

MAXIM

270Mbps SFP LED Driver

MAX3967A

General Description

The MAX3967A is a programmable LED driver for fiber optic transmitters operating at data rates up to 270Mbps. The circuit contains a high-speed current driver with programmable temperature coefficient (tempco), adjustments for LED prebias voltage, and a disable feature. The circuit accepts PECL data inputs, and operates from a single +2.97V to +5.5V power supply.

The SFP LED driver can switch up to 100mA into typical high-speed light-emitting diodes. As temperature increases, the device's modulation current increases with a tempco that is programmable from 2500ppm/°C to 12,000ppm/°C. The modulation current is programmed with a single external resistor.

The MAX3967A's LED prebias voltage is programmable from 400mV to 925mV. The prebias circuit produces peaking current, which improves the LED switching speed.

Complementary current outputs help to maintain a constant supply current, reducing EMI and supply noise generated by the transmitter module. The MAX3967A is available in die form, or in a 4mm x 4mm, 24-pin thin QFN package.

Applications

Multimode LED Transmitters
Fast Ethernet/FDDI
155Mbps LAN ATM Transceivers
ESCON Receivers
SFP Transceivers

Typical Operating Circuits appear at end of data sheet.

Features

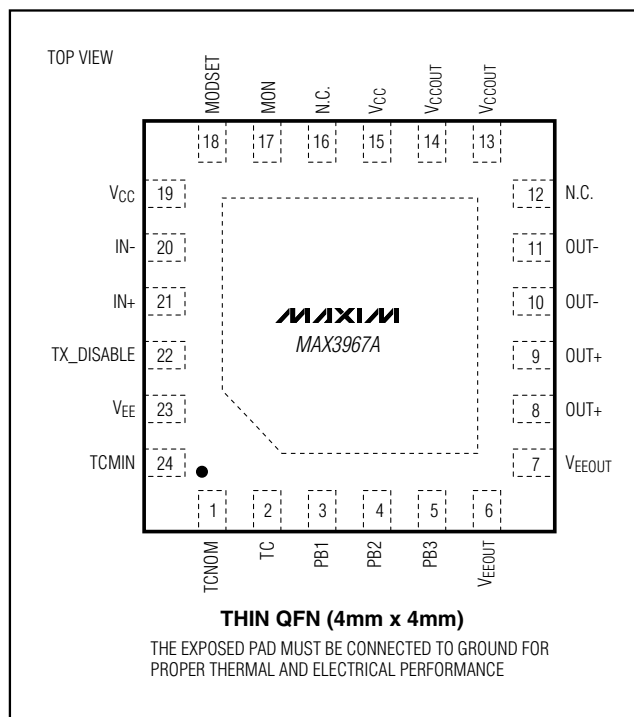
- ◆ TX_DISABLE for SFP Compatibility
- ◆ Single +2.97V to +5.5V Power Supply
- ◆ Adjustable Temperature Compensation
- ◆ Adjustable Modulation Current
- ◆ Complementary Output Reduces Supply Noise
- ◆ Programmable LED Prebias Voltage
- ◆ Available in 24-Pin Thin QFN or Die

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX3967AETG	-40°C to +85°C	24 Thin QFN
MAX3967AE/D	-40°C to +85°C	Dice*

*Dice are tested and guaranteed only at $T_A = +25^\circ\text{C}$.

Pin Configuration



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ABSOLUTE MAXIMUM RATINGS

Supply Voltage at V_{CC} , V_{CCOUT}

(V_{EE} , $V_{EEOUT} = 0V$)-0.5V to +7V

Current into $OUT+$, $OUT-$ -40mA to +160mA

Differential Output Voltage ($OUT+$ to $OUT-$)-3.3V to +3.3V

Voltage at $PB1$, $PB2$, $PB3$,

$IN+$, $IN-$, $OUT+$, $OUT-$, $TX_DISABLE$-0.5V to ($V_{CC} + 0.5V$)

Voltage at $TCMIN$, $TCNOM$, TC , $MODSET$, MON -0.5V to +2V

Continuous Power Dissipation ($T_A = +85^\circ C$)

24-Lead Thin QFN (derate 20.8mW/ $^\circ C$

above $+85^\circ C$).....1354mW

Operating Junction Temperature Range-40 $^\circ C$ to +150 $^\circ C$

Die Attach Temperature+375 $^\circ C$

Storage Temperature Range-50 $^\circ C$ to +150 $^\circ C$

Lead Temperature (soldering, 10s)+300 $^\circ C$

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

(Load as specified in Figure 1; $V_{CC} = +2.97V$ to $+5.5V$ (at the V_{CC} pins); V_{EE} , $V_{EEOUT} = 0V$; $T_A = -40^\circ C$ to $+85^\circ C$, unless otherwise noted. Temperature coefficients are referenced to $T_A = +25^\circ C$. Typical values are at $V_{CC} = +3.3V$, $T_A = +25^\circ C$, unless otherwise noted. Dice are tested at $T_A = +25^\circ C$ only.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Data Input High Voltage		Referenced to V_{CC} , DC-coupled input	-1.165		-0.880	V
Data Input Low Voltage		Referenced to V_{CC} , DC-coupled input	-1.810		-1.475	V
Supply Current	I_{CC}	(Note 1)		30	39	mA
Input Current at $IN+$ or $IN-$			-50		+50	μA
Modulation Current		$R_{MODSET} = 698\Omega$ (Note 2)	$T_A = -40^\circ C$		109	mA
			$T_A = +25^\circ C$		112 126 140	
			$T_A = +85^\circ C$		126 140 155	
		$R_{MODSET} = 3.0k\Omega$ (Note 2)	$T_A = -40^\circ C$		15.3	
			$T_A = +25^\circ C$		17.5 19.8 25	
			$T_A = +85^\circ C$		21.9 24 28.7	
		$R_{MODSET} = 1.0k\Omega$ (Note 2)	$T_A = -40^\circ C$		65 74 89	
			$T_A = +25^\circ C$ $V_{CC} = 3.3V$		79 86 89	
			$T_A = +85^\circ C$		90 96 110	
Prebias Voltage		$PB1$, $PB2$, $PB3 = (open, open, open)$	0.368	0.400	0.451	V
		$PB1$, $PB2$, $PB3 = (V_{EE}, V_{EE}, open)$	0.575	0.625	0.696	
		$PB1$, $PB2$, $PB3 = (V_{EE}, V_{EE}, V_{EE})$	0.848	0.925	1.026	
Temperature Coefficient of Modulation Current		Maximum tempco (TC open)		12,000		ppm/ $^\circ C$
		Nominal tempco (TC shorted to $TCNOM$)		3600		
		Minimum tempco (TC shorted to $TCMIN$)		2500		
Prebias Resistor	$R_{PREBIAS}$		66	78	90	Ω
$TX_DISABLE$ Resistance		Resistance to V_{EE} (Note 3)	50	65	100	$k\Omega$
$TX_DISABLE$ High	V_{IH}		2.0			V
$TX_DISABLE$ Low	V_{IL}				0.8	V
Monitor Gain		I_{MON} / I_{MODSET} , $V_{MON} < 1.1V$, $R_{MODSET} = 1k\Omega$, $TC = TCMIN$	0.92	1	1.08	A/A

Note 1: $R_{MODSET} = 1k\Omega$. Excludes I_{OUT+} and I_{OUT-} , $TX_DISABLE$ high or low.

Note 2: TC connected to $TCMIN$, $PB1 = PB2 = V_{EE}$, $PB3 = open$.

Note 3: The $TX_DISABLE$ pin is internally pulled low. The driver is enabled when $TX_DISABLE$ is left open.

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AC ELECTRICAL CHARACTERISTICS

(Load as specified in Figure 1, unless otherwise noted. $V_{CC} = +2.97V$ to $+5.5V$ (at the V_{CC} pins), $R_{MODSET} = 1k\Omega$, $T_A = -40^{\circ}C$ to $+85^{\circ}C$. Input data shaped by 470MHz 4-pole filter, $PB1 = PB2 = V_{EE}$, $PB3 = open$. Typical values are at $V_{CC} = +3.3V$, TC connected to TCMIN, $T_A = +25^{\circ}C$.) (Note 4)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Data Input Range		Differential input	500		2400	mV _{P-P}
Output-Current Edge Speed		20% to 80%, input is a 12.5MHz square wave	300	615	1230	ps
Output-Current Pulse-Width Correction (PWC)		(Note 5)		-90		ps
Output-Current Data-Dependent Jitter	DJ	266Mbps (Note 6)		140		ps _{P-P}
		155Mbps (Note 7)		150	250	
Random Jitter	RJ			3		ps _{RMS}
TX_DISABLE Assert Time	t_{off}	Time from rising edge of TX_DISABLE to output at 10% of steady state		0.01	0.5	μs
TX_DISABLE Negate Time	t_{on}	Time from rising edge of TX_DISABLE to output at 90% of steady state		0.01	0.5	μs
Power-On Time	t_{init}	Time from $V_{CC} > 2.97V$ to output at 90% of steady state		0.1	2	ms

Note 4: AC characteristics are guaranteed by design and characterization.

Note 5: $PWC = (width_{CURRENT\ ON} - width_{CURRENT\ OFF}) / 2$.

Note 6: Test pattern is a K28.5 (0011 1110 1011 0000 0101) transmitted at 266Mbps.

Note 7: Test pattern is equivalent to a $2^{13} - 1$ PRBS containing 72 consecutive zeros or 72 consecutive ones.

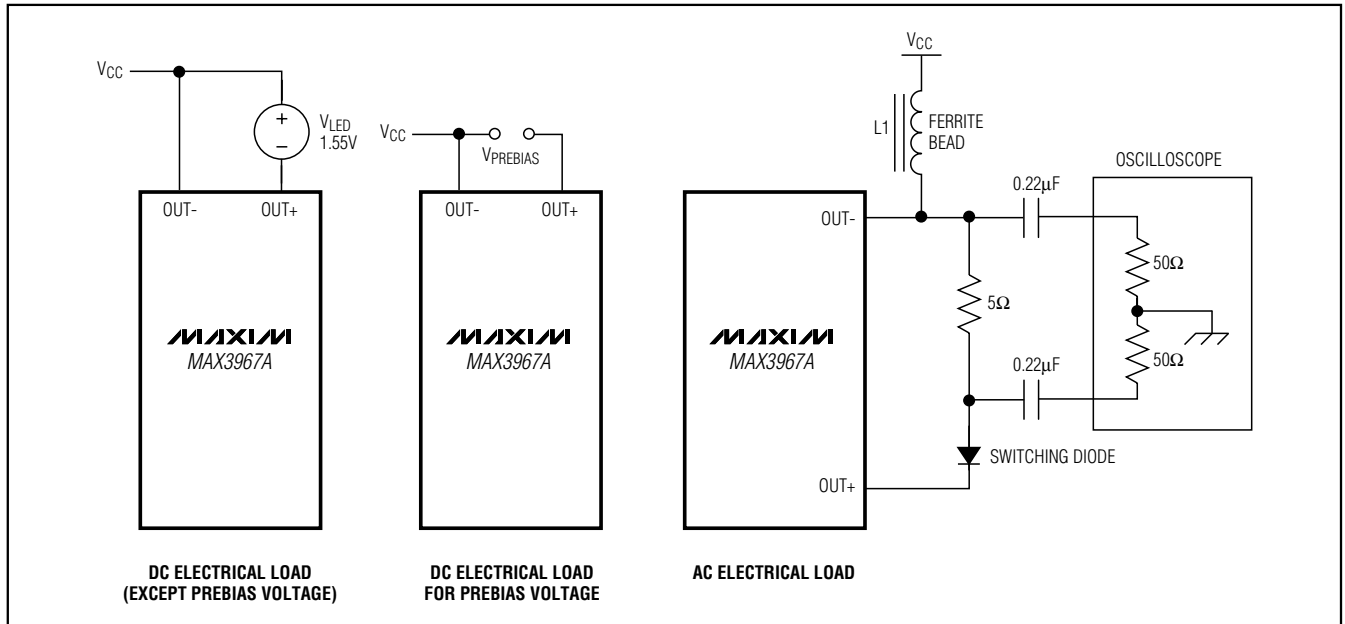
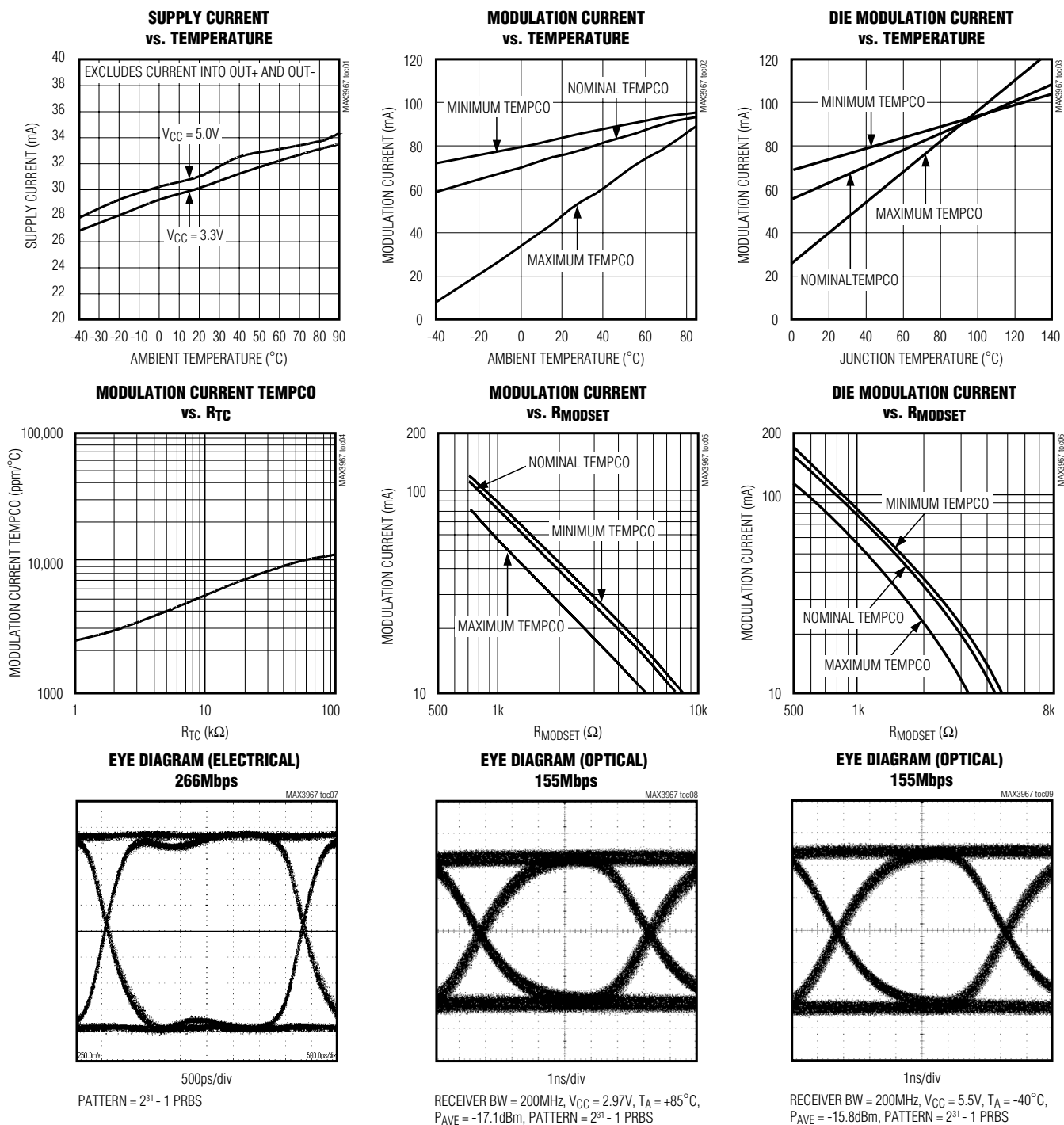


Figure 1. MAX3967A Output Test Loads

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Typical Operating Characteristics

(MAX3967AETG in Maxim evaluation board, $V_{CC} = +3.3V$, PB1 = PB2 = V_{EE} , PB3 = open, TC connected to TCNOM, $R_{MODSET} = 1k\Omega$, $T_A = +25^\circ C$, unless otherwise noted.)

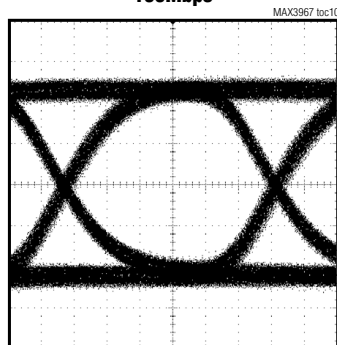


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Typical Operating Characteristics (continued)

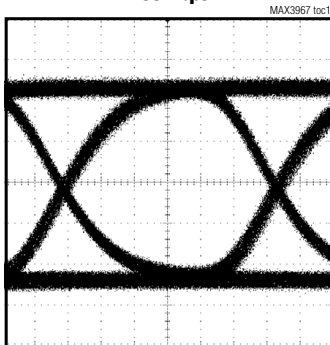
(MAX3967AETG in Maxim evaluation board, $V_{CC} = +3.3V$, PB1 = PB2 = V_{EE} , PB3 = open, TC connected to TCNOM, $R_{MODSET} = 1k\Omega$, $T_A = +25^\circ C$, unless otherwise noted.)

EYE DIAGRAM (OPTICAL)
155Mbps



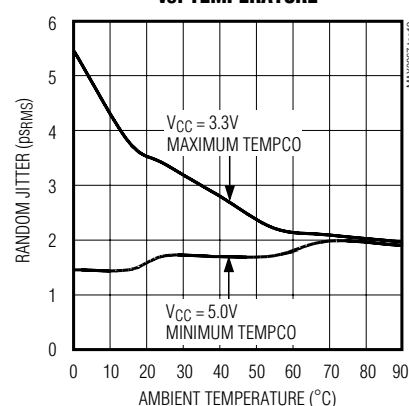
1ns/div
RECEIVER BW = 200MHz, $V_{CC} = 5.5V$, $T_A = +85^\circ C$,
 $P_{AVE} = -17.1dBm$, PATTERN = $2^{31} - 1$ PRBS

EYE DIAGRAM (OPTICAL)
155Mbps

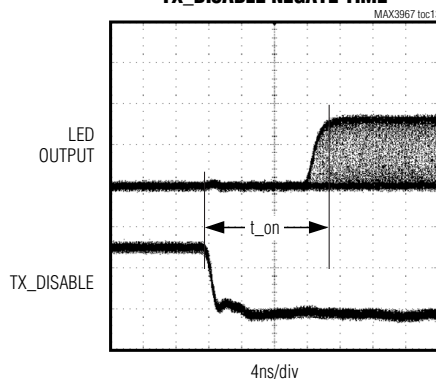


1ns/div
RECEIVER BW = 200MHz, $V_{CC} = 2.97V$, $T_A = -40^\circ C$,
 $P_{AVE} = -15.8dBm$, PATTERN = $2^{31} - 1$ PRBS

RANDOM JITTER
vs. TEMPERATURE

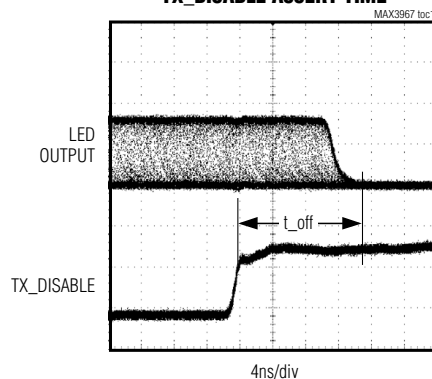


TX_DISABLE NEGATE TIME



4ns/div

TX_DISABLE ASSERT TIME



4ns/div

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Pin Description

PIN	NAME	FUNCTION
1	TCNOM	Shorting TC to TCNOM provides a modulation tempco of approximately 3600ppm/°C.
2	TC	A resistor (R_{TC}) connected between the TC and TCMIN pins sets the tempco of the modulation current. Leaving R_{TC} unconnected provides the maximum tempco.
3, 4, 5	PB1, PB2, PB3	Programs the Prebias Voltage at the OUT+ Pin (Table 1)
6, 7	VEEOUT	Ground for the Output-Current Drivers
8, 9	OUT+	Current Output
10, 11	OUT-	Complementary Current Output
12, 16	N.C.	No Connection. Not internally connected.
13, 14	VCCOUT	Supply Connection for the Output-Current Drivers
15, 19	VCC	Provides Current to the Internal Amplifiers
17	MON	The Current Sourced from the MON Pin is Proportional to the Modulator Current
18	MODSET	A Resistor from MODSET to VEE Programs the LED Modulation Current
20	IN-	Inverting Data Input
21	IN+	Noninverting Data Input
22	TX_DISABLE	Transmit Disable. When high, the current at the OUT+ pins is in the low state. The transmitter is enabled when TX_DISABLE is open.
23	VEE	Ground for internal amplifiers.
24	TCMIN	Shorting TC to TCMIN provides the minimum modulation-current tempco.
EP	Exposed Pad	Connect the exposed pad to board ground for optimal correct electrical and thermal performance.

Detailed Description

The MAX3967A provides a flexible current drive for the modulation of fiber optic light-emitting diodes (LEDs). The circuit is designed to be used with +3.3V or +5V power supplies. The IC provides up to 100mA of modulation current. An adjustable prebias current source sets the LED prebias voltage. An integrated resistor provides passive peaking and optical pulse-width compensation.

Figure 2 shows a block diagram of the MAX3967A, which comprises a reference-voltage generator, modulation-current generator, input buffer with disable, prebias-current generator, main output driver, complementary output driver, and LED-compensation network.

Temperature Compensation

The reference-voltage generator circuit provides two voltage sources that create modulation-current temperature compensation. A positive modulation-current tem-

perature coefficient (tempco) is useful to compensate for the temperature characteristics of typical fiber optic LEDs. The first source has a temperature-stable output. The second source has a temperature-increasing output with a tempco of approximately 12,000ppm/°C (relative to +25°C). A resistor-divider between the two reference generators programs the modulation-current tempco. For maximum modulation-current tempco, leave the TC pin disconnected. For a tempco of approximately 3600ppm/°C, connect TC to TCNOM. To obtain the minimum tempco, connect TCMIN to TC. Intermediate tempco values can be programmed by connecting an external resistor (R_{TC}) between TCMIN and TC.

Input Buffer

The inputs are connected to the PECL-compatible differential input buffer. If left unconnected, IN+ is internally pulled to a PECL low and IN- is pulled to a PECL high, causing low current at OUT+. The input impedance of IN+ and IN- is approximately 50k Ω .

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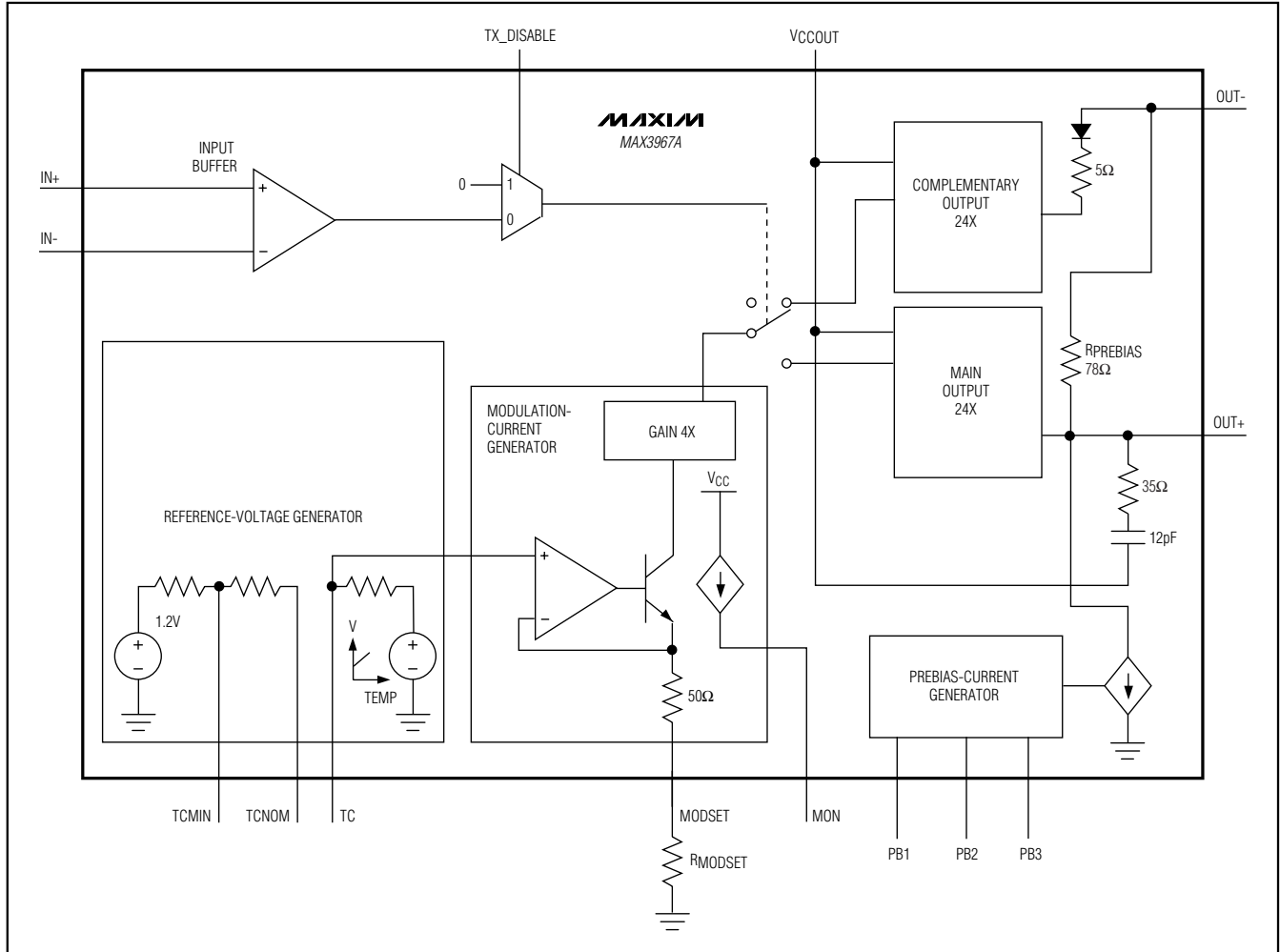


Figure 2. Functional Diagram

Modulation-Current Generator

The modulation-current generator circuit provides control of the modulation-current amplitude. This amplitude is determined by the voltage at the MODSET pin and external resistor R_{MODSET} .

Do not connect bypass capacitors at the MODSET pin. Capacitance at this pin increases high-frequency output noise. The MON pin provides an optional modulation-current monitor. The current sourced from the MON pin is 1/96 of the modulation current. If used, the pin should be connected to V_{EE} through a resistor. The resistance must be chosen so the voltage on MON does not exceed 1.1V. If not used, leave MON open.

Prebias Current Generator

A prebias voltage ($V_{PREBIAS}$) can be applied to the LED to improve switching speed. The prebias current generator creates a current that flows through the 78Ω prebias resistor in the output stage, creating a prebias voltage. The prebias voltage can be adjusted by selectively connecting pins PB1, PB2, and PB3 to V_{EE} . Table 1 describes the functions of PB1, PB2, and PB3.

Output Current Drivers

The modulation-current reference is switched and amplified by the output stages.

LED package lead inductance causes ringing and overshoot, which can be compensated with an RC filter network. The MAX3967A includes 35Ω and 12pF of

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Table 1. LED Prebias Voltage

PB1	PB2	PB3	PREBIAS (V)
Open	Open	Open	0.400
VEE	Open	Open	0.475
Open	VEE	Open	0.550
VEE	VEE	Open	0.625
Open	Open	VEE	0.700
VEE	Open	VEE	0.775
Open	VEE	VEE	0.850
VEE	VEE	VEE	0.925

internal compensation. The compensation network can be optimized by adding additional components between VCCOUT and OUT+.

The MAX3967A includes a complementary output driver, which is switched 180° out of phase with the main output. This configuration helps to maintain constant current flow from the voltage supply, reducing noise and EMI. A large diode and a 5Ω resistor are connected in series with the negative output (OUT-) to emulate the LED load at OUT+.

Peaking Current

The prebias resistor provides peaking current to improve the LED switching speed. The peaking magnitude is given by the following equation:

$$I_{PEAK} = \frac{V_{LED} - V_{PREBIAS}}{78\Omega}$$

The peaking amplitude is equal for rising and falling data transitions.

Design Procedure

Select an LED

For best performance, select a high-efficiency, low-inductance LED. LED inductance causes large voltage swings and ringing.

Program the Modulation-Current Tempco

Select a modulation-current tempco that provides nearly constant LED output power as temperature varies. For the minimum tempco, connect TCMIN to the TC pin. For a tempco of approximately 3600ppm/°C, connect TC to TCNOM and leave TCMIN unconnected. For the maximum tempco, leave TCMIN, TCNOM, and TC unconnected.

See the Modulation Current Tempco vs. RTC graph in the *Typical Operating Characteristics* to program a custom tempco. From the graph, determine the appropriate resistor and connect it between TCMIN and TC.

For example, if an LED requires a 5000ppm/°C tempco, choose RTC of 8.3kΩ.

Program the Modulation Current

Determine the required modulation current at TA = +25°C. Then select the appropriate value of RMODSET from the Modulation Current vs. RMODSET graph in the *Typical Operating Characteristics*.

For example, to program 75mA modulation current, the graph indicates an RMODSET value of 750Ω for maximum tempco (12,000ppm/°C) and 1kΩ for nominal tempco (3600ppm/°C). By interpolation, choose an RMODSET of 792Ω for a tempco of 5000ppm/°C.

Program Prebias Voltage

Determine the LED prebias voltage that produces an acceptable trade-off between peaking current and extinction ratio. See Table 1 for PB1, PB2, and PB3 settings.

Layout Considerations

For optimum performance, total load inductance should not exceed 10nH. Load inductance includes LED inductance, LED package lead inductance, and circuit-board traces. Keep the connections between the MAX3967A OUT pins and the LED as short as possible to minimize inductance.

Chip-and-wire (hybrid) technology reduces package inductance significantly, and provides the best possible performance.

Use good high-frequency layout techniques and a multilayer board with an uninterrupted ground plane. Power supplies should be capacitively bypassed to the ground plane with surface-mount capacitors located near the power-supply pins.

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Input Terminations

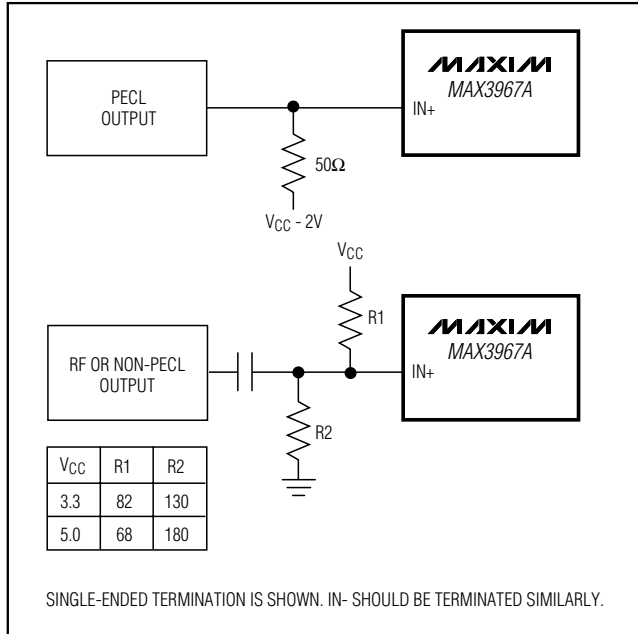


Figure 3. Input Terminations

Exposed-Pad Package

The exposed pad on the 24-pin TQFN provides a very low thermal resistance path for heat removal from the IC.

Chip Information

TRANSISTOR COUNT: 331

SUBSTRATE CONNECTED TO VEE

PROCESS: BIPOLAR

DIE THICKNESS: 15 mils

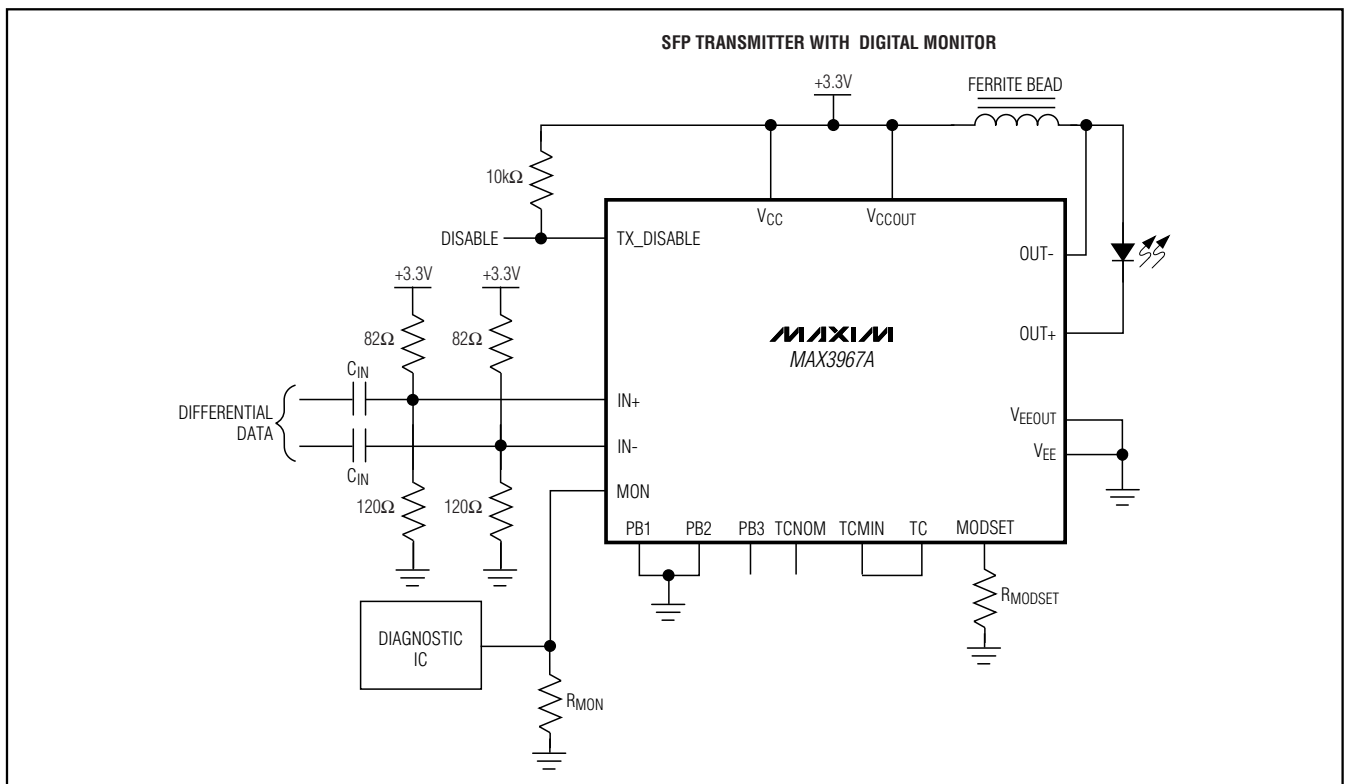
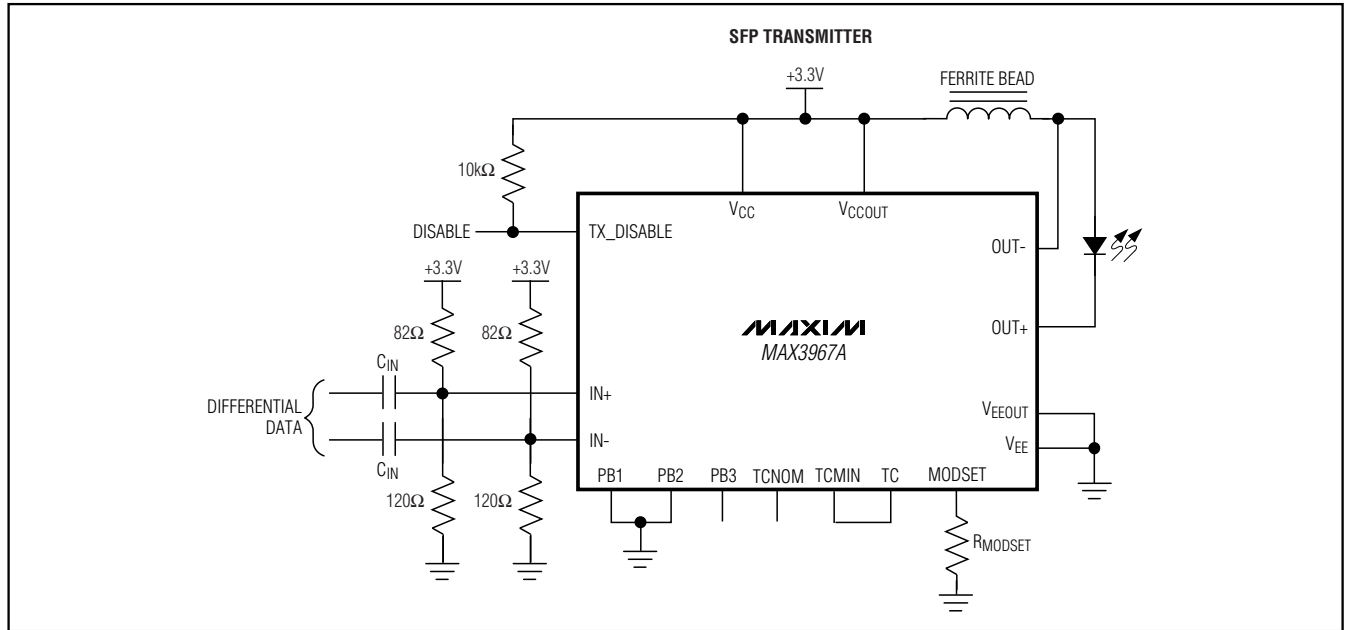
Applications Information

Wire-Bonding Die

The MAX3967A utilizes gold metalization, which provides high reliability. Make connections to the die with gold wire only, using ball-bonding techniques. Use caution if attempting wedge-bonding. Pad size is 4 mils x 4 mils (100μm). Die thickness is typically 15 mils (375μm).

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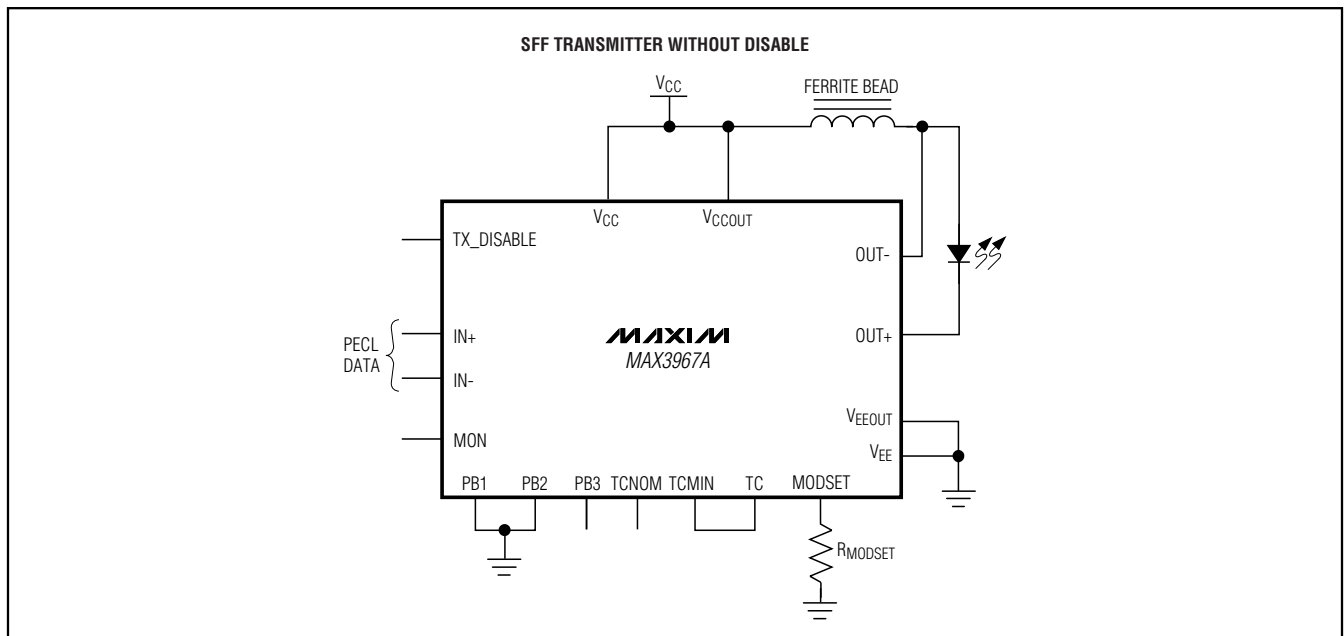
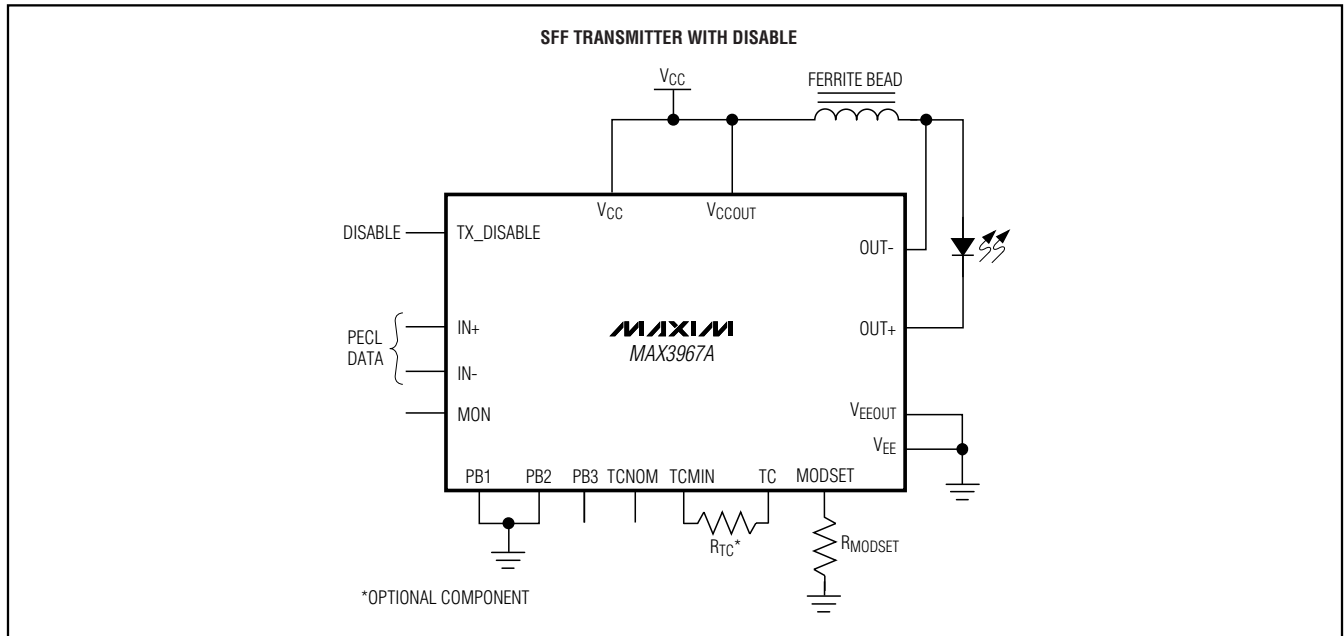
Typical Operating Circuits



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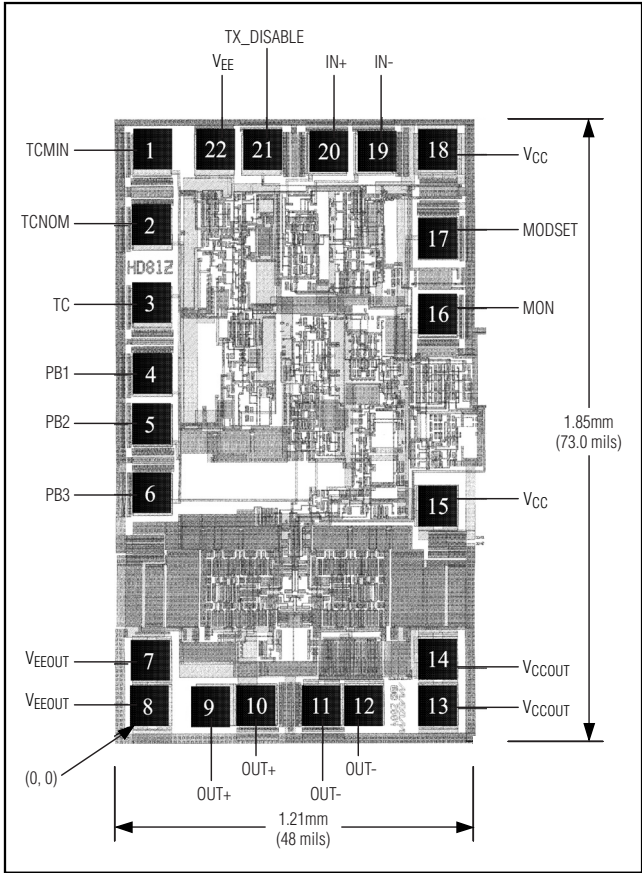
Typical Operating Circuits (continued)

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Chip Topography

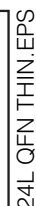


Pad Coordinates

PAD NUMBER	PAD NAME	COORDINATES (μm)	
		X	Y
BP1	TCMIN	0	1464
BP2	TCNOM	0	1268
BP3	TC	0	1060
BP4	PB1	0	876
BP5	PB2	0	744
BP6	PB3	0	560
BP7	VEEOUT	0	116
BP8	VEEOUT	0	0
BP9	OUT+	180	0
BP10	OUT+	296	0
BP11	OUT-	480	0
BP12	OUT-	596	0
BP13	VCCOUT	804	0
BP14	VCCOUT	804	124
BP15	VCC	804	528
BP16	MON	804	1032
BP17	MODSET	804	1240
BP18	VCC	804	1464
BP19	IN-	624	1464
BP20	IN+	492	1464
BP21	TX_DISABLE	308	1464
BP22	VEE	176	1464

MAX3967A

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)

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Package Information(continued)

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to www.maxim-ic.com/packages.)

COMMON DIMENSIONS															
PKG	12L 4x4			16L 4x4			20L 4x4			24L 4x4			28L 4x4		
REF.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.
A	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80
A1	0.0	0.02	0.05	0.0	0.02	0.05	0.0	0.02	0.05	0.0	0.02	0.05	0.0	0.02	0.05
A2	0.20 REF			0.20 REF			0.20 REF			0.20 REF			0.20 REF		
b	0.25	0.30	0.35	0.25	0.30	0.35	0.20	0.25	0.30	0.18	0.23	0.30	0.15	0.20	0.25
D	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10
E	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10	3.90	4.00	4.10
e	0.80 BSC.			0.65 BSC.			0.50 BSC.			0.50 BSC.			0.40 BSC.		
k	0.25	-	-	0.25	-	-	0.25	-	-	0.25	-	-	0.25	-	-
L	0.45	0.55	0.65	0.45	0.55	0.65	0.45	0.55	0.65	0.30	0.40	0.50	0.30	0.40	0.50
N	12			16			20			24			28		
ND	3			4			5			6			7		
NE	3			4			5			6			7		
JEDEC	VGG3			VGGC			WGGD-1			WGGD-2			WGGE		

EXPOSED PAD VARIATIONS							
PKG. CODES	D2			E2			DOWN BONDS ALLOWED
	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	
T1244-2	1.95	2.10	2.25	1.95	2.10	2.25	NO
T1244-3	1.95	2.10	2.25	1.95	2.10	2.25	YES
T1244-4	1.95	2.10	2.25	1.95	2.10	2.25	NO
T1644-2	1.95	2.10	2.25	1.95	2.10	2.25	NO
T1644-3	1.95	2.10	2.25	1.95	2.10	2.25	YES
T1644-4	1.95	2.10	2.25	1.95	2.10	2.25	NO
T2044-1	1.95	2.10	2.25	1.95	2.10	2.25	NO
T2044-2	1.95	2.10	2.25	1.95	2.10	2.25	YES
T2044-3	1.95	2.10	2.25	1.95	2.10	2.25	NO
T2444-1	2.45	2.60	2.63	2.45	2.60	2.63	NO
T2444-2	1.95	2.10	2.25	1.95	2.10	2.25	YES
T2444-3	2.45	2.60	2.63	2.45	2.60	2.63	YES
T2444-4	2.45	2.60	2.63	2.45	2.60	2.63	NO
T2844-1	2.50	2.60	2.70	2.50	2.60	2.70	NO

NOTES:

1. DIMENSIONING & TOLERANCING CONFORM TO ASME Y14.5M-1994.
2. ALL DIMENSIONS ARE IN MILLIMETERS. ANGLES ARE IN DEGREES.
3. N IS THE TOTAL NUMBER OF TERMINALS.
4. THE TERMINAL #1 IDENTIFIER AND TERMINAL NUMBERING CONVENTION SHALL CONFORM TO JEDEC 95-1 SFP-012. DETAILS OF TERMINAL #1 IDENTIFIER ARE OPTIONAL, BUT MUST BE LOCATED WITHIN THE ZONE INDICATED. THE TERMINAL #1 IDENTIFIER MAY BE EITHER A MOLD OR MARKED FEATURE.
5. DIMENSION b APPLIES TO METALLIZED TERMINAL AND IS MEASURED BETWEEN 0.25 mm AND 0.30 mm FROM TERMINAL TIP.
6. ND AND NE REFER TO THE NUMBER OF TERMINALS ON EACH D AND E SIDE RESPECTIVELY.
7. DEPOPULATION IS POSSIBLE IN A SYMMETRICAL FASHION.
8. COPLANARITY APPLIES TO THE EXPOSED HEAT SINK SLUG AS WELL AS THE TERMINALS.
9. DRAWING CONFORMS TO JEDEC MO220, EXCEPT FOR T2444-1, T2444-3, T2444-4 AND T2844-1.
10. MARKING IS FOR PACKAGE ORIENTATION REFERENCE ONLY.
11. COPLANARITY SHALL NOT EXCEED 0.08mm
12. WARPAGE SHALL NOT EXCEED 0.10mm
13. LEAD CENTERLINES TO BE AT TRUE POSITION AS DEFINED BY BASIC DIMENSION "e", ± 0.05 .

-DRAWING NOT TO SCALE-

	
TITLE PACKAGE OUTLINE, 12, 16, 20, 24, 28L THIN QFN, 4x4x0.8mm	
APPROVAL	DOCUMENT CONTROL NO. 21-0139
REV. D	2/2

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