal power.

$P_{I/O} \text{ max}$  represents the maximum power dissipation on output pins where:

$$P_{I/O} \text{ max} = \Sigma (V_{OL} \times I_{OL}) + \Sigma ((V_{DD} - V_{OH}) \times I_{OH}),$$

taking into account the actual  $V_{OL} / I_{OL}$  and  $V_{OH} / I_{OH}$  of the I/Os at low and high level in the application.

**Table 120. Package thermal characteristics**

Symbol	Parameter	Value	Unit
$\Theta_{JA}$	Thermal resistance junction-ambient LQFP100 - 14 × 14 mm / 0.5 mm pitch	43	°C/W
	Thermal resistance junction-ambient WLCSP143	31.2	
	Thermal resistance junction-ambient LQFP144 - 20 × 20 mm / 0.5 mm pitch	40	
	Thermal resistance junction-ambient LQFP176 - 24 × 24 mm / 0.5 mm pitch	38	
	Thermal resistance junction-ambient LQFP208 - 28 × 28 mm / 0.5 mm pitch	19	
	Thermal resistance junction-ambient UFBGA169 - 7 × 7mm / 0.5 mm pitch	52	
	Thermal resistance junction-ambient UFBGA176 - 10× 10 mm / 0.5 mm pitch	39	
	Thermal resistance junction-ambient TFBGA216 - 13 × 13 mm / 0.8 mm pitch	29	

### Reference document

JESD51-2 Integrated Circuits Thermal Test Method Environment Conditions - Natural Convection (Still Air). Available from [www.jedec.org](http://www.jedec.org).

# 8 Part numbering

**Table 121. Ordering information scheme**

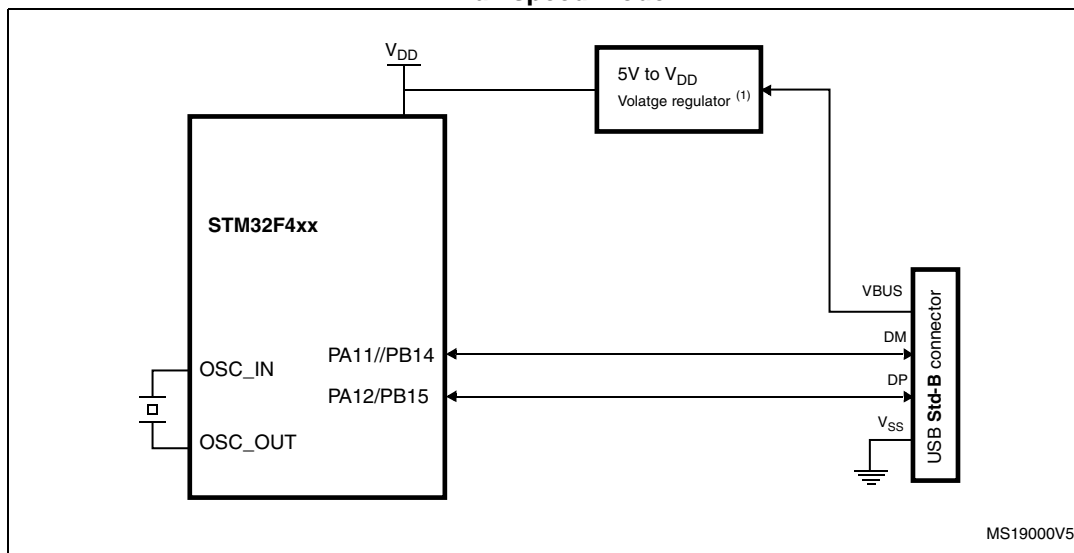
Example:	STM32	F	439	V	I	T	6	xxx
<b>Device family</b>								
STM32 = ARM-based 32-bit microcontroller								
<b>Product type</b>								
F = general-purpose								
<b>Device subfamily</b>								
437= STM32F437xx, USB OTG FS/HS, camera interface, Ethernet, cryptographic acceleration								
439= STM32F439xx, USB OTG FS/HS, camera interface, Ethernet, LCD-TFT, cryptographic acceleration								
<b>Pin count</b>								
V = 100 pins								
Z = 144 pins								
A = 169 pins								
I = 176 pins								
B = 208 pins								
N = 216 pins								
<b>Flash memory size</b>								
G = 1024 Kbytes of Flash memory								
I = 2048 Kbytes of Flash memory								
<b>Package</b>								
T = LQFP								
H = BGA								
Y = WLCSP								
<b>Temperature range</b>								
6 = Industrial temperature range, -40 to 85 °C.								
7 = Industrial temperature range, -40 to 105 °C.								
<b>Options</b>								
xxx = programmed parts								
TR = tape and reel								

For a list of available options (speed, package, etc.) or for further information on any aspect of this device, please contact your nearest ST sales office.

## Appendix A Application block diagrams

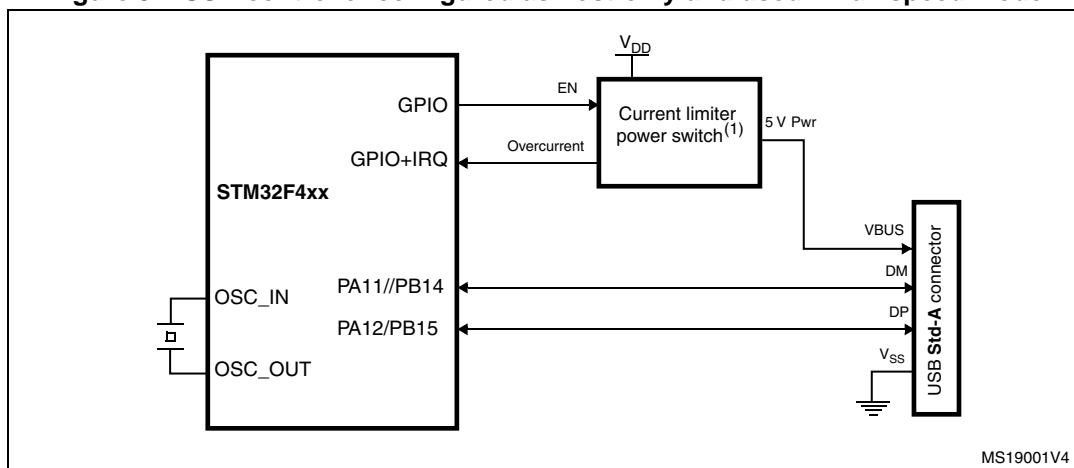
### A.1 USB OTG full speed (FS) interface solutions

Figure 93. USB controller configured as peripheral-only and used in Full speed mode



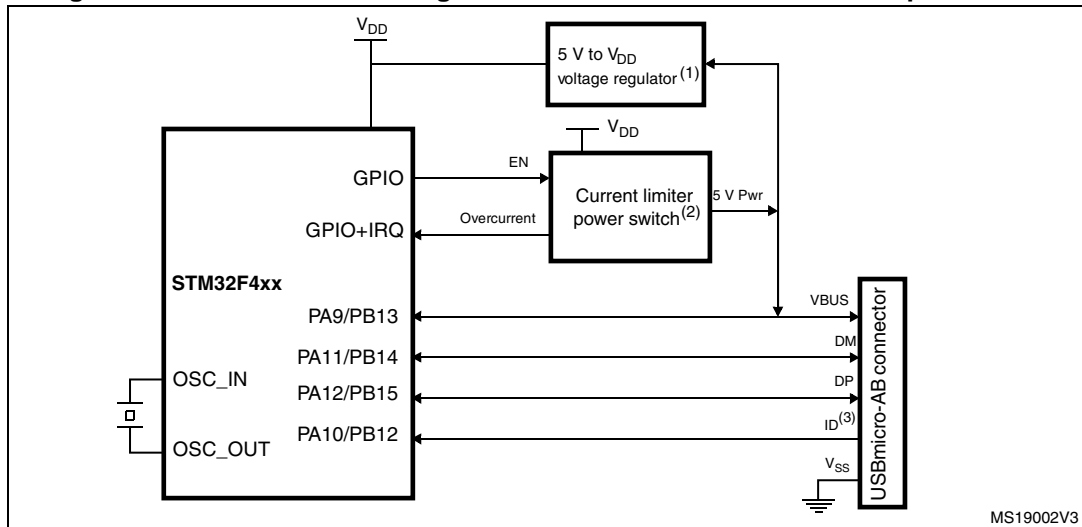
1. External voltage regulator only needed when building a V<sub>BUS</sub> powered device.
2. The same application can be developed using the OTG HS in FS mode to achieve enhanced performance thanks to the large Rx/Tx FIFO and to a dedicated DMA controller.

Figure 94. USB controller configured as host-only and used in full speed mode



1. The current limiter is required only if the application has to support a V<sub>BUS</sub> powered device. A basic power switch can be used if 5 V are available on the application board.
2. The same application can be developed using the OTG HS in FS mode to achieve enhanced performance thanks to the large Rx/Tx FIFO and to a dedicated DMA controller.

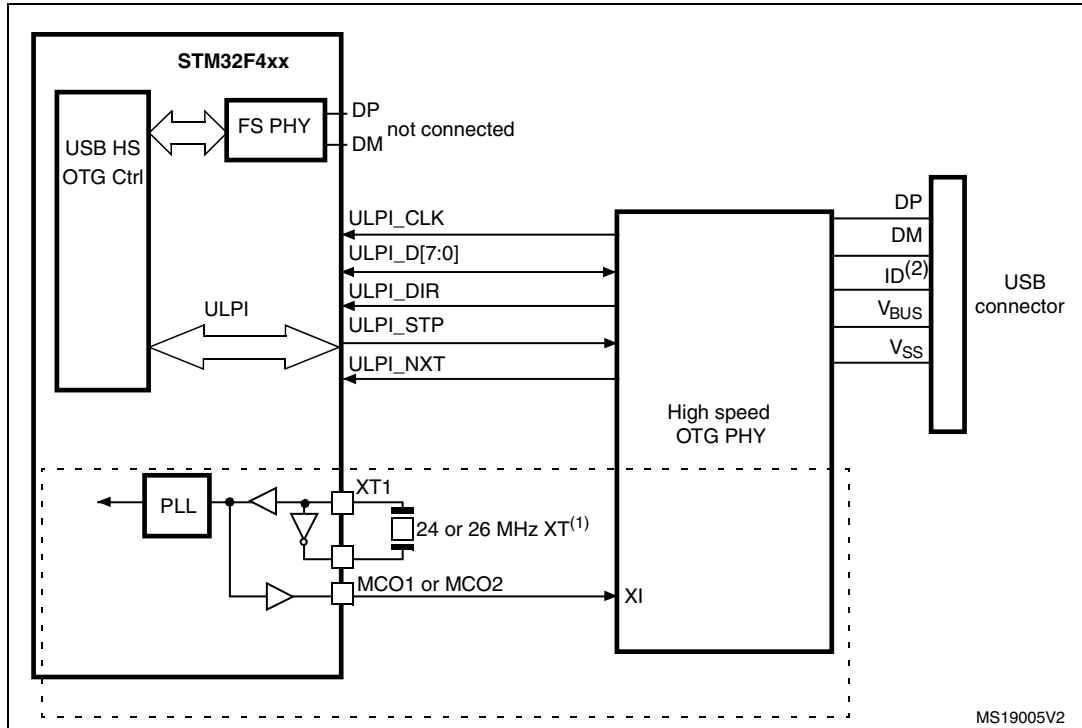
Figure 95. USB controller configured in dual mode and used in full speed mode



1. External voltage regulator only needed when building a V<sub>BUS</sub> powered device.
2. The current limiter is required only if the application has to support a V<sub>BUS</sub> powered device. A basic power switch can be used if 5 V are available on the application board.
3. The ID pin is required in dual role only.
4. The same application can be developed using the OTG HS in FS mode to achieve enhanced performance thanks to the large Rx/Tx FIFO and to a dedicated DMA controller.

## A.2 USB OTG high speed (HS) interface solutions

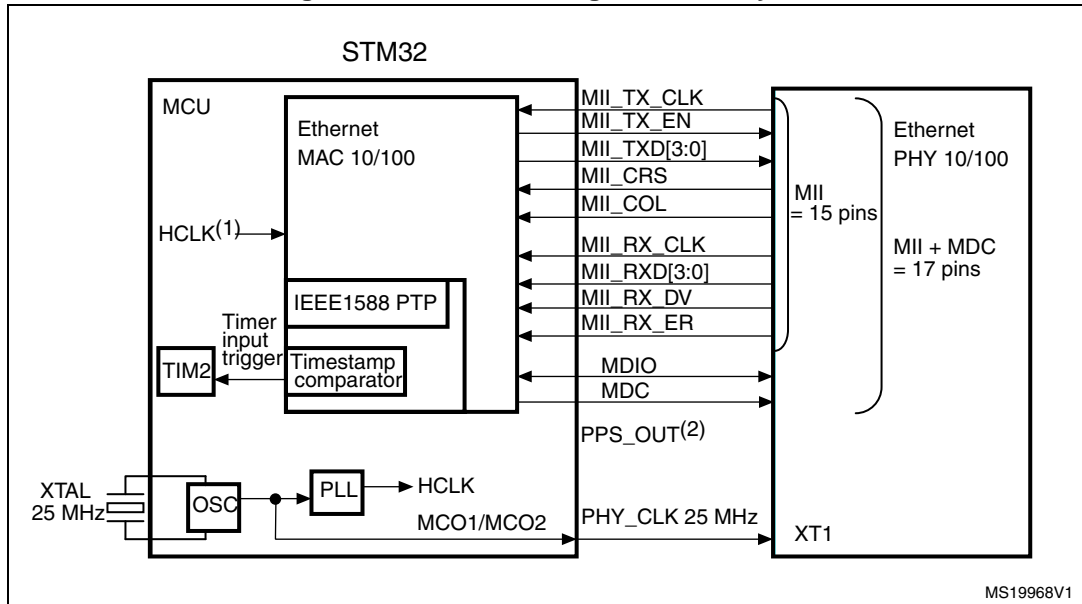
Figure 96. USB controller configured as peripheral, host, or dual-mode and used in high speed mode



1. It is possible to use MCO1 or MCO2 to save a crystal. It is however not mandatory to clock the STM32F43x with a 24 or 26 MHz crystal when using USB HS. The above figure only shows an example of a possible connection.
2. The ID pin is required in dual role only.

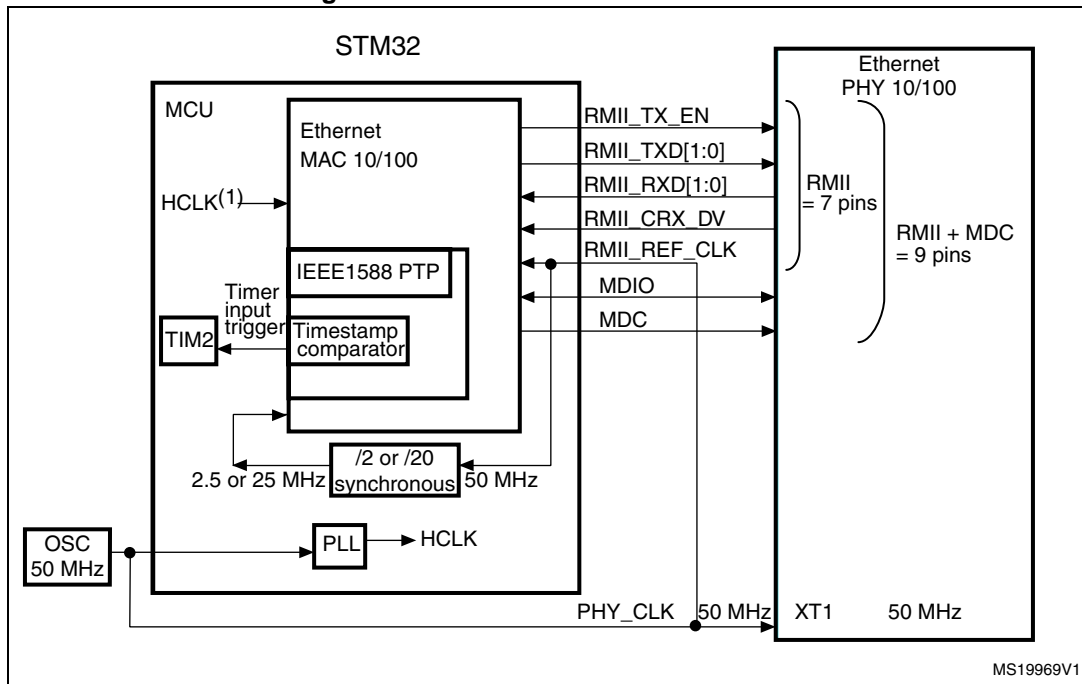
### A.3 Ethernet interface solutions

Figure 97. MII mode using a 25 MHz crystal



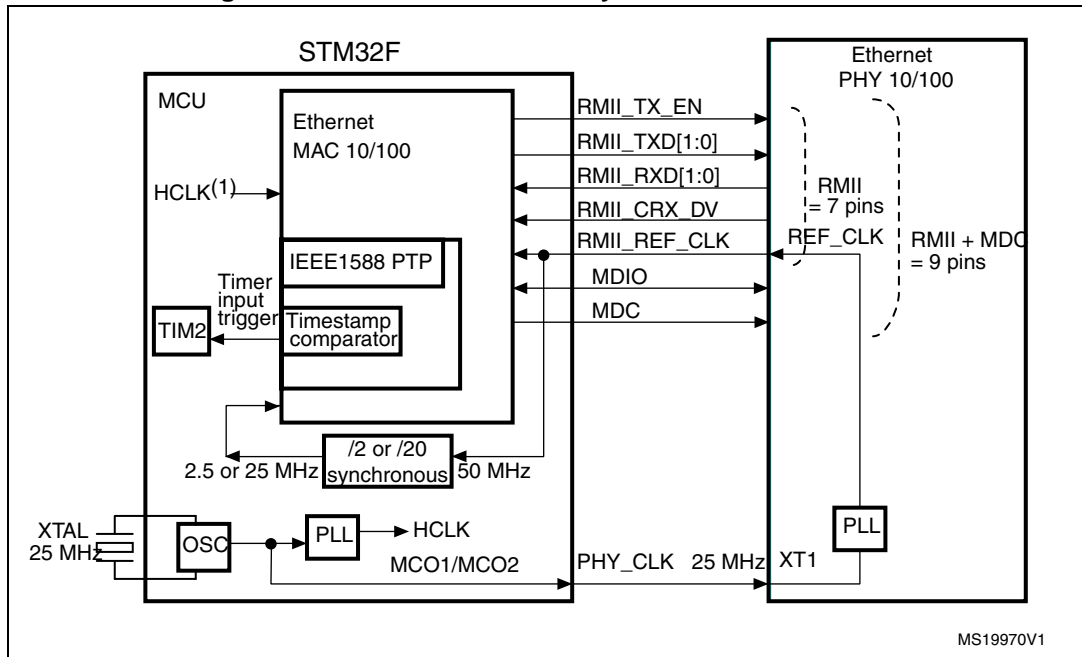
1.  $f_{HCLK}$  must be greater than 25 MHz.
2. Pulse per second when using IEEE1588 PTP optional signal.

Figure 98. RMIi with a 50 MHz oscillator



1.  $f_{HCLK}$  must be greater than 25 MHz.

Figure 99. RMIi with a 25 MHz crystal and PHY with PLL



1.  $f_{HCLK}$  must be greater than 25 MHz.
2. The 25 MHz (PHY\_CLK) must be derived directly from the HSE oscillator, before the PLL block.

## 9 Revision history

**Table 122. Document revision history**

Date	Revision	Changes
12-Aug-2013	1	Initial release.
10-Sep-2013	2	<p>Added STM32F439xx part numbers and related informations.  <b>STM32F437xx part numbers:</b>            Replaced FSMC by FMC added Chrom-ART Accelerator and SAI interface.            Increased core, timer, GPIOs, SPI maximum frequencies            Updated <a href="#">Figure 4: STM32F437xx and STM32F439xx block diagram</a>.            Updated <a href="#">Figure 5: STM32F437xx and STM32F439xx Multi-AHB matrix</a>.            Removed note in <a href="#">Section : Standby mode</a>.            Updated <a href="#">Figure 14: STM32F43x LQFP176 pinout</a>.            Updated <a href="#">Table 10: STM32F437xx and STM32F439xx pin and ball definitions</a> and <a href="#">Table 12: STM32F437xx and STM32F439xx alternate function mapping</a>..            Modified <a href="#">Figure 19: Memory map</a>.            Updated <a href="#">Table 17: General operating conditions</a>, <a href="#">Table 18: Limitations depending on the operating power supply range</a>. Removed note 1 in <a href="#">Table 22: reset and power control block characteristics</a>. Added <a href="#">Table 23: Over-drive switching characteristics</a>.            Updated <a href="#">Section : Typical and maximum current consumption</a>, <a href="#">Table 34: Switching output I/O current consumption</a>, <a href="#">Table 35: Peripheral current consumption</a> and <a href="#">Section : On-chip peripheral current consumption</a>.            Updated <a href="#">Table 36: Low-power mode wakeup timings</a>.            Modified <a href="#">Section : High-speed external user clock generated from an external source</a>, <a href="#">Section : Low-speed external user clock generated from an external source</a>, and <a href="#">Section 6.3.10: Internal clock source characteristics</a>.            Updated <a href="#">Table 43: Main PLL characteristics</a> and <a href="#">Table 45: PLLISAI (audio and LCD-TFT PLL) characteristics</a>.            Updated <a href="#">Table 52: EMI characteristics</a>.            Updated <a href="#">Table 57: Output voltage characteristics</a> and <a href="#">Table 58: I/O AC characteristics</a>.            Updated <a href="#">Table 60: TIMx characteristics</a>, <a href="#">Table 61: <math>I^2C</math> characteristics</a>, <a href="#">Table 63: SPI dynamic characteristics</a>, <a href="#">Section : SAI characteristics</a>.            Updated <a href="#">Table 104: SDRAM read timings</a> and <a href="#">Table 106: SDRAM write timings</a>.</p>

Table 122. Document revision history (continued)

Date	Revision	Changes
24-Jan-2014	3	<p>Added STM32F437AI and STM32F439AI part numbers and UFBGA169 package.</p> <p>Changed INTN into INTR in <i>Figure 4: STM32F437xx and STM32F439xx block diagram</i>.</p> <p>Updated <i>Section 3.15: Boot modes</i>.</p> <p>Updated for PA4 and PA5 in <i>Table 10: STM32F437xx and STM32F439xx pin and ball definitions</i>.</p> <p>Added V<sub>IN</sub> for BOOT0 pins in <i>Table 14: Voltage characteristics</i>.</p> <p>Updated Note 6. added Note 1., and updated maximum V<sub>IN</sub> for B pins in <i>Table 17: General operating conditions</i>.</p> <p>Updated maximum Flash memory access frequency with wait states for V<sub>DD</sub> =1.8 to 2.1 V in <i>Table 18: Limitations depending on the operating power supply range</i>.</p> <p>Updated <i>Table 24: Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator enabled except prefetch) or RAM</i> and <i>Table 25: Typical and maximum current consumption in Run mode, code with data processing running from Flash memory (ART accelerator disabled)</i>.</p> <p>Updated <i>Table 30: Typical current consumption in Run mode, code with data processing running from Flash memory or RAM, regulator ON (ART accelerator enabled except prefetch), VDD=1.8 V, Table 31: Typical current consumption in Run mode, code with data processing running from Flash memory, regulator OFF (ART accelerator enabled except prefetch), and Table 32: Typical current consumption in Sleep mode, regulator ON, VDD=1.8 V</i>.</p> <p>Updated <i>Table 57: Output voltage characteristics</i>.</p> <p>Updated <i>Table 58: I/O AC characteristics</i>. Added <i>Figure 35</i>.</p> <p>Updated t<sub>h(SDA)</sub>, t<sub>r(SDA)</sub> and t<sub>r(SCL)</sub> and added t<sub>SP</sub> in <i>Table 61: I<sup>2</sup>C characteristics</i>.</p> <p>Updated f<sub>SCK</sub> in <i>Table 63: SPI dynamic characteristics</i>.</p> <p>Updated <i>Table 71: Dynamic characteristics: USB ULPI</i>.</p> <p>Updated <i>Section 6.3.26: FMC characteristics</i> conditions. Updated <i>Figure 74: SDRAM read access waveforms (CL = 1)</i> and <i>Figure 75: SDRAM write access waveforms</i>. Added <i>Table 105: LPSDR SDRAM read timings</i> and <i>Table 107: LPSDR SDRAM write timings</i>. Updated <i>Table 104: SDRAM read timings</i> and <i>Table 106: SDRAM write timings</i> and added note 2. <i>Table 110: Dynamic characteristics: SD / MMC characteristics</i>.</p>
31-Jan-2014	4	<p>In the whole document, minimum supply voltage changed to 1.7 V when external power supply supervisor is used.</p> <p>Updated conditions in <i>Table 63: SPI dynamic characteristics</i>.</p> <p>Added Z<sub>DRV</sub> in <i>Table 68: USB OTG full speed electrical characteristics</i></p>

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